UNIVERSIDADE DE BRASÍLIA FACULDADE DE TECNOLOGIA DEPARTAMENTO DE ENGENHARIA ELÉTRICA

SPATIO-TEMPORAL PREDICTION OF ELECTRIC POWER SYSTEMS INCLUDING EMERGENT RENEWABLE ENERGY SOURCES

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DISSERTAÇÃO DE MESTRADO EM ENGENHARIA ELÉTRICA

PUBLICAÇÃO: 554/2014 DM PPGEE

BRASÍLIA/DF: MARÇO - 2014.

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DISSERTAÇÃO DE MESTRADO SUBMETIDA AO DEPARTAMENTO DE ENGENHARIA ELÉTRICA DA FACULDADE DE TECNOLOGIA DA UNIVERSIDADE DE BRASÍLIA, COMO PARTE DOS REQUISITOS NECESSÁRIOS PARA A OBTENÇÃO DO GRAU DE MESTRE EM EN-GENHARIA ELÉTRICA.

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BRASÍLIA-DF, 12 DE MARÇO DE 2014.

FICHA CATALOGRÁFICA

MILANEZI JR, JAYME				
FOLHA CATALOGRÁFICA - Spatio-Tem	poral Prediction of Electric Power			
Systems including Emergent Renewable Ener	Systems including Emergent Renewable Energy Sources.			
	[Distrito Federal] 2014.			
xxviii, 146p., 297mm (ENE/FT/UnB, Mes	tre, Engenharia Elétrica, 2014).			
Dissertação de Mestrado - Universidade de	Brasília.			
Faculdade de Tecnologia.				
Departamento de Engenharia Elétrica.				
1. Predição de Séries Temporais	2. Sistemas de Potência			
3. Fontes Alternativas de Energia	4. Processamento de Sinais			
I. ENE/FT/UnB	II. Publicação 554/2014 DM PPGEE			

REFERÊNCIA BIBLIOGRÁFICA

MILANEZI JR., J. (2014). Spatio-temporal Prediction of Electric Power Systems including Emergent Renewable Energy Sources. Dissertação de Mestrado em Engenharia Elétrica, Publicação 554/2014 DM PPGEE, Departamento de Engenharia Elétrica, Universidade de Brasília, Brasília, DF, 146p.

CESSÃO DE DIREITOS

NOME DO AUTOR: Jayme Milanezi Junior.

TÍTULO DA DISSERTAÇÃO DE MESTRADO: Spatio-temporal Prediction of Electric Power Systems including Emergent Renewable Energy Sources. GRAU / ANO: Mestre / 2014

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Jayme Milanezi Junior SQN 409, bloco I, CEP 70857-090 Brasília - DF - Brasil. Não posso imaginar que uma vida sem trabalho seja capaz de trazer qualquer espécie de conforto. A imaginação criadora e o trabalho, para mim, andam de mãos dadas; não retiro prazer de nenhuma outra coisa.

(Sigmund Freud)

DEDICATÓRIA

À minha esposa, Gal, e aos nossos futuros filhos.

AGRADECIMENTOS

À minha esposa, Maria Galberianny, que, com amor, companheirismo e verdadeira doação, tolerou meu frequente distanciamento em virtude dos numerosos trabalhos que empreendi durante este mestrado. A você, que sempre confiou na minha capacidade, me apoiando e fortalecendo a minha moral para este bom combate, agradeço de todo o meu coração. Amo você.

Agradeço ao meu orientador e ex-colega de profissão, Prof. João Paulo, com cuja amizade conto desde os tempos em que ombreamos no Instituto Militar de Engenharia (IME). Enquanto Professor, sempre me dispensou inestimável ajuda no âmbito do mestrado, tendo me apresentado um instigante e desafiador campo de pesquisa, e dirimindo várias de minhas dúvidas em tempo real. Pelo exemplo e incentivo incansáveis, sou-lhe extremamente grato.

Aos meus colegas na Agência Nacional de Energia Elétrica (ANEEL), principalmente aos meus companheiros na Superintendência de Pesquisa e Desenvolvimento e Eficiência Energética (SPE), por terem prestigiado os meus esforços ao longo da preparação desta dissertação. Ao Superintendente, Máximo Luiz Pompermayer, agradeço pela confiança em que o meu trabalho na SPE prosseguiria sem maiores percalços, mesmo com o desenvolvimento concomitante das tarefas relacionadas ao mestrado. Igualmente, gostaria aqui de deixar meus agradecimentos a Aurélio Calheiros Melo Junior, Assessor da SPE, por ter sempre demonstrado apoio e incentivo às tarefas acadêmicas, e aos demais colegas de sala e de toda a ANEEL, que foram grandes amigos, verdadeiros vetores de renovação e bom humor ao longo desta jornada. Um especial agradecimento a Lucas Dantas Xavier Ribeiro, que veio a se tornar um fundamental colaborador em alguns tópicos da presente dissertação, auxiliando-me com algumas sintaxes de MATLAB e na organização e processamento de dados em sua distribuição espacial. Muito obrigado a todos vocês.

Ao meu coorientador, Prof. José Antônio Alves Gomes, do Instituto Nacional de Pesquisas da Amazônia (INPA), por ter me apoiado com o arcabouço teórico relacionado a um importante vetor energético que estamos aqui apresentando como uma possível fonte renovável, as enguias elétricas. À pesquisadora Renata Schmidt, da Universidade Federal do Amazonas (UFAM), pela colaboração dada juntamente com o Prof. José Gomes, e ao Prof. José Luiz Cerda-Arias, que nos forneceu os dados de 29 subestações elétricas da cidade de Leipzig - Saxônia, Alemanha, também dedico meu sincero reconhecimento. Deixo aqui também um obrigado a Ronaldo Sebastião Ferreira Junior, que me prestou grande auxílio na concepção de alguns circuitos para a parte de reciclagem de ondas rádio e na própria parte de reciclagem de luz indoor.

Aos meus colegas em inovações tecnológicas e matérias de mestrado Marco Antônio Marques Marinho, Edison Pignaton, Antonio Rubio Serrano, Danilo Fernandes, Bruno Pilon, Ricardo Kehrle e Stephanie Alvarez, agradeço pela parceria, seja nos estudos em conjunto, seja naqueles debates teóricos sobre os modelos estudados, cujas discussões sempre foram da maior relevância, sempre por demais construtivas. Ao Prof. Ricardo Zelenovsky e a Fábio Stacke Silva, integrantes da banca avaliadora, sou grato pelas valorosas revisões do texto final desta dissertação, as quais tornaram o texto de leitura sem dúvida mais aprazível.

Aos meus pais, Jayme e Lúcia, que, com amor e paciência, me deram desde cedo o ensinamento de que o estudo, a honestidade e a superação pessoal moldam os únicos caminhos que edificam e valorizam permanentemente nossas vidas. Aos meus irmãos Pollyanna, Marcos e Thiago, pelo suporte emocional e conversas revigorantes. Vocês são um grande esteio de carinho e amparo na minha vida.

Agradeço, por fim...

... a Deus, se ele existir.

RESUMO

PREDIÇÃO ESPACIAL TEMPORAL DE SISTEMAS ELÉTRICOS DE POTÊNCIA INCLUINDO FONTES RENOVÁVEIS EMERGENTES

Autor: Jayme Milanezi Junior Orientador: João Paulo Carvalho Lustosa da Costa Coorientador: José Antônio Alves Gomes Programa de Pós-graduação em Engenharia Elétrica Brasília, março de 2014

A atividade de planejamento de sistemas de potência inclui, como um de seus maiores desafios, a predição do comportamento da carga. Com a finalidade de otimizar o investimento ante os dados de consumo, as empresas do setor elétrico lançam mão de várias técnicas de previsão da evolução da demanda que devem atender. No presente trabalho, o tema da predição espacial e temporal da carga é enfrentado, estudando e incorporando, simultaneamente, a tendência hoje já observada de inclusão de fontes em microgeração distribuída. Três fontes renováveis e emergentes de geração foram consideradas como geradoras de energia pelos consumidores: enguias elétricas, painéis fotovoltaicos para aproveitamento da luz solar e de interiores, e antenas para reciclagem da energia existente nas ondas eletromagnéticas de radiodifusão. Quatro métodos preditivos foram empregados para prever o comportamento da carga: modelo Auto-Regressivo (AR), Auto-Regressivo com Variável eXógena (ARX), Auto-Regressivo de Média Móvel com Variável eXógena (ARMAX) e Redes Neurais Artificiais (ANN). Os dados de consumo foram as máximas demandas semanais registradas em 8 Subestações da cidade de Leipzig (Saxônia, Alemanha), durante os anos de 2001, 2002, 2003 e 2004. O dado exógeno considerado foi a temperatura, em valores diretos e logarítmicos. Das 209 semanas existentes entre 2001 e 2004, as 200 primeiras destinaram-se ao ajuste dos coeficientes nos modelos AR e ao treinamento da rede neural; as 9 semanas restantes foram destinadas à comparação de resultados. A aplicação das técnicas deu-se, assim, em dois estágios: no primeiro, os dados reais da rede de Leipzig foram considerados, e no segundo estágio trabalhou-se com novos valores de demandas máximas, originadas pela inserção de valores hipotéticos de energia recebida das três fontes citadas. Em ambos os estágios, o modelo ARMAX foi o de melhor precisão na previsão de dados. O sistema de redes neurais demonstrou ser um sistema sub-ótimo de previsão.

Palavras Chave: Predição de Séries Temporais, ARMAX, Redes Neurais, Geração por Enguias Elétricas, Reciclagem de Energia.

ABSTRACT

SPATIO-TEMPORAL PREDICTION OF ELECTRIC POWER SYSTEMS INCLUDING EMERGENT RENEWABLE ENERGY SOURCES

Author: Jayme Milanezi Junior Supervisor: João Paulo Carvalho Lustosa da Costa Co-supervisor: José Antônio Alves Gomes Programa de Pós-graduação em Engenharia Elétrica Brasília, March of 2014

Power systems planning activities include load behavior prediction as one of its most challenging tasks. In order to optimize investments related to consumption data, utilities from the Electrical Sector resort to several forecasting techniques so that they can predict the power demand which these utilities must support. Along the present work, issues related to the spatial and temporal predictions are faced, considering, simultaneously, the observed trend of microgeneration spread. Three emergent renewable sources were proposed to be taken on by consumers: electric eels, photovoltaic solar panels for outdoor generation and indoor light energy harvesting, and antennas for radio frequency energy recycling. Four predictive methods were employed in order to forecast load evolution: Auto-Regressive (AR), Auto-Regressive with eXogeneous inputs (ARX), Auto-Regressive Moving Average with eXogeneous inputs (ARMAX) models and Artificial Neural Networks (ANN). Consumption data were the maximum weekly power demands registered over 8 Power Substations from the city of Leipzig (Saxony, Germany), during the years 2001, 2002, 2003 and 2004. The exogeneous variable adopted was temperature, in realistic and in logarithmic values. During the 209 weeks which are comprised between 2001 and 2004, the first 200 weeks served to coefficients adjustments, with regards to AR models, and the training of the neural network, in the case of ANN. The last 9 weeks were destinated for results comparison. Techniques were undertaken in two stages: firstly, only realistic data from Leipzig Substations were considered, and in the second stage, new values for maximum power demands were obtained by means of simulations upon the three emergent sources. In both stages, ARMAX model returned the fittest results, whereas ANN characterized itself as a sub-optimal prediction system.

Keywords: Time Series Prediction, ARMAX, Artificial Neural Networks, Electric Eels Generation, Energy Recycling.

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List of Symbols, Nomenclatures and Abbreviations

 a, a_n : Logarithmic value of annual increasing rate

 A_{em} : Maximum effective irradiation area

AM: Air Mass

ANATEL: Agência Nacional de Telecomunicações

ANEEL: Agência Nacional de Energia Elétrica

ANN: Artificial Neural Networks

AR: Auto-Regressive Model

ARMAX: Auto-Regressive Moving Average with eXogeneous Elements

ARMAX-AIC: ARMAX Akaikepsilas Information Criterion

ARMAX-FPE: ARMAX Akaikepsilas Final Prediction Error

ARMAX-NSSE: ARMAX Normalized Sum of Squared Error

ARX: Auto-Regressive with eXogeneous Elements Model

ARX log[T]: Auto-Regressive with eXogeneous Elements Model; the exogeneous element, or input, is logarithmic values of minimum weekly temperatures.

a-Si: Amorphous Silicon

A(z): z-polynomyal transfer function related to the AR part, whose input is past values of Y(z)

BHJ: Bulk Heterojunctions

B(z): z-polynomyal concerning to the eXogeneous part

 C_1, C_2 : Capacitors for the boost converter circuit

CCFL: Cold-Cathode Fluorescent Lamp

CFL: Compact Fluorescent Lamp

 $c_{i,k}$, $\cos(\theta_{i,k})$, $\rho_{X_iX_k}$: Sample correlation coefficient, or correlation coefficient

 c_n , c: Initial amount of the unit generation: antenna, eel or square meter of PV panel Cov $\{X, Y\}$, σ_{XY} : Covariance of X and Y

c-Si: Crystalline Silicon

 $c_{x,y}, \cos(\theta_{x,y}), \rho_{xy}$: Correlation coefficient of X and Y

C(z): z-polynomyal concerning to the MA part

 D_0 : Maximum directivity considering the irradiation pattern of the antenna

 D_1, D_2, D_3 : Diodes for the boost converter circuit

 d_1 : Realistic power demand void of generating sources (example)

 d_2 : Realistic power demand with generating sources (example)

DC: Direct Current

 d_i, d_j : Deviation matrix or deviation vector

DTV: Digital Television

e: natural logarithmic basis: e = 2.71828182846...

 e^{α} : absolute value of z

 $e_c d$: Iradiation efficiency of the antenna

e(t), e(n): noise signal values

 $E\{u[n]\}, \mu_n, \mu(n), \overline{u}[n]$: Mean or Expected Value of n

FEM: Forecasted Error Method

 $f \star f$: Autocorrelation function of the PMF f

 $f \star g$: Cross-correlation function of the PMFs f and g

f * g: Convolution function of the PMFs f and g

FIR: Finite Impulse Response filter

 $F_{\rm s}$: Switching frequency of the converter

F(z): z-transform of the function f(x)

 g, g_t : Power microgeneration actuating on the t-th week

 G_r : Gain of the receiver antenna

GSM: Global System for Mobile communications

 G_t : Gain of the transmitter antenna

G(z): transfer function related to the eXogeneous part, whose input is U(z)

H(z): transfer function related to the MA part, whose input is W(z)

IEEE: Institute of Electrical and Electronics Engineers

IIR: Infinite Impulse Response filter

 I_{\max} : Maximum current

 I_{out} : Current flowing over the output terminals of the circuit

 $I_{\rm sc}$: Short Circuit Current

k: noise summed to an originally pure ascendant exponential, conferring a step-shape for it

 K_{ind} : Estimated coefficient for the inductor ripple current relative to the maximum output current

L: Boost converter inductance

LED: Light Emitter Diode

LFCE-INPA: Laboratório de Fisiologia Comportamental e Evolução do Instituto Nacional de Pesquisas da Amazônia

log[T]: Logarithmic value of the minimum registered temperature for a week

 $L_{\mathbf{x}}$: Light intensity, in Lux

MDL: Minimum Description Length

MPP: Maximum Power Point

 M_{pv} : Number of square meter of PV panels generating power

 M_{pv}^{n} : Number of square meter of PV panels generating power in the n - th week

MSE: Minimum Square Error

 n_a : Number of past output values which are input of the AR filter

 n_b : Number of past eXogeneous values which are input of the X filter

 n_c : Number of past noise values which are input of the MA filter

 n_e : Number of eels

 n_k : Number of delay instants betwen the forecasted moment and the first exogeneous value in the future

NPD: Number of Positive Deviations

OECD: Organization for Economic Cooperation and Development

OPV: Organic Photovoltaic Cell

P3HT: Poly-3-hexyl Thiophene

PCBM: Phenyl-C₆₁-Butyric Acid Methyl Ester

PDF: Probability Density Function

 P_e : Amount of power produced by the eels

 P_{e-peak} : Power delivered by the batteries only during the peak

PMF: Probability Mass Function

 $P_{\mu g}$: Power microgeneration, corresponding to the difference between d_1 and d_2

 P_r : Effective power in the receiver antenna

PS: Power Substation

PSs: Power Substations

 P_t : Power transmitted by the transmitter antenna

 P_{tm} : Maximum power available from a transmitter antenna

PV: Photovoltaic

 q^{-1} : delay-shift operator, such that $y(t).q^{-1} = y(t-1)$

 R^2 : Determination Coefficient

 R_a : Water resistance

RF: Radio Frequency

ROC: Region of Convergence

 R_{XX} : Correlation between dimensions of vector X

 $r_{X,Y}(t,t+\tau), \sigma_{XY}(\tau)$: Cross-covariance function between X(t) and Y(t), lag of τ

STC: Standard Test Conditions

T: Temperature, or the minimum registered temperature for a week

t: Time, in weeks, from the moment in which the first source unit started to generate

UMTS: Universal Mobile Telecommunications System

u(n), u(t), u: deterministic signal values

U(z): deterministic input data

Var $\{X\}, \sigma_X^2$: Variance of X

 $V_{\rm i}, V_{\rm in}$: Voltage along the input terminals of the circuit

 V_{inmin} : Minimum voltage along the input terminals of the circuit

 V_{max} : Maximum voltage

 $V_{\rm oc}$: Open Circuit Voltage

 $V_{\rm out}$: Voltage along the output terminals of the circuit

w: Number of weeks in which generation is present; w is counted backwards, from the last week of 2004 to the past

 W_i : Power density of the incident wave

Wp: Watt peak

w(t): noise input data

W(z): z-transform of noise input data

X: Random variable X

X(t): Stochastic process which associates the observed values of the random variable X in the time-domain

X(z): z-transform of x(t)

Y: Random variable Y

 $\hat{y}(k)$: predicted value of y(k)

y(n), y(t), y: output signal values

Y(z): output data

Z: Random variable Z

 Z_t : Output values of the filter $\psi(B)$

 $\alpha, \beta, \gamma, \kappa$: Scalar constants

 β : angle of z

 Γ : Impedance loss factor of the antenna

 $\Delta V_{\rm out}$: Ripple voltage requirement

 $\delta(n)$: positive values of k into the algorithm; whether δ is negative, k is not sumed to the exponential

 $\epsilon(k,\theta) :$ Forecasted error taking into account the moment of prediction k and the parameter vector θ

 θ : parameter vector

 $\theta_{i,k}$: angle formed between deviation vectors d_i and d_k

- λ : wavelength of the RF wave
- π : 3.1415926535...
- $\hat{\rho}$: Polarization mismatch factor
- Σ_X : Covariance matrix of vector X
- $\Sigma_{XY}(\tau), r_{XY}(\tau)$: Cross-covariance matrix considering X(t) and Y(t) with the lag τ
- $\tau :$ Time-shifting constant
- $\varphi(k)$: regression vector
- $\psi_1, \psi_2, \ldots, \psi_n$: weights of the transfer function $\psi(B)$
- $\psi(B){:}$ transfer function with the advance-shift B operator

Chapter 1 INTRODUÇÃO

1.1 Motivação

A energia elétrica não só está intimamente ligada ao desenvolvimento humano, riqueza e conforto; ela pode ser entendida como a própria base e expressão de tais conceitos, até porque ela é assim vista no campo da análises econômicas. A vida do homem foi radicalmente modificada pelo advento da eletricidade, e não é mais possível viver satisfatoriamente sem ela.

Por todo o mundo, grandes parcelas da energia total gerada são dedicadas aos setores da indústria e transporte, que em muitas economias são os setores energeticamente mais intensivos. Esta é uma simples e efetiva explicação de por que os países mais ricos apresentam os maiores índices de consumo energético per capita. De acordo com [1], os 30 países desenvolvidos que integram a OCDE são historicamente os maiores consumidores de energia. Adicionalmente a isto, temos que as taxas de crescimento do Produto Interno Bruto (PIB) são dados intrinsecamente energéticos: em [2], é destacada uma relação quase linear entre o crescimento do consumo energético mundial e o crescimento do PIB do planeta, a partir de uma análise sobre os anos de 2003 a 2007. Intensidade energética, indicador que diz quanta energia é necessária para produzir um crescimento no PIB, é mais alta em países que têm maiores porções da população com acesso a itens de consumo energeticamente intensivos [3].

Apesar das modernas e questionáveis teorias sobre uma suposta nova relação entre energia e desenvolvimento econômico, todas as evidências aqui apresentadas apontam para uma inegável conexão entre prosperidade e consumo energético. Segundo [4], a eletricidade, no Brasil, respondeu por 17 % do consumo final energético total no ano de 2012. A mesma fonte menciona a evolução de alguns indicadores da economia brasileira ao longo do período 2003-2012: durante este decênio, a Oferta Interna de Energia (OIE), PIB e população em milhões de habitantes evoluíram segundo dados da Tabela 1.1:

Na Tabela 1.1, **tep** significa *tonelada equivalente de petróleo*, unidade de energia equivalente a 11.63 MWh. A oferta energética para o mercado interno, que é um dado per

Table 1.1: Oferta Interna de Energia (OIE), Produto Interno Bruto (PIB) e População, de 2003 a 2012, no Brasil [4]

	Unidade	2003	2004	2005	2006	2007	2008	2009	2010	2011	2012
OIE	10^6 tep	201.9	213.4	218.7	226.3	237.8	252.6	243.9	268.8	272.3	283.6
PIB	10^9 US \$	1426.1	1507.5	1555.2	1616.7	1715.2	1803.9	1797.9	1933.4	1986.2	2003.5
População	10^{6} hab.	176.6	178.7	180.8	182.9	185.0	187.2	189.4	191.6	193.2	194.7
OIE/PIB	$tep/10^3$ US\$	0.142	0.142	0.141	0.140	0.139	0.140	0.136	0.139	0.137	0.142
OIE/capita	tep/hab.	1.143	1.194	1.210	1.238	1.285	1.350	1.288	1.403	1.410	1.457

capita, costumeiramente aumenta mesmo com o aumento da população e, além disso, a OIE se mantém em consonância com a evolução do PIB. Devido à crise econômica global que teve seu clímax em 2008, que trouxe à tona consequências para as taxas do PIB ao redor do planeta, PIB e OIE para 2009 ficaram abaixo da tendência observada. Apesar disso, a razão OIE/PIB permeneceu aproximadamente constante durante todos os anos, mesmo com tais perturbações oriundas da economia, como demonstrado para esta década, uma vez que a amplitude da variação máxima pouco superou 4% da média de 0.140. Isto confirma o conhecido princípio pelo qual o desenvolvimento econômico está profundamente relacionado ao crescimento da oferta energética.

Sendo a eletricidade uma das principais formas de energia, ela e a oferta de energia total devem variar da mesma maneira. E, uma vez que os benefícios econômicos são esperados apenas quando a oferta de energia interna aumenta, a eletricidade pode igualmente servir de referencial para o aumento desse benefício. Desta maneira, a necessidade de se compreender o gerenciamento da energia elétrica surge como um tema crucial, tanto para a Economia quanto para a Engenharia.

Para ser consumida, a eletricidade deve ser gerada, ter sua tensão transformada e ser transportada por enormes distâncias, especialmente extensas em países como o Brasil, cujos centros de geração tradicionais estão situados muito longe daqueles de consumo. A eletricidade é uma forma de energia bastante sensível à infraestrutura. A potência elétrica total que as linhas de transmissão podem suportar por quilômetros é fundamentamentalmente dependente dos parâmetros destas linhas, como as propriedades nominais dos transformadores, por exemplo. E, uma vez que as linhas de distribuição até mesmo em centros urbanos acarretam pesados investimentos em infraestrutura, erros ou impropriedades no planejamento da expansão de tais ativos podem gerar perdas econômicas por anos a fio. A necessidade de manter a capacidade de transporte e distribuição instantânea sempre acima dos valores consumidos justifica a microgeração distribuída como uma decisão vantajosa, em face da oferta incremental da geração no sistema como um todo, tal como será discutido na Seção 1.2 e demonstrado no Capítulo 4. Em muitas situações, microgeradores possibilitam à carga assumir um padrão de consumo graficamente mais achatado, ou seja, com menor variância nos valores de potência demandada, dado que esses microgeradores sustentam parte da carga no horário de pico, em que o valor unitário do kWh é mais caro.

Além do mais, microgeradores contam agora com um ambiente favorável para prosperar em países que permitem que consumidores forneçam para a rede o excedente de sua produção. No Brasil, desde a publicação da Resolução Normativa nº 482, da Agência Nacional de Energia Elétrica (ANEEL), em abril de 2012, consumidores particulares podem produzir energia, sendo reembolsados pelas parcelas que eles cederem à rede em suas próximas contas de energia. Esse aspecto permite-nos imaginar a microgeração distribuída como uma tendência, ao menos no mercado brasileiro.

Ao dedicarmos igual atenção ao consumo e à microgeração distribuída, nós cobrimos um rol mais extenso de possibilidades futuras, o que aumenta nossa capacidade de prever os investimentos necessários a serem realizados sobre a rede em casos concretos. Esse é o principal interesse do presente estudo, que provê uma forma de minimizar erros relativos a quanto e onde os ativos das linhas de potência devem ser expandidos. Consideramos o problema da exatidão na previsão dos valores futuros de demanda em ambientes urbanos, de forma a possibilitar a predição ótima no consumo futuro nos domínios temporal e espacial.

1.2 Premissas e Objetivos

Baseado na descrição acima sobre soluções energéticas, formulamos agora um esquema visual relacionado às contribuições do presente trabalho. A Fig. 1.1 mostra um exemplo simplificado do atual padrão de rede, no qual existem vários geradores de grande porte que estão, na maioria das vezes, distantes dos maiores centros de consumo. Existe um considerável custo associado ao transporte de grandes blocos de energia, tal qual aqui ilustrado. Apesar de a produção em si ter custo eminentemente baixo em face da fonte considerada - a hidrelétrica -, o custo de transportar potência ao longo de várias centenas de quilômetros é por vezes significativo. E, à medida que a energia transportada vai chegando aos ambientes urbanos, milhares de ramificações surgem dado que a rede vai se capilarizando. Considerando os ativos dedicados ao transporte e distribuição, o investimento total a fazer em infraestrutura energética é proporcional à complexidade da rede, às distâncias envolvidas e, acima de tudo, ao total de potência entregue pela fonte.



Figure 1.1: Esquema reduzido do sistema elétrico, compreendendo geradores hidrelétricos distantes, longas linhas de transmissão, distribuição e consumo

A Fig. 1.2 ilustra uma situação bastante similar à da Fig. 1.1, mas agora com vários atores de microgeração espalhados pelo sistema. Ao longo do presente estudo, três vetores energéticos são considerados como sendo adotados pelos consumidores de uma determinada cidade. Referir-nos-emos a cada uma destes vetores como "fontes", pois, até no caso da reciclagem energética, supõe-se que os vetores envolvidos gerem um quantum de potência extra dentro do balanço energético geral. Não consideraremos, na presente dissertação, soluções energéticas tradicionais, mas fontes renováveis verdadeiramente emergentes, como as enguias elétricas, captação da energia das ondas de rádio (RF) e reciclagem de energia de luz em ambientes fechados por meio de painéis fotovoltaicos.

A elevação da oferta geral de energia provavelmente terá, como consequência, a postergação dos investimentos atinentes à ampliação dos ativos da cadeia elétrica de suprimento. Fig. 1.3 mostra o que significa esta nova realidade. Como dito, se os consumidores não produzirem sequer uma parcela do consumo de sua carga, a totalidade do esforço recairá sobre as empresas do setor. Por outro lado, caso este consumidor produza ao menos uma pequena parte de sua própria energia, a maior parte do peso de manter



Figure 1.2: Esquema da Fig. 1.1, agora incluindo enguias elétricas nadando em um rio e recicladores RF sobre as casas, ambos fornecendo energia para o sistema elétrico interligado, enquanto um analisador de dados reúne os dados de consumo das Subestações envolvidas



Figure 1.3: Comparação entre as redes com e sem microgeradores distribuídos: a pessoa representa os geradores, a corda é comparada com as linhas de transmissão, um bloco pesado representa a carga, e o contrapeso auxiliar representa a microgeração: (a) a pessoa efetua o esforço sozinha, e (b) o esforço de sustentar a carga é parcialmente dividido entre ela e o contrapeso auxiliar.

o funcionamento do sistema ainda recai sobre os ativos tradicionais, mas agora este esforço resta diminuído.

A ideia central da reciclagem de energia, assim como a da própria microgeração distribuída, é a de aliviar a rede de alguns investimentos no curto prazo, ao mesmo tempo em que diminui o risco de insuficiência do suprimento. Quanto maior o contrapeso auxiliar na Fig. 1.3(b), menor será a força requerida da pessoa puxando a carga; ainda, a corda suportará com maior segurança os padrões de tração, e a polia ficará em condições de girar sem grandes problemas.

As contribuições do presente trabalho são baseadas em prospectar dois tópicos principais: investigação sobre os efeitos da microgeração distribuída a ser inserida por toda a rede, e a análise da predição deste consumo em séries temporais. Ambos os objetivos são agora separadamente descritos.

O primeiro objetivo desta pesquisa está relacionado com a avaliação energética da microgeração a ser espalhada pela rede. Duas formas emergentes de reciclagem de energia são combinadas com uma nova fonte de energia renovável. A primeira tecnologia de reciclagem é relativa ao aproveitamento da energia fotovoltaica da luz em ambientes fechados, por meio de painéis instalados no interior de edificações. O segundo reciclador de energia aproveita a potência eletromagnética das Rádio Frequências (RF) existente no espaço, com ênfase à sua exploração em ambientes predominantemente urbanos. Podemos dizer que ambos os tipos consistem em reciclagem de energia proveniente do espectro electromagnético - o primeiro reciclando a luz, e o segundo, ondas rádio. O terceiro elemento analisado, que é uma fonte de energia em si e não um vetor de reciclagem, é a enguia elétrica, ou poraquê. Este peixe de água doce é muito observado na Amazônia brasileira e converte a energia química de seu alimento em tensão elétrica ao longo de seu corpo. Estes três vetores energéticos são avaliados tecnicamente, sendo estimadas as possíveis contribuições energéticas de cada uma delas para a rede.

O segundo objetivo desta dissertação é a análise da predição de séries espaciais-temporais, que contam com métodos matemáticos úteis em predizer valores de séries temporais. Da cidade de Leipzig (Saxônia, Alemanha) obtivemos dados diários de consumo ao longo de 4 anos, envolvendo 8 Subestações de potência (SE). A razão de adotar Leipzig como a cidade de análise neste trabalho foi a disponibilidade de um preciso e coeso conjunto de dados diários. Quatro métodos foram testados de forma a tornar possível a identificação do mais adequado para prever a informação: modelo Auto-Regressivo (AR), Auto-Regressivo com um elemento eXógeno (ARX), Auto-Regressivo de Médias Móveis com um elemento eXógeno (ARMAX) e Redes Neurais Artificiais (ANN). Os três primeiros métodos são modelos Box & Jenkins. O elemento exógeno presente nos modelos ARX e ARMAX foi a temperatura: primeiramente, seu valor real foi adotado, e em seguida foram incluídos seus valores logarítmicos.

Na sequência, dois principais estágios foram empreendidos, sendo a microgeração distribuída considerada apenas no segundo estágio. Todos estes passos buscaram determinar qual o método mais preciso para predição de valores futuros: AR, ARX, ARMAX ou ANN.

Previsões espaciais advieram dos resultados obtidos na predição temporal, uma vez que os futuros valores encontrados para a carga no tempo são aplicados a cada uma das 8 SEs consideradas.

Como o escopo da presente dissertação tem dois focos diferentes e harmônicos, devemos descrever os prováveis benefícios finais deste trabalho. Uma vez que os dados sobre consumo oriundos de algumas SEs estiverem disponíveis, o processamento de futuros valores de demandas é uma tarefa possível, e ainda com a indicação de qual método - se ANN, AR, ARX ou ARMAX - é o melhor para tanto. Assume-se que os dados relativos às potência demandada estão à disposição da empresa de distribuição de energia elétrica. Uma vez que esta informação está ao alcance, valores futuros de demanda podem ser determinados com razoável precisão. Os benefícios relacionados à predição dividem-se em duas frentes:

- como exemplo de aplicação no curto prazo, temos que se a empresa prevê para as próximas semanas uma elevação em uma das demandas que venha a oferecer qualquer ameaça à continuidade do suprimento, a empresa pode se prevenir contra este evento agindo sobre os gargalos. Por exemplo, consideremos uma previsão de que em duas semanas um pico de potência tem uma certa probabilidade de ocorrência de tal forma que possa se aproximar dos limites de especificação de um dos transformadores. Nesse caso, poderia ser uma solução ativar um segundo transformador na respectiva SE em vista dessa previsão específica.
- como exemplo de aplicação no médio prazo, considerando um conjunto de dados
mensais ou anuais, se a empresa verifica que, de acordo com a tendência, um grupo de transformadores deve suportar potências que não podem atender dadas suas especificações, eles tornam-se gargalos do sistema, devendo aí a rede ser expandida.

No exemplo atinente ao curto prazo, haveria ainda a possibilidade de que o sistema viesse a pagar uma taxa mais cara pelo Wh adicional gerado pelos consumidores, no caso de se tratar de uma semana crítica. Obviamente, tal decisão deve estar amparada pelas disposições regulatórias vigentes.

Dado que serão caracterizados valores plausíveis para as fontes renováveis em questão - enguias, RF e reciclagem de luz em ambientes fechados -, ganha-se informação para futuros estudos que venham a adotá-las.

1.3 Como esta Dissertação está Organizada

O Capítulo 3 inicia o estudo da dissertação propriamente dita, abordando as três fontes renováveis que serão empregadas: conversão fotovoltaica, reciclagem de ondas eletromagnéticas RF e enguias elétricas. Para cada uma destas fontes, expõe-se a teoria geral e o que há no estado da arte. As medições da potência passível de ser fornecida ajuda a moldar os circuitos de gerenciamento da energia aplicáveis.

Tanto no Capítulo 4 como no 5, um conjunto de dados de demandas de potência abarca 4 anos (2001-2004), ou 209 semanas. As máximas demandas de potência semanal basearam-se em registros pormenorizados de 8 SEs, totalizando 209 dados sequenciados para cada SE. Os capítulos 4 e 5 também mencionam como se deram as aplicações dos sistemas ANN e Box & Jenkins (AR, ARX and ARMAX) sobre os dados reais observados. Uma vez que estes capítulos procedem a cálculos relacionados com estes sistemas, o leitor deve considerar a possibilidade de visitar os Apêndices A e B caso deseje entender ou relembrar a transformada z e as estruturas AR, ARX e ARMAX.

O capítulo 4 descreve a análise preditiva da carga tal qual os dados foram apanhados das SEs de Leipzig, isto é, sem que ainda fosse considerada a inserção das fontes emergentes no sistema elétrico. Para os modelos AR, ARX e ARMAX, a ordem ótima do modelo foi selecionada a partir do critério da Descrição por Dimensões Mínimas (DDM). Os valores de temperatura e seus logaritmos foram analisados como possível variável *eXógena* (o X de ARMAX). Produzimos, então, os modelos AR, ARX e ARMAX para as últimas 9 semanas, e para cada SE. Falando agora das estruturas ANN, de redes neurais, ela foi treinada ao longo das primeiras 200 semanas, produzindo saídas para as últimas 9 semanas de dados do período total de 209 semanas. Assim sendo, em nome de uma comparação justa, utilizamos a mesma quantidade de dados de entrada e previmos a mesma quantidade de informação, usando tanto estruturas ANN quanto ARMAX. Todos os modelos foram, então, comparados uns com os outros segundo a mesma quantidade de dados.

O Capítulo 5 insere as fontes renováveis nas casas, indústrias e prédios públicos. As mesmas informações de 209 semanas (2001-2004) foram admitidas como dados de entrada. Projetamos possíveis valores energéticos oriundos das fontes para diferentes períodos ao longo das últimas 157 semanas, em instantes iniciais aleatórios. Novos valores para demandas máximas semanais foram então obtidas a partir da diferença entre os valores reais medidos e os valores calculados para cada fonte renovável. Esses novos conjuntos de valores foram, assim, adotados como o novo rol de dados de entrada. Posto que durante 2001 ainda não havia energia proveniente das fontes emergentes, mas apenas durante o período 2002-2004, e considerando que as fontes renováveis entram com energia na rede progressivamente, foram simuladas novas tendências para as máximas demandas de potência tal como enxergadas pelas SEs. Em seguida, nós treinamos a malha ANN e ajustamos os modelos AR, ARX e ARMAX de acordo com as primeiras 200 semanas. Avaliamos, então, a habilidade de adaptação de cada método no sentido de predizer os consumos semanais para as últimas 9 semanas de 2004.

A existência de microgeração distribuída não é informada para a empresa de distribuição, nas aplicações dos modelos ANN, AR, ARX e ARMAX; a empresa deve apenas detectar as novas tendências da carga e adaptar-se a elas.

Tanto no capítulo 4 como no 5, foi estabelecido que a rede neural conta com duas camadas escondidas. Nossos códigos MATLAB detectaram o melhor número possível para cada camada escondida. Portanto, as previsões via ANN foram obtidas com essa quantidade ótima de neurônios por camada.

O Capítulo 6 contém considerações sobre o experimento em geral e oferece uma conclusão objetiva com relação a qual método foi o de melhor desempenho para cada um dos experimentos. Há, ainda, dois apêndices com um útil arcabouço teórico que ampara várias decisões tomadas ao longo desta dissertação. O Apêndice A revisa Séries Estacionárias e discute conceitos atinentes à estacionariedade de séries temporais, que são requisitos para os filtros lineares utilizados nos Capítulos 3 e 4. O Apêndice B contém a Transformada Z, compreendendo modelos auto-regressivos (AR), de média móvel (MA), ARX (AR com variável exógena) e ARMAX. Por fim, o Apêndice C elenca vários resultados de predição espacial no que tange à redução percentual da carga no horário de ponta para os bairros em que se situam as Subestações de Potência. Alguns destes resultados são mostrados no final do Capítulo 5; porém, em face da imensa quantidade de exemplos possíveis a partir da variação de alguns parâmetros de crescimento das fontes, um Apêndice específico foi idealizado.

Chapter 2 INTRODUCTION

2.1 Motivation

Electric energy is not only intimately related to human development, richness and comfort; it can be understood as the basis of such concepts, even because it is as such considered in the realm of economic analyzes. Human life was radically modified by the advent of electricity, and it is no longer possible to live in a satisfying way without it.

All over the world, large parts of energy supply are set apart to industry and transport, which are in many economies the most energetic intensive sectors. That is a simple and effective explication why the richest countries have the highest energetic demands per inhabitant. According to [1], the 30 developed countries which integrate the OECD are historically the greatest energy consumers. In addition to it, growing-rates for GDP are intrinsically energetic data: in [2], it is pointed out an almost linear relationship between the worldwide increasing energetic demand and GDP growth, considered the years from 2003 to 2007. Energetic intensity, the indicator which establishes how much energy is necessary to produce GDP, is higher in countries which have larger portions of population with access to energetic intensive consumer goods [3].

Despite the modern and questionable theories about the new relationship between energy and economic development, all these evidences show the undeniable connection between wealth and energy consumption. According to [4], in Brazil, electricity stands for 17 % of the final energy consumption in 2012. The same document states, for the year 2012, the evolution of some indicators of Brazilian economy over the period 2003-2012: during these 10 years, Domestic Energy Supply (DES), Gross Domestic Product (GDP) and population in millions of inhabitants have evolved as shown in Table 2.1:

In Table 2.1, **toe** stands for *ton oil equivalent*, energy unit equivalent to 11.63 MWh. The energy supply for internal inhabitants usually increases despite of the growing population, and, moreover, the DES keeps very pace with GDP evolution. Due to the worldwide economic crisis which has had its climax in 2008, carrying forth consequences for the GDP rates all over the globe, GDP and DES for 2009 were below the expected

Table 2.1: Domestric Energy Supply (DES), Gross Domestic Product (GDP) and Population from 2003 to 2012 in Brazil [4]

	Unit	2003	2004	2005	2006	2007	2008	2009	2010	2011	2012
DES	10^6 toe	201.9	213.4	218.7	226.3	237.8	252.6	243.9	268.8	272.3	283.6
GDP	10^9 US \$	1426.1	1507.5	1555.2	1616.7	1715.2	1803.9	1797.9	1933.4	1986.2	2003.5
Population	10^6 inhab	176.6	178.7	180.8	182.9	185.0	187.2	189.4	191.6	193.2	194.7
DES/GDP	$toe/10^3$ US\$	0.142	0.142	0.141	0.140	0.139	0.140	0.136	0.139	0.137	0.142
DES/capita	toe/inhab	1.143	1.194	1.210	1.238	1.285	1.350	1.288	1.403	1.410	1.457

trend. Despite of that, DES/GDP remained approximately constant all over the years, even with the perturbation arised by the economy, as it is demonstrated in this decade, since the amplitude of maximum variation barely overcame 4% of the mean 0.140. This confirms the very known principle by which economic development is deeply related to the increase of energy offer.

As electric energy is one the main forms of energy, electricity and the overall energy supply are supposed to arise or come down together. Furthermore, since economic benefits are expected only when energy offer goes up, electricity must do the same. Therefore, the need related to understanding electricity management comes up as a crucial subject, in Economy as much as in Engineering.

In order to be consumed, electricity must be generated, transformed and transported over wide distances, which are specially wide in cases as the Brazilian, whose centers of generation are very far away from those of consumption. It consists on a form of energy much sensitive to infrastructure. The amount of electrical power transmission lines can transport over kilometers is fundamentally dependent on its physical parameters, like transformers properties, for instance. Since distribution power lines even into urban environments represent huge infrastructure investments, errors or improprieties in planning such assets expansion may breed economic losses for years. This need of keeping available power always above instantaneous consumption justifies electric power microgeneration as an advantageous measure, due to the incremental power offer to the overall balance, as discussed in Section 1.2 and demonstrated in Chapter 4. In many situations, microgenerators enable the load to assume a flatter behavior, i. e., with less variance on power demand values, given that microgenerators can sustain part of the load in the daily peak, when the value of unitary kWh is more expensive.

Furthermore, power microgenerators count nowadays on a favorable environment to prosper within countries in which consumers are allowed to furnish to the grid the surplus of their power generation. In Brazil, since the edition of the Resolution n^o 482

from Brazilian Electricity Regulatory Agency (*Agência Nacional de Energia Elétrica* - *ANEEL*, in portuguese), in April 2012, private people are allowed to produce part of their own energy, being refunded for the power delivered to the grid in their next bills. This aspect makes possible for us to imagine micropower generation as a trend, at least in the Brazilian market.

Dedicating equal attention to consumption as to microgenerated power, we cover a wider set of future possibilities, which enhances our capability of forecasting investments in real cases. This is the main concern of the present study, which provides a way of minimizing errors related to how much and where the assets of power lines must be expanded. We take into account the problem related to the accuracy on future power demands into urban environments, in order to provide the optimal prediction of future consumptions, in spatial and temporal domains.

2.2 Premises and Objectives

Based on the previous description about the mentioned renewable energy solutions, we can formulate a visual scheme related to the contributions of the present work. Fig. 2.1 depicts a simplified example of the current patterns of grid, with several concentrated generators that are, in many situations, far away from the great consumption centers. There is a considerable cost with regards to the transport of the generated electric energy. Although the production itself might be actually cheap, as in hydroelectric power station, the cost of carrying this energy over about several hundreds of kilometers is occasionally significant. Besides, as the power grid gets to the big cities' environments, a plenty of ramifications come up as the grid branches off more and more. Taking transport and distribution assets into account, the total investment to make is proportional to the complexity of the grid, the distances and, above all, the total power drawn by the load.

Fig. 2.2 illustrates a situation which is similar to that of Fig. 2.1, but now with many spread micropower generation systems. Along the present study, three energetic sources are considered to be spread over a city. We refer to each one of these technologies as "sources", because even the recycling mechanisms are supposed to provide extra power, considering the overall energetic balance. In this study we do not consider conventional renewable energy solutions, but rather emergent renewable energy solutions, such as electric eels, radio-frequency (RF) waves and indoor light energy harvesting systems.



Figure 2.1: Reduced scheme of the electric system, comprehending far away hydro-generators, long transmission lines, distribution and consumption.

The enhancement of the overall generation probably will have, as a consequence, the postponement of investments concerning to the enlargement of the assets of electric supply chains. Fig. 2.3 illustrates what would be this new reality. As said, whether consumers do not produce even a fraction of their electric load, the totality of the effort falls on the public utility. On the other hand, if this consumer produces at least a little part of their own energy, the bulk of the energetic burden continues to lie with the utility, but this effort is now shortened a little.

The core idea of recycling energy, as well as that of the distributed microgeneration, is to relieve the grid from investments in the short-term, simultaneously diminishing the risk of supply insufficiency. The higher is the auxiliary weight within Fig. 2.2(b), the softer will be the force which is required from the man; furthermore, the rope has a greater assurance in supporting the traction, as well as the pulley is more likely to roll without problems.

The contributions of the present work are based on the probing of two main topics: investigation about the effects of the micropower generation to be inserted all over



Figure 2.2: Scheme of Fig. 1.1, now including eels into the river and an RF recycler on top of houses, both delivering generating power for the interconnected power system, while a data analyzer gathers up the consumption data from the Power Substations.



Figure 2.3: Comparison of the grid with micro-generators: the man represents generators, the rope is compared to transmission lines, a heavy block represents the load, and the auxiliary counterweight stands for the renewable sources: (a) the man pulls it alone, and (b) the burden of load is partially shared between him and the auxiliary weight.

the grid, and spatio-temporal series predicition analysis. Both objects are hereafter separately described.

The first goal of this research is related to the energetic evaluation of micropower generation to be inserted all over the grid. Two emergent forms of recycling energy are combined to a novel renewable source. The first recycling technology is energy haversting from indoor light, by means of solar panels installed within buildings. The second energy recycler manages the eletromagnetic power from Radio Frequency (RF), which is spread all over the space, with emphasis to urban environments. We can say that both types consist on recycling energy of electromagnetic spectrum, the first one recycling light and the second one recycling radio frequencies. The third analyzed element, which is an energy source itself rather than a recycling scheme, is the electric eel. This freshwater fish is rife in Brazilian Amazonia and converts chemical energy existing in their food into voltage along their body. All these three vectors are technically analyzed, and we assess the possible energetic contribution from each one to the grid.

The second goal of this dissertation is the spatio-temporal series analysis, which relies on useful mathematical methods to predict values of time series. The city of Leipzig (Saxon, Germany) has furnished its electrical data over 4 years involving 8 Power Substations (PS). The reason for adopting Leipzig as our city in this study was the availability of such set of data in details.

Four methods were tested in order to make it possible identifying the more suitable one to forecasting information: Auto-Regressive (AR) Model, Auto-Regressive with an eXogeneous Input (ARX), Auto-Regressive Moving Average with ans eXogeneous Input (ARMAX) and Artificial Neural Networks (ANN). The three first methods are Box & Jenkins Models. The eXogeneous input within ARX and ARMAX was temperature: priorly, its real measured value was adopted, and then we handled its logarithmic values.

In sequel, two main stages were undertaken, microgeneration being considered only in the second stage. All these tasks were developed in order to determine the more precise method: AR, ARX, ARMAX or ANN.

Spatial forecasts arose from the temporal results, since the future values of load in time are applied to each one of the 8 considered PS.

As the scope of this work has two different and harmonic focuses, we should describe the benefits of such study in the way it is idealized. Since many data about electrical consumption of a set of Power Substations is available, the processing of the future values is possible, even with the indication of which method - if ANN or ARMAX is the best one for that. We take for grant that demand power data are within the reach of the electric distribution utility. Once this information is available, futures values of demand are determined with resonable precision. The benefits with respect to prediction comprise two fronts:

- as an example of application in the short-term, if the utility forecast any raise within future power demands, which might offer any threat to the assurance of supply, the utility can prevent such issue, acting upon bottlenecks. For instance, consider a forecast which informs us that two weeks from now a maximum value is likely to occur in such a way that it may approach the boundaries of transformer capacity. In this case, it might be a solution to turn on an existing second transformer in the Power Substation for this specific event.
- as an example of application in the mean-term, considering therefore a monthly or annual set of data, if the utility finds out that an increasing number of transformers are about to face power demands that they would not support given their specifications, these transformers become bottlenecks, and therefore the grid must be expanded.

In the short-term example, there is still the possibility of paying some more for each additional Wh generated by the consumers, whether the present week is a critical one. Obviously, such decision should be supported by the regulatory laws.

Since we are going to feature realistic values for the mentioned renewable sources eels, RF and indoor light harvesting systems -, there is a gain of information for future studies which intend to adopt these energetic sources.

2.3 How this Dissertation is Organized

Chapter 3 initiates the proper study existing in the present dissertation, approaching the three renewable sources which will be employed: photovoltaic conversion, electromagnetic waves recycling. For each source, we approach the theory and the state of the art. Measurements of deliverable power helps to determine the specifications for the applicable management circuits.

In both chapters 4 and 5, the set of power demands covers 4 years (2001-2004), or 209 weeks. Weekly maximum power demand data counted on 8 PSs detailed demand registers, performing 209 sequenced data for each PS. Chapters 4 and 5 also cover the application of ANN and Box & Jenkins (AR, ARX and ARMAX) models upon real recorded data. Since these chapters carry out calculations related only to these models, the reader must consider take a look at the Appendices A and B in order to review *z*-transform and ARMAX concepts, respectively.

Chapter 4 describes the predictive analysis of the load as they were extracted from Leipzig Power Substations, i. e., with no emergent power microgenerators being considered. For the AR, ARX and ARMAX models, the optimal order of parameters were selected via Minimum Description Length (MDL) criterion. Temperature values and their logarithm were analyzed as a possible *eXogeneous* variable, therefore ARMAX with X at the end. We have then produced AR, ARX and ARMAX models for the last 9 weeks, and for each PS. Talking now about ANN structure, we trained it over the first 200 weeks, returning outputs for the last 9 weeks of data from the entire period of 209 weeks. Therefore, in order to perform a fair comparison, we utilized the same quantity of input data and forecasted the same amount of information, either making use of ANN and ARMAX models. All models were then compared each other counting on the same set of data.

Chapter 5 inserts the renewable sources into houses, industries and public buildings. The same 209 weeks data (2001-2004) were admitted as input data. We projected possible energetic values from the sources for different periods into the last 157 weeks, in random initial instants. New values for maximum power demands are then obtained from the difference between the realistic values and those coming from the renewable sources. These new sets of values have thus been adopted as our set of input data. Since during 2001 there is not yet input from renewables, but only during the 2002-2004 period, and given that these renewable sources enter in the grid progressively, we simulated a new trend for the power demand values seen by the PS. In the sequel, we trained our ANN and adjusted our AR, ARX and ARMAX models according to the first 200 weeks. We have then evaluated the adaptation ability of each method to forecast the weeks demands for the 9 last weeks of 2004.

The existence of power microgeneration is not informed to the grid in ANN, AR, ARX and ARMAX applications; the energy utility has the mission of only detecting the new trend for power demands, and adapting itself to the new curves.

In both chapters 4 and 5, we established that our ANN system has two hidden layers. We have exhaustively searched for the number of neurons which gives the best performance. Therefore, ANN forecasts were obtained with these optimal quantity of neurons per layer.

Chapter 6 contains considerations about the overall experiment and accomplishes an objective conclusion with respect to the best model achieved for each one of the experiments.

There are still two appendices with useful concepts that support the proposals all over this dissertation. Appendix A goes over Stationary Series and discusses concepts about temporal series stationarity, which are requisites for the linear filters used in Chapters 3 and 4. Appendix B contains Z-Transform concepts, comprising auto-regressive (AR) model, moving average (MA) model, ARX (AR with eXogeneous variable) and AR-MAX. Finally, Appendix C gathers several results of spatial prediction of percent peak load reductions for the districts in which the Power Substations lie. Some results are initially shown in the final part of Chapter 5; however, given the huge quantity of possible examples which consider the source parameters variation, a specific Appendix was idealized.

Chapter 3 EMERGENT RENEWABLE SOURCES

In this chapter we examine three sources of electric energy, which are divided into two energy recyclers, which are the Radio Frequency (RF) energy harvesting and the indoor light energy harvesting, and an energetic source itself, which is the electric eel. From now on, despite of the fact that we are dealing with one source and two recycling forms, all the three energy vectors are going to be treated as "renewable sources". For each one of them, the theory and state of the art are overviewed, and data which result from measurements are exposed.

3.1 RF Energy Harvesting

3.1.1 Theory and State of the Art

The main objective of the RF energy harvesting system is to convert the RF power from the space into usable direct current (DC) electrical source [5, 6]. In order to achieve that, a system with several stages is required. The first element is the RF source, which is the TV and Radio broadcast transmitters in our cities. The next necessary element is our RF receiver with its *matching circuit*. While specifying it, one must pay heed to the exploited band, since the manegement circuit is placed at the input stage to equalize the impedance between the antenna and the next component of the circuit. Next, the voltage booster and the rectifier are the items of our highest concern, since our circuit for energy management may constitute a point of power losses. Finally, the energy storage system, which is a battery. Fig 3.1 shows the sequence of elements as here described.

If one antenna serves to furnish power specifically to another antenna, it embodies a dedicated system, which does not apply to the energy harvesting definition [7]. Recycling power occurs when the energetic proposal of the source does not correspond to its original utilization. For instance, we suppose an antenna to broadcast eletromagnetic signal all over the space. In doing so, a certain part of the antenna irradiation can be absorbed by any device, like antennas which aim to scavenge the energy from the wave, while ignoring the signal which is conveyed by the transmitter. But an antenna itself



Figure 3.1: Components of a RF energy harvesting system: the *rectenna* is an antenna with a RF-DC interface [5]. Therefore, the receiver must be integrated to a matching circuit, a voltage booster and the rectifier, whose output is often connected to a battery.

must receive signals from electromagnetic irradiation rather than harvesting power. The adoption of a different objective for the device is the core idea of recycling energy. As we will see in Section 3.2, a lamp which irradiates light which will be harvested by a solar panel is equally considered to make part of a harvesting energy scheme.

According to [7], as ambient RF levels are lower than those that can be provided by a dedicated RF source, the efficiency of the harvesting system and its minimum startup power are of critical importance. RF energy from TV broadcasts is 100 times weaker than solar power [8]. In addition, compared to solar energy, which can only obtain power during daytimes under fine weather, RF energy from TV broadcasts can obtain power 24 hours, except during the maintenance period.

Radio and TV broadcast signals are designed to cover the entire range of human activities, and thus usually they transmit RF intense waves [9]. The amount of received power P_r depends on the distance decay d, the wavelength λ , the gains of transmitter and receiver, respectively, G_t and G_r , and obviously on the transmitted power P_t , as in the Friis transmission equation:

$$P_r = \left(\frac{\lambda}{4\pi d}\right)^2 G_t G_r P_t. \tag{3.1}$$

This equation assumes the free-space propagation. Therefore, in a real environment, the amount of received power will be much less [9]. The banded input RF power

density is calculated by summing all the spectral peaks across the band. These levels provide a snapshot of source availability, and for this reason they are used as a harvester designer starting point (each band will define the input impedance of a rectenna) [7]. Still from (3.1), one can perceive that the higher the frequency, the lower the received power.

Selection of a proper operating frequency band for the proposed RF energy harvesting system is critical since it will affect the overall size of the receiving antenna and operating range of the system [10]. The maximum available power from an antenna P_{tm} is directly proportional to the maximum effective area A_{em} . This relation can be written as [11]

$$P_{tm} = A_{em} W_i, \tag{3.2}$$

where W_i is the power density of the incident wave. An expression that relates the antenna maximum effective area to the different antenna parameters needs to be taken into consideration when designing a receiving antenna, which is given in [12]

$$A_{em} = e_{cd} (1 - |\Gamma|^2) (\frac{\lambda}{4\pi})^2 D_0 |\hat{\rho}_w \, \hat{\rho}_a|^2, \qquad (3.3)$$

where λ is the wavelength of the RF source, e_{cd} is the irradiation efficiency, $(1 - |\Gamma|^2)$ and $|\hat{\rho}_w \hat{\rho}_a|^2$ account for the losses due to impedance and polarization mismatches, respectively, and D_0 is the maximum directivity considering the irradiation pattern of the antenna. In order to maximize the captured power, both the transmitting and receiving antennas should have same polarization [5], which in Eq. (3.3) is related to the $\hat{\rho}$ terms. This equation is very important given that W_i , the power density of the incident wave, is not supposed to change; if we intend to preserve the arriving RF power, each term of (3.3) must be analyzed with attention. Furthermore, the lower the input power, the lower the efficiency of the rectifier circuit [9].

Even though the transmitted power at a broadcast station has the order of more than several kilowatts, the received power will be the order of microwatts to milliwatts. This would explain why the main utilization to RF harvesting energy is providing energy to wireless sensors, which constitute very tiny loads. They work usually isolated and the amount of the necessary energy to put them in operation is considerably low. Generally, a sensor node consumes more than 10 mW of power in order to transmit a packet over a wireless link [9]. In many instances, only a few miliwatts of power are needed to power wireless sensors [13]. Most commonly used wireless sensor nodes consume dozens μW in sleep mode and hundreds μW in active mode [14].

Examples of RF energy harvesting systems efficiency. Despite advancements in end-to-end (i.e., input RF to output DC), only a few power conversion attempts with low input RF power levels at true ambient RF energy harvesting have been reported. For example, one relatively efficient rectenna, utilizing a modified omnidirectional patch antenna, has an efficiency of 18 % with a single-tone input RF power of 20 dBm [15]. This helps us to comprehend how limited harvesting RF waves energy can be. In order to enrich our analysis, the maximum dBm power which is transmitted from the antennas is shown in Table 3.1.

Table 3.1: Frequency bands of energy harvesting and respective maximum dBm power

Standard	Frequency Bands	Band of Interest	Max. Power (dBm)
DTV	470 - 862MHz	470 - 862MHz	70
GSM	900, 1800 MHz	925-960 / 1805.2-1879.8 MHz	40
UMTS	$2100 \mathrm{~MHz}$	2110-2170 MHz	40
Wi-Fi, Bluetooth	$2.4~\mathrm{GHz}$	$2.4~\mathrm{GHz}$	30, 20
New Wi-Fi	$5~\mathrm{GHz}$	$5~\mathrm{GHz}$	NA

A 7 days measurement of the characteristics of TV broadcast RF energy harvesting was carried out in [9]. The measurement was performed in the balcony of a laboratory located 6.6 km away from the TV source; this tower broadcasts TV signals over the UHF band. The amount of power harvested over 7 days had mean at 20 μ W. A 1 k Ω resistor was used as a load resistor, being the one which can extract the maximum power from the energy harvester.

In terms of GSM signals, we may expect, indoors everywhere or outdoors on an elevated level, a power density between 0.01 and 1.0 mW/m² ($10^{-3} \sim 10^{-1} \mu$ W/cm²), taking into account distances between 25 m and 100 m from a GSM900 base station [16]. If we consider the power integrated over the downlink frequency band (935 ~ 960 MHz), i. e., the summed power density, we may expect a total power density between 0.1 and 3.0 mW/m². The power density received from GSM1800 base stations are, up to 100 m, in the same order of magnitude as those received from GSM900 base stations at a single frequency or summed for low traffic situations [17]. In [18], a work was carried out at 900 MHz with 50 Ω impedance, using a resonance circuit transformation coupled with a Schottky diode. This scheme yields a DC output voltage of over 0.3 V for an input power level of -26 dBm (2.5 μ W). A DC voltage of 0.8 V was achieved from RF input power level of -20 dBm (10 μ W) at 868.3 MHz through simulation with no load [19]. A Cockcroft-Walton multiplier circuit was used and produced 1.0 V DC voltage onto a 200 M Ω load for an input power level of 1.0 μ W at a fixed frequency of 2.4 GHz [20].

According to [21], experimental results involving the operation frequency of 945 MHz, conversion efficiencies of -5 dB, 0 dB and +5 dB corresponded, respectively, to 3 %, 5 % and 7 %, with R = 100 k Ω , and 20 %, 23 % and 25 % for R = 50 k Ω . This analysis presupposes the employment of a two-stage Dickson voltage multiplier. The efficiency at 0 dBm was 21 %, whereas around -5 dBm the efficiency was 3.2 % [16].

A set of simulations and one practical implementation were realized in [6] with fixed RF at 945 MHz \pm 100 MHz, which is close to the down link center (947.5 MHz) radio frequency range of GSM900 band. The voltage obtained at the final node (VDC4) of the multiplier circuit was recorded for various input power levels from -40 dBm to +5 dBm, with power level interval (spacing) of 5 dBm. Different load resistors were simulated (50 k Ω , 100 k Ω , 200 k Ω and 1 M Ω), but only one of them was tested (100 k Ω). Figure 3.2 shows the set of results.



Figure 3.2: Resulting V_{out} according to input dBm [6].

There is still an example of incident power in high buildings from [7]. Measurements within the Department of Electrical and Electronic Engineering building at Imperial College of London were taken on the 11th floor of the south stairwell. These are shown in Fig. 3.3. One must note that it is related to the end-to-end efficiency, which is the quocient between the time-averaged output (i.e., equivalent DC) power into the storage element and the time-averaged input RF power; in other terms, it depends strongly on the efficiency of the harvesting system as shown in Fig. 3.1, but also are relevant as examples for the end-to-end efficiency.



Figure 3.3: End-to-end efficiencies for ambient RF energy harvesting [7].

3.1.2 Measurements

In May 11th 2013, several measurements were undertaken with the aim of evaluating the use of RF waves as an ubiquitous energy source. Values of incident power, in dBm, were measured from four points of Brasilia-DF (Brazil) with the aid of a spectrum analyzer. The measurements were all carried out approximately 1.7 meter above the soil.

The four surveyed points stay near two main targets of research: the Digital Tower $(15^{\circ}41'57.67"S/47^{\circ}49'46.47"O)^{1}$ and the TV Tower $(15^{\circ}47'25.49"S / 47^{\circ}53'29.45"O)$. For each target, two sites were chosen to verify the incident power in dBm values. The sites selected to take measurements from the Digital Tower were Place 1 $(15^{\circ}41'29.70"S/10^{\circ})$

 $^{^1\}mathrm{All}$ places here described had their coordinates drawn from Google Earth.

47°51'13.96"O), on the border of the road leading to the Digital Tower, and Place 2 ($15^{\circ}41'59.46''S/47^{\circ}49'50.38''O$), right in front of this target. In order to take measurements over RF power irradiance arriving on points around the TV Tower, places 3 ($15^{\circ}47'25.13''S/47^{\circ}53'28.88''O$) and 4 ($15^{\circ}47'39.61''S/47^{\circ}53'11.72''O$) were chosen. These places are depicted in Fig. 3.4.



Figure 3.4: Map of Brasilia-DF (Brazil) with the indicated 4 places on which RF spectrum intensities were measured.

Each one of the four places was visited three time during that day, in order to register morning, afternoon and evening dBm values. The dBm power level took into account the corresponding frequency, and therefore several measurements over different frequencies were made so that we could feature the dBm behavior according to the part of RF spectrum. The graphs made for each place, as shown in Fig. 3.5, comprised values over an entire day.



Figure 3.5: Measurements of incident dBm power according to frequency. Each place encloses morning, afternoon and evening data, which are mixed in each graph.

At a first glance, Fig. 3.5 allows us to conclude that the TV Tower, nearer to places 3 and 4, conveys the highest measured amount of dBm, since only in these places positive dBm values were found. TV Tower constitutes the point from which flows the greatest amount of watts, given that place 2, in the parking are of Digital Tower, and place 3, in the parking area of the TV Tower, are situated in very similar distances to their respective target. One aspect to be observed is that the Digital Tower might be out of service, but it was not possible to confirm such information.

Significant low values are obtained from sources which are not transmitter antennas: according to [22], the Earth receives 1 nW/m^2 from a magnitude +3.5 star, which corresponds to -60 dBm. These values are not useful, either for the slight power, or given that the source is unknown. Therefore, a better analysis arises from the set of the dBm values which overcome -15 dBm: the set of these values are depicted in Fig. 3.6. The measurements of the four places are here gathered and their points are described by different colors. One can notice the predominance of Place 3 with several points,



whereas Place 2 has just one point in this set of data.

Figure 3.6: Results which overcome -15 dBm, here called "best results", with values properly signalized in order to show participation of each place in this group.

The most concentrated part of the spectrum graph stood around 90 MHz. In Fig. 3.6, all measurement points with regards to places 1 and 4 caught signals around 90 MHz, as well as the only value of Place 2 which shows up here. Place 3 bears points here and in the other parts of the graph. The main conclusion is, since we intend to avail ourselves of the incident RF power, the range of frequency 90-100 MHz is the most indicated interval to achieve an efficient energy harvesting scheme. All observed places have relatively good dBm values into 90-100 MHz, and the best dBm values are also in this range. Therefore, the higher probability of success involves this range.

There are 24 points appearing in this graph, three being in overlap. From the 24 points, 9 were taken in the morning, 8 were taken in the afternoon, and 7 were recorded in the evening. Accounting on this observation, we assume that the period of the day is not relevant for obtaining higher or lower values of dBm. The distance to the target and the target itself are considered much more important.

3.1.3 Expected results given the measurements

We now call upon the known expression shown in (3.4):

$$dBm = 10 \log_{10} \left(\frac{P}{1 \text{ mW}}\right) \tag{3.4}$$

This expression provides a rapid way to calculate the incident power as a function of the measured dBm. Table 3.2 contains a brief set of examples of values which come from the simple application of (3.4).

dBm	-15	-10	-6	-3	0	1	2	3	5
Power (mW)	0.032	0.100	0.251	0.501	1.000	1.259	1.585	1.995	3.162

Table 3.2: Incident power as a function of dBm values - examples

Crossing the average efficiency of 17 % [15] with an incident power of -3 dBm, which is near to the mean over the graph shown in Fig. 3.6, we achieve the average power of 86 μ W. During 24 hours, the total energy harvested achieves 2.058 mWh. This amount of daily power will be multiplied for the number of antennas, resulting on the daily available energy for the complete group of RF harvester antennas.

3.2 Indoor Light Energy Recycling

3.2.1 Theory and State of the Art

There are many sources of energy that are not, or barely, used for power generation. According to [8] and [23], technical aspects enable luminous energy to be one of the most promising power sources, given that their power density related to area or volume is among the highest ones. In a outdoor environment, at noon, the ratio of available energy for a solar cell reaches 15 mW/cm^2 .

On the other hand, a small piezoelectric generator using shoes movements achieves $330 \ \mu W/cm^3$; a thermoelectric generator (10 °C gradient) produces about 40 $\mu W/cm^3$; acoustic noise generator (100 dB) falls short of 1 $\mu W/cm^3$ [23].

We evaluate if luminous energy from artificial light constitutes a potentially exploitable source in terms of harvesting systems. According to [24], the harvesting involving indoor environment provides 3 orders of magnitude lower than outdoor environment. In recent publications [24], [25], [26], [27], [28], [29], [30], [31], [32] there has been an increasing interest in energy harvesting in indoor environments. However, in [25], [26], [27], [28], [32], [33], [34], [35], the goal was supplying a sensor node. Due to the importance of charging portable electronic devices, in this section, we perform measurements in order to verify the possibility of supplying the battery of a cell phone using artificial light. As an additional contribution, we also propose a circuit to charge a cell phone using artificial light and we show the efficiency of our proposed circuit by means of simulations.

In [24], photovoltaic devices are classically optimized for the solar spectrum. Energyefficient fluorescent and LED lighting have largely replaced incandescent light bulbs, so that the spectral profile of artificial light has changed from broad spectra, originated by low temperature blackbodies, to sharply peaked narrow spectra. Table 3.3 presents a comparison between crystalline Silicon (c-Si), amorphous Silicon (a-Si) and P3HT/PCBM bulk heterojunctions organic PV (OPV) panels in experiments with light from solar light, Compact Fluorescent Lamp (CFL), Cold-Cathode Fluorescent Lamp (CCFL) and LED [24]. Based on Table I, an a-Si solar panel is more suitable for indoor artificial light in comparison with other types of panels.

Table 3.3: Maximal Efficiency Values for Crystalline Silicon, Amorphous Silicon, and Organic BHJ Solar Cells under different spectral illumination [24].

			Luminous	Source	
		Solar	CFL	CCFL	LED
	$\operatorname{crystalline-Si}$	49~%	50~%	52 %	54~%
PV Material	amorphous-Si	37~%	74~%	70~%	80~%
	OPV	28~%	63~%	59~%	63~%

In [27], an amorphous silicon photovoltaic cell (PV) has a relatively high efficiency at low light intensity levels compared to other types of cells. Such property makes them particularly suitable for indoor and low power light environments. Other types of panel, as c-Si, have higher performance in an outdoor environment compared to the indoor environments due to the scarce spectrum of the artificial light as exemplified in Fig. 3.7 [24].

3.2.2 Open Circuit and Short Circuit Measurements in Indoor Environments for the a-Si Panel

Measurements were performed with the a-Si solar panel with no load connected to it. Our goal here is to observe only the behavior of the a-Si solar panel using indoor artificial light.



Figure 3.7: Typical solar light AM 1.5 and cold-cathode fluorescent light spectra.

Given that incandescent light bulbs are falling into disuse in many countries, in this Section only fluorescent and LED lamps are considered. This factor is determinant for choosing a-Si panel, since it offers a higher efficiency to produce power from such spectrum as shown in the previous section. Another reason to choose a-Si panel lies on its low cost. From now on, all presented results were obtained by means of measurements obtained in our laboratory.

A panel of amorphous Silicon (a-Si) with maximum current $I_{\text{max}} = 200$ mA and maximum voltage $V_{\text{max}} = 6$ V and with size of 95 mm x 110 mm is employed in the experiments. Over 54 measurements were performed with the a-Si panel being under different light intensity levels. The variables of interest are open circuit voltage, V_{oc} , and short circuit current, I_{sc} , in function of incident lux L_{x} . A white 8 W LED light bulb (color temperature 4.700 K) was used for the measurements. Figures 3.8 and 3.9 show the curves obtained during the measurements. Intensity and spectrum of light were kept constant, whereas distance between panel and LED was varied, in order to provide several light intensity levels.

From the curves of Figs. 3.8 and 3.9, Eqs. (3.5) and (3.6) can be written with the relationship between $V_{\rm oc}$ and $L_{\rm x}$ and between $I_{\rm sc}$ and $L_{\rm x}$:

$$V_{oc} = -0.68789 + 0.68159 \cdot ln(L_x), (3.5)$$
$$I_{sc} = 0.00074 + 0.00084L_x, (3.6)$$

where $\ln(\cdot)$ is the natural logarithm operator.



Figure 3.8: Absorbed Light Intensity L_x [lux] versus Open Circuit Voltage V_{oc} [V] of the solar panel 95 mm x 110 mm for a LED 8 W lamp.



Figure 3.9: Absorbed Light Intensity L_x [lux] versus Short Circuit Current I_{sc} [mA] of a solar panel 95 mm x 110 mm for a LED 8 W lamp.

From Eqs. (3.5) and (3.6) we obtain values of the coefficient of determination (R^2) , respectively, equal to 99.52 % and 99.65 %. Therefore, both equations have a very good approximation with the curves in Figs. 3.8 and 3.9. Significance of the independent variables are 0.01 % in either cases. Significance and coefficient of determination were obtained by means of linear regression, which comprises the principle of residuals normality. All over this section, the obtained data samples bore this principle, which is also understood as the variance constancy, or homocedasticity.

Inverting terms in (3.5) results in $(L_x) = 1.0092 \cdot e^{1.46716 \cdot V_{oc}}$. Since solar cells are voltage limited current sources, the relation between the light intensity L_x versus short circuit current (I_{sc}) is linear as shown in Figure 3.9.

3.2.3 Charge Profile of a Cell Phone Battery

Measurements were performed connecting the cell phone directly to the power outlet through its charger. Our focus here is to identify the behavior of the battery. Experiments employed a Samsung B3210 cell phone, that uses a 800 mAh Lithium-Ion rechargeable battery. The battery was initially completely discharged in order to recharge it using an electrical socket from the electrical power system, whose output is 220 V in 60 Hz.

The whole process comprises a four-charging regime [36]: the first stage corresponds to a constant current charge while voltage increases; the second stage is related to the saturation, when voltage stabilizes whereas the current lowers; the third phase is cut-off, voltage decays softly. Finally, the stand-by mode occurs when the battery draws a low-level current from the source in order to recuperate voltage after some charge leakage. At the beginning of the charging process, the current is maximum and decreases until a time instant in which the current falls abruptely, when the second stage starts. These two first stages are pictured in Figure 3.10.



Figure 3.10: Current and voltage versus time of a cell phone charging at 220 V, 60 Hz.

At first, current underwent a linear decay, until the 73rd minute. Thereafter, it obeyed to another linear expression, corresponding to the less steeper descendent along lower values (saturation). The expression of current I as a function of time t in minutes is given by expressions (3.7) and (3.8). Time instants were taken at each minute. At each time instant, the current and the voltage were measured, making possible to calculate the instantaneous transfer power. Cell phone was kept turned off during the entire process.

For
$$0 \le t \le 73$$
, $I = -0.859t + 439.71$, (3.7)

For
$$74 \le t \le 203$$
, $I = -0.194t + 111.26$. (3.8)

The approximation error in terms of determination coefficients R^2 for expressions (3.7) and (3.8) are, respectively, 95.59 % and 99.07 %.

The internal control system of the cell phone manages the input voltage from the DC source, which delivers 5 V all the time. The first stage comprises an increasing voltage regime, as the current is managed to stay in a narrow set of values by the control system. When the voltage achieves approximately 4.15 V, this system steps in to reduce the current towards lower levels, avoiding damages. This moment corresponds to the observed discontinuity over voltage and current values, when the second phase of charging the battery begins. This is in accordance with Fig. 3.10.

The cell phone internal circuit control aims to preserve the battery physical safety, since the chemical structure of the Li-ion battery does not allow overcharging [36]. Considering the profiles of current and voltage as observed above, it is possible to imagine that there is an equivalent R which is not constant and might replace the entire cell phone, in order to produce the same final circuit current. Hence, the cell phone could be substituted by a black box, into which there is a time-variable resistance whose value increases as the stored energy arises, due to the behavior of the cell internal control system. The higher the internal voltage source of the battery, the higher is the apparent equivalent resistance, which has the behavior shown in Fig. 3.11. This is based on a mathematical rather than a physical comparison, which nevertheless bears concrete results into the modeled circuit. The instantaneous values of this variable impedance were obtained by the division of voltage upon current. The expression for R in function of time, in the right part of graph (from the 75th minute on), is now $R = 44.713e^{0.002t}$.



Figure 3.11: Apparent Impedance (Ω) of cell phone seen by the source.

In Fig. 3.12, the instantaneous values of delivered power were obtained by multiplying the current by the voltage of 5 V, which is delivered from the outlet. Therefore, Fig. 3.12 depicts the power transferred to the cell phone, instead of its battery. This was made to visualize in which conditions such transferring power would be more effective, since we consider that the external voltage source, either the outlet or the boost converter associated with the a-Si PV panel, is supposed to deliver 5 V all the time. Fig. 3.12 illustrates how the power evolves instantaneously from the panel to the load.



Figure 3.12: Instantaneous transferred power to the cell phone as seen by the source.

One may notice that the most favourable situation to transfer energy to the cell phone occurs during the final part of the saturation stage. Since the apparent impedance arises, the cell phone will draw a lower current, diminishing the probability of voltage to plunge, considering that the source is a solar panel. Hence, hereafter we adopt the premise according which the higher is the energy stored in the battery (higher equivalent impedance), the more effective will be the power transfer. This corresponds to the right part of the graphic shown in Fig. 3.11. Furthermore, the PV panel has a higher internal impedance, and such characteristic offers the possibility of increasing the impedance load and the power simultaneously.

3.2.4 Indoor Experiment Using Artificial Light and Results

In this experiment, the cell phone is charged considering indoor light. In order to perform measurements in indoor scenarios, the environment in Figure 3.13 is suggested. This figure shows an office-desk with a lamp in its support. The a-Si panel was installed on the bottom of the lamp support, inside of which an 8 W LED lamp was installed to provide artificial light. The center of the LED lamp was considered to be 13 cm apart from one of the panel borders. The cell phone is connected to the solar panel and a boost converter, using it as its charger, in the same way while charging it directly by the grid.



Figure 3.13: Overall deployment of cell phone, amperimeter (in series) and solar panel under a light-support in an office for electrical current measurement.

An enlarged solar panel, which delivers 2.8 V (V_i) and power of 3.92 W at lower luminosities, was considered. These hypothetical values for the panel aim to allow the examination of possible results in the recycling process. Moreover, the panel concerning to this stage of the experiment is made of two modules of 24 cm x 35 cm each (therefore, 0.168 m² of total area). Considering the environment of Figure 3.13, light beam must be directed to the plan of the desk and not to the panel, since the charging process should not disturb the usage of the desk for reading and other activities.

A boost converter was designed and simulated in order to step up the solar panel voltage allowing the battery charge. Fig. 3.14 shows the circuit designed to charge capacitor. According to [37], the commercial available boost converters use recommended capacitor and inductors values, stated in the datasheets. This simple boost converter was designed with an oscillator of 200 kHz whose duty cycle is 80 % (D = 0.8), since $V_{\rm out}$ is 5 V, therefore five times the value of $V_{\rm i}$. The cell phone will only behave as a



Figure 3.14: Designed boost converter producing 5 V in steady state

valid load when the boost output voltage is around 4.7 V - 5 V; below those voltage levels, it is likely to behave as an open circuit. Given the constraint of 2.8 V output voltage, we estimate more accurately the value of L in order to achieve the amount of power that the panel should furnish, which is 3.92 W, as well as to determine the oscillator frequency. The critical inductance is defined as the inductance at the boundary edge between continuous and discontinuous modes and is defined as (3.9).

$$L > \frac{V_{\text{inmin}} \cdot (V_{\text{out}} - V_{\text{min}})}{F_{\text{sw}} \cdot I_{\text{out}} \cdot V_{\text{out}}^2} [H]$$
(3.9)

In (3.9), V_{inmin} is the minimum input voltage, V_{out} is the desired output voltage, I_{out} corresponds to the desired maximum output current; F_{s} is the switching frequency of the converter, and K_{ind} is the estimated coefficient that represents the amount of inductor ripple current relative to the maximum output current. A good estimation for

the inductor ripple current is 20 % to 40 % of the output current, that is, $0.2 < K_{\text{ind}} < 0.4$. Considering feeding cell phone with power, $V_{\text{out}} = 5$ V. It was adopted $K_{\text{ind}} = 0.3$, $F_{\text{sw}} = 200$ kHz and $V_{\text{inmin}} = 1.0$ V, with $I_{\text{out}} = 100$ mA. The resultant minimum value is 26 μ H. If our choice for V_{inmin} is 1.16 V, then the minimum L is 32 μ H. These observations allow us to choose an inductor of 33 μ H.

According to [38], since the capacitor's equivalent series resistance (ESR) affects efficiency, low-ESR capacitors will be used for best performance. A load resistance of 50 Ω was adopted, according to the verified values for R in the beginning of the second stage of Fig. 3.11. An approximated expression for the required capacitance as a function of ripple voltage requirement, ΔV_{out} , D, switching frequency F_s and output voltage, V_{out} is given by

$$C \ge \frac{V_{\text{out}} \cdot D}{F_{\text{s}} \cdot \Delta V_{\text{out}} \cdot R}.$$
(3.10)

Considering ΔV_{out} , which is the maximum ripple output voltage, as 5 % (or 0.05), assuring the minimum output voltage to stay over 4.75 V, the minimum value of output capacitor is 8 μ F. Since the constant RC, shown in (3.11), requires commonly a higher value for resultant C in order to avoid the deterioration of voltage value, it is possible to choose C2 to be 20 μ F.

However, with these values of R and C2, even when C1 equals 100 μ F the result makes I_{out} to decay less than 0.5 %. Considering that low levels of resistance load might demand higher capacitances in order to avoid I_{out} to lower significantly, the value of C1 is chosen to be 100 μ F. All the components are part of the circuit shown in Fig. 3.14.

$$i_c = \frac{E}{R} e^{-\frac{t}{RC}} \tag{3.11}$$

Fig. 3.15 depicts the blue line which represents the boost output voltage, whereas the black line shows the effective values of voltage and current absorbed by the cell phone.

Considering the conditions of the boost and the panel, which must provide 2.8 V minimum, the time to achieve steady state does not overcome 0.3 ms.

In a limited voltage current source, as the solar panel, in the flat part of the $V \ge I$ graphic, the bulkier the resistance coupled, the higher is the power transferred to



Figure 3.15: Effective voltage and current in steady state over a load of resistive load R of 50 Ω .

load. If power is RI^2 , then it is interesting to use load with higher values of resistance, provided that it does not make voltage to drop excessively. According to Figure 3.11, R increases when full charge of the cell phone has just been completed. Figure 3.16 is in accordance with Figure 3.11 showing that during the period "A", the battery will receive less energy than during period "B", considering equal durations.



Figure 3.16: Bars indicating equal periods of time A and B as demonstration that transferred power is influenced by resistance (PB>PA).

According to our simulations, about 0.5 W can be steadily transmitted to a battery, day and night, since illumination over day hours is fairly more intense than by night.

It can be noticed that in just one day (24 hours) the proposed system can return 12 Wh to a battery, which overcome the overall energy needed to its full charge, which is approximately 3.72 mWh. Much more than this can be achieved by just making smaller the distance between lamp and the a-Si panel: being the panel 1.5 cm closer to the LED lamp, current increased 36 % and, therefore, 85 % in power.

Our simulations will consider two cases: solar panel will generate power only from inside houses and buildings, and therefore we can consider to produce 12 Wh per day and per square meter. Despite our panel has 0.168 m^2 , the enhancement of this area will not likely result in additional energy to be drawn; 168 cm^2 probably occupies a single lamp completely. For the sake of conservatism, we will adopt 1 m² of a-Si panel to generate the amount of 0.5 W continuously. A second simulation is also projected: in this situation, the panel will harvest power from indoor light only at night, generating energy outdoor during the day hours. In order to provide the respective energetic measurement, the next section is developed.

3.2.5 Expected results for outdoor environments

According to [39], monocrystaline, policrystaline and amorphous silicon solar cells are capable to produce, in the Maximum Power Point (MPP), an electric current of about 32 mA/cm^2 , being the voltage in the range [0.46 V 0.48 V]. Cells arranged in parallel sum the current, whereas cells arranged in series sum voltage. Taking into account 1 m^2 we can put in position 64 modules of 156cm^2 in series, for example, obtaining 5.0 A over [29.44 V 30.72 V], resulting in [147.2 W 153.6 W]. This values are in accordance with the commercial specifications of module 80 Wp, which has dimensions $0.6 \text{ m} \times 1.0 \text{ m}$.

Standard Test Conditions (STC) include irradiance of 1.000 W/m^2 , Air Mass (AM) 1.5 and temperature of 25°C. Values of MPP are calculated by the manufacturer under STC. These considerations allow us to adopt the overall average efficiency of 15 %, i. e., for each incident 100 W onto the PV panel, approximately 15 W will be effectively converted into useful power.

Photovoltaic generators may provide power which vary in function of the external conditions they undergo, the intensity of incident light and the ambient temperature. These data are supposed to oscillate substantially in a single day and from one day to another. Information about broader periods can help to achieve average results. Since the application of this study is thought to be in a Brazilian city, we resort to [40] in order to obtain the distribution of global daily irradiations in annual averages, which is shown in Table 3.4.

Brazilian Geographic Region	Annual Average of Global Irradiance $[\rm Wh/m^2]$
North	5462
North-East	5688
Middle-West	5630
South-East	5478
South	5015

Table 3.4: Annual Average of Daily Irradiance [Wh/m²] by Brazilian Region [40].

This set of values justifies the adoption of 5000 Wh/m^2 as the total incident power originated from solar beam during a day, still being a conservative value. Crossing this parameter with the average efficiency of 15 %, we achieve the daily available power of 750 Wh/m^2 . The corresponding weekly amount of energy is 5.25 kWh/m². This value will be taken on as the main reference related to projections for PV generation within Brazilian cities.

3.3 Electric Eel

3.3.1 Theory

The third analyzed element, which is an energy source itself rather than a recycling scheme, is the electric eel. This freshwater fish is rife in Amazonia and converts chemical energy existing in their food into voltage along their body constantly, emitting higher values when exposed to stronger stimuli. Values of voltage along the body of this animal are creditable of investigation: a single discharge can reach several hundreds of volts. The utilization of electric eels for energetic objectives is, nonetheless, still inexpressive. We discuss why and feature eels' properties in order to conceive a worthy scenario for its employment as a renewable source.

3.3.2 State of the Art

3.3.2.1 Electric eel properties

Electric eels can live in freshwater. Desirable values for pH are within 5.0 - 7.0, and water electric condutivity may vary between 30 - 100 μ S, being about 25° C the expected temperature of water for the fishes lying in Amazonia [41]. Electric eels have no lungs, but rather a respiratory organ inside the mouth. Hence, they are mandatory air-breathers. The interval between two breathes is about five minutes. Their body can reach 2.5m and they can weigh up to 20 kg. If they are kept in an aquarium, about 30 % of the water must be renewed in order to preserve their healthy. For electric eels, the ideal aquarium may have 1.5m of water depth [42].

The electric eels has the scientific denomination *Electrophorus electricus* and bears three electric inner organs which only act simultaneously under an attack or when the eels aims to kill a prey, which are the cases of the strong discharge. The amplitude of this discharge current can reach 1 ampère, for a large eel. The respective voltage can achieve the intensity of 600 volts. Therefore, during this time, the electric eels can emit a power of approximately 0.6 kW. However, under normal conditions, the eel emits few volts only for seeing what exists in the environment during navigation, and also for communicating. The chemical principle behind the voltage generation is shown in Fig. 3.17. Inside the eel (*Electrophorus*) there is a modified muscle, the electric organ, whose fundamental cellular unit is the electrocyte. These cells are very similar to muscle fiber cells in which they can produce electric signals, but, unlike muscle fiber cells, they cannot contract. Such cells contain a high concentration of Potassium ions (K^+) , a low concentration of Sodium ions (Na^+) , and a comparable amount of negatively charged ions. The cell membrane is permeable to K^+ ions but not to Na⁺ ions. When in this inactive state, the cell membrane has a negative voltage, avoiding potassium ions to flow out the cell [43].

Acetylcholine, a neurotransmitter, can activate the electrocyte cells. It is secreted through nerves on one side of the cell, causing ion channels to open on that side. Na⁺ ions are then able to rapidly enter the cell via these channels. This influx of positive charge modifies the equilibrium potential of the cell. To reestablish this equilibrium, the K⁺ ions leave the cell on the other side, still very permeable to K⁺.

About 6000 of these electrocytes cover approximately 4/5 of the eel's body length, and when these cells are activated, the eel produces the voltage amplitude. The head of



Figure 3.17: Electrocytes behavior which yields an external electric voltage: (a) cell membrane is permeable only to K^+ ions; (b) Acetylcholine activates the electrocyte cell, making Na⁺ ions to come inside it [43]. In the case of *Electrophorus electricus*, there exists only the posterior membrane, which is excitable.

the eel has a positive charge, and the negative charge stands on the tail. The emitting voltage are also responsible for giving to the eels a substitutive sense of vision, as the sonar, for a bat, since eels do not see sufficiently with the eyes. Objects having different conductivities distort the electric field that the eel produces, thus making the eel aware of such objects' presence. Furtheromore, voltage discharges perform an important role in social communication. Even the gender identification among electric eels depends on the characteristics of discharges profile from each one of them [44]. Fig. 3.18 illustrates the path current lines that the eels emit.

Electric eels use low values of electric current to perceive the world around them, and they make it 24 hours daily. With regards to Fig. 3.18, any scheme of harvesting energy from these fishes should provide conductors in proximity to the head and the tail, in order to shorten the path of energy through water.

3.3.2.2 Available Papers

Consulting the available papers in IEEE about eels, only one paper was found [45] which studies these fishes with regards to their capability of producing electricity. This article provides some interestig data, as for instance the energy produced per unity of tissue volume. Electric energy produced per cubic centimeter of tissue is 370 microjoules per impulse in a 60-centimeter eel and only 100 microjoules per impulse in a 180-centimeter


Figure 3.18: Lines of electric field generated along the eel's body. The head is the positive pole, and the tail bears the negative pole, from where electric current goes off through water until the head [44].

eel. Therefore, the capability of producing energy per volume of tissue decreases with higher volumes. Such impulses, however, are not the source of power to be considered in this study, since we will handle the continuously emitted electrical power, whose magnitude are in the order of few miliwatts [45].

One of the greater merits of this article is the establishment of an equivalent circuit for the eel, which is shown in Fig. 3.19. In this model, E is the effective electromotive force of the 10-centimeter segment of electric tissue, and r is the effective internal resistance. R' is the effective resistance of all closed circuits within the nonelectric tissue of the fish, and R is the external load. The tissue is a complex network of electromotive force and resistance. The substitution of the simple model is justified by Thevenin's theorem. The application implicitly assumes that E, r, R' and R are independent of the current.



Figure 3.19: Equivalent circuit of the electric eel [45].

On the best knowledge of the author, there are not experiments aiming to harvest energy from eels with a clear market insertion model, as proposed in this study. A possible explication is that these fishes are not indicated for household captivity. Moreover, their expected amount of energy, in steady state, seems to be not worth it in face of the necessary investments.

3.3.3 Measurements

With the goal of harvesting energy from the eel and measuring the electric power, an experiment was carried out in Laboratório de Fisiologia Comportamental e Evolução do Instituto Nacional de Pesquisas da Amazônia (LFCE-INPA), in the city of Manaus-AM (Brazil). An electric eel measuring about 90 cm was inserted into an aquarium of approximated sizes 120 cm x 50 cm x 50 cm. Two plaque-shaped electrodes were installed on each side of the larger width of the aquarium. The correspondent resistance of each electrode is 4 Ω . Figures 3.20, 3.21 and 3.22 show, respectively, the photo from the real experiment, a picture which describes the whole deployment of the experiment, and the respective equivalent circuit.



Figure 3.20: Real experiment carried out in LFCE-INPA, wherein the electrodes are metal plaques with a resistance of 4 Ω each. This photo was made while the electric eel swam in order to look for the prey, which is the little fish in the above part of the aquarium, near to the surface.

The resistances which appear in Fig. 3.21, 3 Ω each, were included in order to improve the voltage measuring accuracy, and as an artifice to protect voltimeter against high voltage values, in the case of a possible low impedance on the rest of the circuit. An oscilloscope was connected to the voltimeter terminals.



Figure 3.21: Picture which outlines the overall deployment of the experiment: two metal plaques actuating as electrodes were placed on each side of the largest width, which is 120 cm. Two resistances were then purposefully inserted in the part of circuit the voltimeter was connected.



Figure 3.22: Equivalent circuit of Fig. 3.21, illustrating the eel as the voltage source. R_a is the varying water resistance, and 8 Ω corresponds to the total electrodes resistances in series. The resistances of 3 Ω were inserted to protect the voltimeter and to improve measuring accuracy.

The entire experiment enclosed three steps: (i) the eel was surveyed while it was calm; (ii) brief stimuli of stress were applied to the eel; (iii) a little living prey was offered to the electric eel. The duration of the steps (i) and (iii) was 120 seconds each, whereas the situation (ii) lasted 60 seconds.

The values of voltage appearing in the graphs of the following subsections were obtained by considering that the condutivity of freshwater is between 10 μ S/cm and 70 μ S/cm, as stated in [46]. Electrical condutivity between 20 μ S/cm and 70 μ S/cm are expected. The average value of **40** μ S/cm was therefore combined with the mean distance of the eel to the electrodes in order to characterize conveniently the resistance R_a .

3.3.3.1 Electric eel under still water

In the first situation, the electric eel was let in the aquarium with no external intentional stimulum. This step aimed to feature the voltage along the eel's body while it is only recognizing the environment or trying to communicate. The eel was able to move easily trough the aquarium. When, in a given moment, the eel turned around and swam in the opposite direction, the recorded voltage values changed from negative values to positive ones.



Figure 3.23: Voltage values with the eel swimming freely into the aquarium.

Voltages emitted by the eel appear in Fig. 3.23. The inversion of voltage signal is due to the eel to turn around into the aquarium, while it swam. Electric equivalent DC current in this situation were around 1 mA. The mean value of voltage in this stage of

the experiment stood around 10 V. An eel whose size is 90 cm can, therefore, in steady state, furnish 10 mW for the power managing circuit. The advantage of the eel over the other renewable sources is exactly the level of voltage by which power is provided, since the other sources provide power with few milivolts.

The amount of power delivered in this situation can be enhanced by means of employing a larger eel, since the amount of deliverable power increases with the weight of the eel, although not linearly, as we beheld in [45]. Hence, it is valid to affirm that 10 mW in steady state are available if no expressive losses in the aquatic path current occur.

3.3.3.2 Electric eel under stress

In this step, the eel underwent some stress. The stimuli consisted in agitating water randomly during 60 seconds with a bar.

It is obvious that such type of stimulum is not sustainable in the long-term, since it draws from the electric eel an amount of power that this fish cannot provide in steady-state. As shown in Fig. 3.17, there is a physical limit for the Acetylcholine to activate the electrocyte cell, as the amount of ions is not infinite.

Furthermore, it is not imaginable to stress the electric eel permanently, which would be inhuman and, besides, inefficient, given that the eel could learn the external inputs and stopping of answering as we desire it to do. The main point is that our managing circuit must be prepared to avail the energy coming from the fish even in the stress situation.

3.3.3.3 Electric eel receiving a prey

During this step of the experiment, a little fish was offered as a prey to the electric eel. In the very beginning of this process, the eel emitted a large voltage pulse. In the sequence, the electric eel decreased voltages intensity until the end of recording values.

Nevertheless the eel provided high values while it desires to kill the next meal, it depends also on how hungry it is. Forecasting confidently the amount of energy in this situation would be almost impossible. Fig. 3.24 shows voltages values during the second and the third described situations: (a) electric eel under stress, and (b) eel receiving a prey. One can notice that results for voltage values in (a) are more spaced each other; on the other hand, they are more intense.



Figure 3.24: Voltage values with (a) the eel being stressed, and (b) eel trying to kill a little prey that is placed in the aquarium.

We consider hereafter only the values extracted from the step 1 observed in the previous experiment. We suppose that, since electric eels can provide such values of voltage on steady-state, they cannot furnish high values of voltage all the time. It is assumed that the overall energy available we can reliably count on in the end of process is higher as the variance of voltage values is lower.

3.3.3.4 Structure for electric eels energy harvesting

The premise according which at least 30 % of the water must be renovated weekly rises questions of the energetic cost in doing so. The ideal situation is not expending energy with this activity. Captivity systems are likely to demand this water to be changed and, therefore, imposing an undesirable energetic duty.

This study considers to keep electric eels in the river, generating energy from there. As shown in experiment under still water, eels emit a "stand-by" voltage 24 hours daily, indicating therefore the amount of electrical power we can steady count on. Despite in this case the electric eels produce voltage values that are considerably lower, such energy is considered reliable.

One solution is to build a structure within which the eels could swim with no hindrances, but not being allowed to execute large displacements. The system illustrated in Fig. 3.25 comes up as a possible solution. It consists on a basket with no obstacles to the water to flow inside it, working as a swimming-pool since it constrains the electric eels to go out. So, water can come in and out of this structure, unlike the eels. It is supposed to work as a type of corral for the electric eels inside it. The surface of water should stand approximatley on the half height of it, enabling the eels to come upwards to breathe. Two conductive structures must be installed into this aquarium-corral: one of them consists in a nude wire circumscribing the inner part of the walls. The other metal piece is again a nude wire which circumscribes a cylinder installed in the middle of the struture, so that it remains fixed in the center of the swimming area. Both cable in the inner part of the walls and cable fixed on the central cylinder have to be connected to external terminals of a battery, so that one must reach the positive pole and the other must connect the negative pole of the battery. Alternatively, the cables in this basket can be substituted by conductive bars.



Figure 3.25: Perspective view of the basket, or aquarium-corral. The brown lines represent conductors, either nude cable or bars, each one being connected to a different polarity of the external circuit of power management.

Fig. 3.25 depicts the perspective of the basket. Fig. 3.26 illustrates the upper view, without (a) and with the eel (b). One can notice that, no matter the point of the aquarium-corral the eel is, both the head and the tail are near, respectively, to one



Figure 3.26: Upper views of the basket: (a) view of the basket without eels inside; (b) view of the basket with one eel swimming inside the corral. At any moment of the displacement of the eel, both head and tail are likely to be near to the collecting current points, which are the conductive cables.

wall and to the central cylinder. This proximity is of great interest given that it seeks to mitigate the impact of R_a , which is the resistance of the water path current.

In what concerns to the eels in this dissertation, we shall assume the overall harvesting scheme which is represented in Fig. 3.27. A bundle of baskets are deployed along a river which crosses the city, each basket containing one or more electric eels inside it. Connecting cables, either in parallel or in series, gather the energy coming from the baskets, leading them to the nearest Power Substation. The PS then manages the total eels instantaneous power and then injects it in the grid.

3.3.3.5 Eels available power

Electric eels under still water provide 10 mW steadily, as illustrated in Fig. 3.23 if we consider the average current of 1 mA.

The number of electric eels to be distributed along the river can vary reasonably. Despite the indeterminate number, a set of assumptions are welcome at this point. One assumption refers to the number of eels by basket: two or more eels can draw the electrical current among them, so that it is perhaps worth avoid joining eels in the same corral. As such, one electric eel per basket is assumed as a premise.

Other aspect is related to the form by which the energy produced by the eels will be stored. This dissertation aims to predict the series of maximum power demand values,



Figure 3.27: Proposed scheme for energy harvesting from eels in a river, which comprises eels inside the baskets. All the output cables are connected to a power manager circuit, which carries out this power to be conditioned.

which are likely to occur during the peak, i. e., the hours of the day in the end of afternoon and beginning of evening. In Brazil, the peak starts at 17h30, in average, and always lasts for 3 hours. If the eels produce energy along all the day and the power is used immediately, with no batteries in the substation to gather it up, then the amount of power received from the eels will vary around $P_e = 0.01n_e$ [W], where P_e is the power furnished by the eels and n_e is the number of eels.

Diversely, if the power can be stored, we can concentrate the Watts produced during 23 hours of the day and inject it into the grid only during one of the 3 peak hours. We can assume that for each 24 hours of storing energy, in only one hour we will use the energy. As such, say that the $P_{e-peak} = 24n_e$ [W], where P_{e-peak} is the power delivered by the batteries solely during the peak. Moreover, the period of one hour here described does not need to be continuous: it can be sliced over 15, 10 or 5 minutes, preferentially when the higher values of power demand are observed.

The importance of this consideration for our purposes is related to optimizing eels energy. Maximum power demand values are highly likely to occur during the peak. If we imagine we have 400 electric eels connected to one PS, storing energy all the day results in $P_{\rm e-peak} = 400 \times 24 \times 0.01 = 98$ W. This amount of power is near to 0.15 %,

in many cases, of the most severe peak values.

Undoubtedly, there must be a coefficient of storage losses, but it will be taken for slight due to the main goal of this study, which is about time series prediction, and because the size of our electric eel, being 90 cm long, may suffer changes. Therefore, as another result of this chapter, the energy will be considered to be $P_{e-peak} = 24n_e$ [W], as above-mentioned.

3.4 Summary of the three energetic sources

Table 3.5 provides a little scheme with a global view upon the main energetic charcteristics of each source. The RF energy harvester system, indoor light recycling circuit and electric eels have different profiles, according to the Table 3.5. All those are weekly results.

Table 3.5: Summary about three energetic sources: RF harvesters, indoor light recycling, electric eels.

Source	RF antennas	Indoor light	Solar PV	Eels
Daily hours of power (h)	24	12	12	24
Generation(G)/Recycling(R)	R	R	G	G
Weekly produced energy	0.144 Wh/ant.	$*42 \text{ Wh/m}^2$	5250 Wh/m^2	1.68 Wh/eel

* 42 Wh/m² refers to the weekly energy from indoor light when the panel is used only 12 hours a day. In the case we harvest indoor energy 24 hours, the result is 84 Wh/m².

Chapter 4 ENERGY DEMAND PREDICTION IN A STANDARD ELECTRIC POWER GRID

4.1 Overview

As said in Chapter 1, two stages of measuring are undertaken along this dissertion. The first stage handles power demand prediction with no renewable microgeneration sources being considered. Microgeneration sources are included only in the model during the second stage, which is described and developed in Chapter 5.

This chapter runs upon the first stage, in which Auto-Regressive (AR) model, Auto-Regressive with eXogeneous Inputs (ARX), Auto-Regressive Moving Average with eXogeneous Inputs (ARMAX) and Artificial Neural Networks (ANN) system are compared to each other, in order to determine which one is more suitable for predict data.

In what concerns to physical maintenance of the grid assets and energy quality, the most important parameter is the maximum power registered, since it indicates the most severe situation that the grid must to attend satisfactorily. Mean power data are more likely to be useful for economic analyses and about the evolution of market consumption. In this study, we consider that the maximum values prediction is the core activity, and therefore all the efforts will be dedicated to forecast its values. Thus, from now on, if we say "value" or "demand value" as isolated expressions, they must be understood as the maximum power demand in the respective week.

Eight PSs in Leipzig, Germany, had their demands registered from 2001 to 2004. These 4 years yield 209 registers for each PS. All power demand values were originally recorded in periods of 15 minutes each, emboding therefore sets of 96 daily power registers. From this domain, each upper daily value was fetched, and the maximum weekly power demand was the higher value among the seven daily maxima between sunday and saturday. Therefore, we hereafter consider only these maximum weekly values, which come from realistic meaurements from Leipzig city, in order to proceed to weekly demands prediction. The matrix corresponding to the demand values is therefore 209 \times 8, since there are 209 weeks for the 8 PSs. The experiment with Box & Jenkins models and ANN systems predict values only for the last 9 weeks for each Substation.

The parameter estimation of $[n_a, n_b, n_c, n_k]$ (see Appendix B for ARMA models) was determined by means of *Minimum Description Length* (MDL), which is better than ARMAX-NSSE, ARMAX-AIC and ARMAX-FPE for model order selection [48, 49]. The term n_a is related to the number of past outputs y(t) to be computed in the next prediction, whereas n_b is related to the number of exogeneous inputs in prediction, i. e., how many values of the exogeneous variable will be considered at each step; n_k informs how many instants before the forecasted output the first exogeneous value is considered. The number n_c expresses how many random noise inputs are brought into the model at each moment. Fig. 4.1 illustrates quickly how do the input and past output data influence the next output at each instant during the filter operation.



Figure 4.1: Visual representation of the set of values which influences the next forecasted output y(t) at each moment: underlining (a) values which influence output y(k+5) and, and in the next instant, (b) values which influence y(k+6).

The experiment starts employing AR, ARX and ARMAX models. Firstly, we have tested the AR, model which handles solely the past outputs values for the futures outputs to be forecasted. In sequel, we probed the validity of the insertion of temperature as an exogeneous variable, assembling ARX and ARMAX models. For the ARX, we tested temperature and logarithm of temperature values. This step was helpful since the prior supposition is that electric consumption increases with lower temperatures, in cold weather countries, as Germany. Fechner's law states that subjective sensation is proportional to the logarithm of the stimulus intensity [50]; therefore, since human perception of cold is logarithmic [51], we tested the suitability of the logarithm of temperature as a better exogeneous variable.

After determining n_a and n_b by means of MDL, we have established the number of entries for the ANN system. The reason for determining n_a and n_b at prior is due to the intended perfect parallel comparison between ARMAX and ANN: if we want to predict values based on the n_a and n_b terms, it is worth to fix the same input data for ANN. Clearly, we are considering that the same input values are likely to produce output values also in the ANN system.

The ANN structure is a Multi-Layer Perceptron (MLP) mesh of two hidden layers. Our MATLAB codes ran in order to achieve the best number of neurons in each hidden layer, limiting them to 6 neurons maximum. The ANN forecasts were based on these optimal quantity of neurons per layer. Then, the resulting MLP underwent the trainning period (first 200 weeks), returning outputs for the last 9 weeks. As far as we can see, our ANN system provided the best results for an ANN system bearing two layers, with 200 input data for the following 9 data forecasting.

The indicative numbers of each one of the 8 PSs are hereafter described by letters from A to H, by means of the following correspondence: A = 1104, B = 1108, C = 1109, D = 3092, E = 3219, F = 3244, G = 4012, H = 4254.

Consumption Data by Power Substation (PS). The 8 observed PSs present weekly maximum power demands as shown in Figure 4.2. The first week contains January 1st 2001 and the 209th week contains December 31st 2004. The length of demands vary over time, but the pattern is quite repetitive: as time indication approaches multiples of 52, there being the end of year and the coldest weeks of winter, electric consumption is enhanced. This fact inspired us to seek the participation of temperature as a relevant exogeneous input variable. In Fig. 4.2, graph for Substation H illustrates the period of time the forecast upon which forecasts will be done, for both AR models and ANN system, and for each one of the 8 PSs.



Figure 4.2: Maximum electric power demands over the 209 weeks: letters on the top indicate the PS. Yellow bar on H substation graph shows the forecasted period for every PS.

4.1.1 AR, ARX and ARMAX Models

First of all, we are supposed to seek the orders n_a for AR model, and n_a , n_b and n_k for ARX and ARMAX. One can find n_a and n_b given the delay factor n_k ; however, n_k is still undefined. One might adopt $n_k = 1$, in order to forecast the next week power demand; alternatively, $n_k = 9$ makes sense if we are going to forecast the 9 last weeks of the entire period. As a counterpoint, very large values of n_k are likely to limit our forecasting power, since it shortens the amount of available input data. The exogeneous information which is going to be considered in this study is temperature in its real and logarithmic values.

Encoutering n_k was possible by means of computing loss functions for ARX models of different orders. A loss function LF is calculated for each combination $[n_a \ n_b \ n_k]$. For each hypothetical n_k , we have assembled a matrix V, which is a matrix $(n_a \times n_b +$ $1) \times (4)$, whose lines are possible sets of $[n_a \ n_b \ n_k \ LF]$. Then, for each different n_k , a combination $n_a \ n_b$ was assigned taking into account the minor value of LF.

In sequel, we varied n_k from 1 until 208, given that we have taken on 209 sequenced values. We have thus obtained 208 LFs. We picked the smallest overall LF in order to achieve the final set $[n_a \ n_b \ n_k]$.

For every PS, one different V matrix was assembled, and therefore we should obtain a different set of $[n_a \ n_b \ n_k]$ for each different PS. However, all values coincided: $n_a = 10$, $n_b = 10$ and $n_k = 1$, being the optimal model for the 8 surveyed PSs.

Let us look at one of ARMAX generated equations to notice the role of each element, including n_c , the moving average factor. For Power Substation G, the following set of expressions was achieved.

$$\begin{split} \mathbf{A}(\mathbf{z}) &= 1 - 0.5861 \ \mathbf{z}^{-1} - 0.6502 z^{-2} - 0.5762 z^{-3} - 0.3488 z^{-4} + 0.8517 z^{-5} + 0.6843 z^{-6} + \\ &\quad 0.2958 z^{-7} - 0.04026 z^{-8} - 0.8855 z^{-9} + 0.2598 z^{-10}, \\ \mathbf{B}(\mathbf{z}) &= -229.8 \ \mathbf{z}^{-1} + 118.4 z^{-2} - 27.16 z^{-3} + 407.6 z^{-4} + 17.88 z^{-5} - 250.2 z^{-6} - \\ &\quad 100.3 z^{-7} - 7.266 z^{-8} - 90.54 z^{-9} + 161.4 z^{-10}, \\ \mathbf{C}(\mathbf{z}) &= 1 - 0.2894 \ \mathbf{z}^{-1} - 0.5455 z^{-2} - 0.9145 z^{-3} - 0.7849 z^{-4} + 0.6069 z^{-5} + 0.8731 z^{-6} + \\ &\quad (4.1) \qquad 0.5833 z^{-7} + 0.4015 z^{-8} - 0.9066 z^{-9} - 0.0225 z^{-10}. \end{split}$$

The meaning of $[n_a \ n_b \ n_c \ n_k]$ becomes clearer now: there are $n_a = 10$ elements in A(z), the auto-regressive component. Other $n_b = 10$ elements come from the exogeneous input variable, which is delayed in $n_k = 1$ instant. The first element of B(z), $[-229.8z^{-1}]$, is delayed in one instant in relation to the first element of A(z), which is 1. For the set of expressions in (4.1), Akaike FPE is 370.8 and fit to estimation data is 57.14%. Fit estimation data corresponds to the prediction focus.

If we made $n_c = 1$, instead of $n_c = 10$, Akaike FPE would be 518.7, and the prediction focus would fall to 49.72%. On the other hand, making $n_k=9$, with $n_a = n_b = n_c = 10$, we would produce FPE = 448.2 and a prediction focus of -331.9%. These observations proves the reasonableness of our choice in equalizing $n_a = n_c$. One interesting aspect in selecting $n_k = 1$ is that it allows to enlarge our set of input data, since large values for n_k constrains our first possible exogeneous input to come in the system only in the occurence y(b + k + 1), given that the exogeneous input set of values is complete only in this moment.

4.1.1.1 eXogeneous Input: Temperature

After determining the general model order, we compare now AR with two ARX models, being the first ARX dependent on temperature data and the second ARX dependent on logarithm of temperature ¹. The logarithm was obtained by means of summing 273° C to the pure temperature data and, in sequel, applying the logarithm upon the result, in order to avoid problems with Celsius negative values. We probed if temperature has a logarithmic impact, in face of human perception. After that, an ARMAX model was tested with the same parameters n_a , n_b and n_k ; the parameter n_c , related to the Moving Average component, received the same value for n_a , which corresponds to the number of past input values.

We have examined performances of AR, ARX (X = Temperature) and ARX (X = Log Temperature) with two goals: testing if the addition of an exogeneous input would really be advantageous, and, whether positive, assessing if the Log Temperature would actually overcome Temperature as a good exogeneous variable.

Table 4.1 shows the set of results to each one of the 8 Power Substations. Prediction focus is higher the more accurate is the model, and assumes greater values for ARX with

¹Temperature data were obtained from http://www.wunderground.com/

	AR		ARX		ARX	
\mathbf{PS}			(Temperature)		(Log Temperature)	
	pred. focus	FPE	pred. focus	FPE	pred. focus	FPE
А	44.39~%	134.1	47.71 %	130.1	48.48~%	126.3
В	35.70~%	336.3	42.68~%	290.7	42.98~%	287.6
С	66.26~%	327.5	68.77~%	305.1	68.78~%	304.8
D	46.84~%	144.1	51.59~%	129.9	52.03~%	127.5
Е	40.93~%	168.1	46.24~%	151.4	46.74~%	148.6
\mathbf{F}	40.97~%	328.1	45.80~%	301.3	46.10~%	298.0
G	45.99~%	515.6	49.90~%	498.2	49.70~%	486.1
H	46.05~%	761.7	50.70~%	691.7	50.99~%	683.7

Table 4.1: AR and ARX models performances in terms of prediction focus and FPE, considering temperature and log temperature as eXogeneous variables in ARX cases.

logarithmic values of temperatures. The same may be said for Akaike's Final Prediction Error (FPE) criterion, which provides a measure of model quality by simulating the situation where the model is tested on a different data set [47].

These results were obtained by making $n_a = 10$, $n_b = 10$ and $n_k = 1$, and they allow us to adopt ARX with the eXogeneous data as log[T], where T is the temperature. The definition of the logarithm of temperature as the main exogeneous input was a necessary step before exploiting ANN system, since in the neural networks the exogeneous input is also present. ANN and ARMAX dealed only with logarithmic values of temperature.

4.1.2 Artificial Neural Networks

The Artificial Neural Networks (ANN) used in this work comprises two hidden layers. In order to encounter the best number of neurons to integrate each hidden layer, we established first the limit of 6 neurons per layer, whereas the minimum number of neurons was fixed in 2. In sequel, the ANN system was trained considering all possibilities in terms of quantity of neurons per layer. The number of simulations for the 2-hidden layers in this scheme is 5^2 , i. e., (limit - 1)².

After each single simulation, the results were compared to the real data, and the difference was assigned to each one of the 72 elements of an error matrix. The mean square error of the 72 results was then obtained. ANN system achieved the best

performance for 6 neurons in the first layer and 2 neurons within the second hidden layer. Such deployment of neurons was adopted as that one which would give us the final best result coming from our ANN system.

Input data for our ANN system were similar to the input data for AR models as possible. Since the obtained AR schemes receive data from 10 past outputs and 10 exogeneous values, the same was made with respect to the ANN system. Therefore, for each trainning/validation step, given an expected result y(t), we counted on 10 past output values, y(t-1) until y(t-10), and 10 past exogeneous input values, u(t-1)until u(t-10), in the same way it was carried out for ARX models. No noise was added, hence $n_c = 0$.

4.1.3 Comparison Between AR, ARX, ARMAX and ANN Results

After predictions, we achieved the error for each element of the forecasted 9×8 matrix. This error matrix consists in the forecasted values by AR, ARX, ARMAX and ANN, subtracted by the realistic power demands. One error matrix was assembled for each model.

In sequel, we calculated the Mean Square Error (MSE), that is, the sum of all squared errors divided by 72, the number of matrix elements. Another criterion was also tested: the number of forecasted results standing above the real power demand value, yielding a "Positive Deviation". This value constitutes an important outcome, since our aim is to prevent maximum power values in the grid. This criterion is therefore called security criterion. The Number of Positive Deviations (NPD) is the amount of predictions which stood above the effective maximum weekly demand.

For $n_a = n_b = n_c = 10$ and $n_k = 1$, ARMAX model presented the best performance among all tested models. Table 4.2 shows MSE and NPD values for AR, ARX, ARX log[T] and ARMAX log[T], where log[T] indicates that the model considers the logarithm of temperature as the exogeneous variable.

One may notice that the MSE increases from ARMAX to ARX log[T], continuing to raise through ARX, AR and ANN. The number of values above the real measured ones, NPD, follows the same sequence. One observation we must do is that the ideal NPD would be 36, in face of the mathematical accuracy; the further from 36, the more biased is the method, because it means that our final outputs stay sistematically below

Table 4.2: Mean Square Errors (MSE) and Number of Positive Deviations (NPD) for AR, ARX, ARX log[T] and ARMAX log[T], $n_a = 10$, $n_b = 10$ and $n_k = 1$

Parâmetro	AR	ARX	ARX $\log[T]$	ARMAX $\log[T]$	ANN
MSE	2.7409	2.6457	2.5800	2.1771	4.6844
NPD	29	29	32	38	21

or above the target. ARMAX is the less biased, whereas ANN seems to be the most one. Since our specific goal is to provide the grid to support occasional extra power demands with optimal investments, we can elect NPD parameter as important for our purposes.

Absolute values of prediction deviations with regards to the real values were also registered in order to provide an effective comparison among all models. Matching all models to each other, we assigned a combination of PS/week whose result is the name of the model which provided the lowest absolute value for the error. The overall result is shown in Figure 4.3. The superiority of ARMAX is patent: ARMAX prevailed in 34 occurrences, or 47.22 % of the 72 events. The model which has provided the second more accurate set of results was AR, with 12 occurrences (16.67 %). Both ARX log[T] and ANN appeared 9 times (12.50 % each), whereas ARX has shown up 8 times (11.11 %).

week nº	Power Substation								
WEEK II-	Α	В	С	D	E	F	G	Н	
201	AR	ARX log	ARMAX	ARX log	ARMAX	AR	ARMAX	ARMAX	
202	AR	AR	ARX	ARMAX	ARMAX	ANN	ANN	ARMAX	
203	ARX	ARMAX	ARX log	ARMAX	ARMAX	ARMAX	AR	ARX	
204	ARX	ARX log	AR	ARX	ARMAX	ARX	AR	ANN	
205	ARX	ARMAX	ARMAX	AR	ARMAX	ARMAX	ARX log	ARMAX	
206	ANN	ANN	ARMAX	ANN	ARX log	ARMAX	ARX log	ARX log	
207	ANN	AR	ARX	AR	ARX log	ARMAX	ANN	ANN	
208	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	
209	ARMAX	ARMAX	ARMAX	ARMAX	AR	ARMAX	ARMAX	AR	

Figure 4.3: Models which have provided the minor deviation absolute values, for each week and PS.

The absolute deviation values for each best model, given a PS and a week, was divided by the actual measured demand value. We remark that each deviation value was calculated subtracting the effective measured data from the predicted value by a model. Therefore, negative values within Fig. 4.4 mean that predicted value was less than the realistic one.

week nº	Power Substation								
WEEK II-	Α	В	С	D	E	F	G	Н	
201	<mark>-6,60</mark> %	5,95%	1,84%	0,14%	-2,62%	-6,55%	-0,51%	-0,17%	
202	-6,89%	0,27 %	-1,25%	-1,91%	-0,21%	1,52%	1,36%	-6,54%	
203	-0,37%	0,51%	0,24%	-2,08%	0,48%	4,19%	5,46%	0,75%	
204	-9,52%	1,61%	0,62%	-8,99%	-7,81%	-0,92%	-2,60%	7,35%	
205	-31,43%	-0,02%	-2,13%	13,71%	7,35%	-2,69 %	-1,25%	0,01%	
206	12,03%	-3,51%	2,17%	0,81%	0,64%	0,05%	0,21%	0,14%	
207	-3,76%	-1,37%	-0,05%	-2,32%	1,43%	3,68%	-2,91%	1,41%	
208	0,25%	-7,03%	-1,33%	-4,24%	- <mark>9,79</mark> %	-14,15%	-5,54%	-18,47%	
209	1,09%	7,62%	4,33%	6,53%	-0,20%	-0,83%	-3,87%	-5,03%	

Figure 4.4: Absolute deviation values from Fig. 4.2 divided by the respective measured values, in percentages

Among the set of 72 predictions, only 5 result errors stood above 10 % in absolute value. Errors below 5 % occurred in 50 cases, more than in 2/3 of the entire set. Errors below 1 % returned a remarkable quantity of 23 occurrences, that is to say, one for each three predictions presented such accuracy.

Despite ANN and ARX $\log[T]$ have offered the same number of best predictions, if we carry out a match exclusively between ARX $\log[T]$ and ANN, we obtain Figure 4.5, which shows whether ANN or ARX $\log[T]$ offer the best results.

ARX log[T] prevailed in 56 occurrences, or 77.78 % of the 72 events. We can suppose that ANN behaves as a sub-optimal forecasting model, since the mean error is always elevated in face of AR models.

Given that ARMAX log[T] and ARX log[T] present the lowest MSE and less biased NPD, we can consider that the best between both is the best of all models. The result

wook nº				Power Su	ubstation			
Week II-	Α	В	С	D	E	F	G	H
201	ARX	ARX	ARX	ARX	ARX	ARX	ARX	ARX
202	ARX	ARX	ARX	ANN	ARX	ANN	ANN	ARX
203	ARX	ARX	ARX	ARX	ARX	ARX	ARX	ARX
204	ARX	ARX	ARX	ARX	ARX	ARX	ARX	ANN
205	ARX	ARX	ARX	ARX	ARX	ARX	ARX	ANN
206	ANN	ANN	ARX	ANN	ARX	ARX	ARX	ARX
207	ANN	ANN	ARX	ARX	ARX	ARX	ANN	ANN
208	ARX	ANN	ARX	ANN	ARX	ARX	ARX	ARX
209	ARX	ARX	ARX	ANN	ARX	ARX	ARX	ANN

Figure 4.5: Score of minor deviation absolute values, matching ARX $\log[{\rm T}]$ versus ANN, for each week and PS

is shown in Figure 4.6, in which ARMAX proved to be the best model in 48 occurrences (66.67 %).

wook nº	Power Substation									
week II=	Α	В	С	D	Е	F	G	Η		
201	ARX	ARX	ARMAX	ARX	ARMAX	ARMAX	ARMAX	ARMAX		
202	ARMAX	ARX	ARMAX	ARMAX	ARMAX	ARX	ARMAX	ARMAX		
203	ARX	ARMAX	ARX	ARMAX	ARMAX	ARMAX	ARX	ARX		
204	ARMAX	ARX	ARX	ARX	ARMAX	ARX	ARX	ARMAX		
205	ARMAX	ARMAX	ARMAX	ARX	ARMAX	ARMAX	ARX	ARMAX		
206	ARMAX	ARMAX	ARMAX	ARX	ARX	ARMAX	ARX	ARX		
207	ARX	ARMAX	ARMAX	ARMAX	ARX	ARMAX	ARX	ARX		
208	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX		
209	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX	ARMAX		

Figure 4.6: Score of minor deviation absolute values, from the match ARMAX log[T] versus ARX log[T], for each week and PS

For all these reasons, ARMAX log[T] is concluded to be the best model for utilization over data without microgenerators analysis.

Figure 4.7 makes possible an overall perception about the performance of each model here described. This Figure depicts the forecasted values for the Power Substation D, in counterpoint with the real data, only for the final 9 weeks.



Figure 4.7: Measured data and forecast data for Power Substation D (3092), during the 9 last weeks of 2004.

Curves from Fig. 4.7 do not illustrate continuous values, but rather discrete outcomes; however, in order to make it more visually comprehensible, lines were designed to connect the output values.

We perceive that forecast data from ARX and ARX log[T] curves are almost the same, differing by a little offset; particularly in this PS, ARX seems to be nearer to the measured data. However, ARX log[T] has been considered more successful over the entire set of PSs, and therefore we can conclude that this PS in particular may constitute an outlier with regards to ARX and ARX log[T]. Values from AR projections are very close to ARX and ARX log[T], and like them, AR makes a peak value in the 5th week, as well as ARMAX and ANN.

However, ARMAX was the only model capable of forecasting the peak of the 8th week. The ANN curve did it also, however, it was more likely to be a recuperation from the erratic plunge occurred in the 7th week, where the measured data indeed get lower, but not so sharply.

Another remark can be made upon the 4th week, where ARMAX apparently drifts apart from the measured data. After this deviation, it lasts only 3 weeks to get the trend again, whereas AR, ARX and ARX log[T] seem to present a very slower reaction. The unstability of ANN is not observed among the AR models (AR, ARX and ARMAX).

Curves from Fig. 4.8 are now applied to the remaining Power Substations, in order to enable a complete assessment of the prediction models over the entire set of PS.



Figure 4.8: Measured data and forecast data for Power Substations A, B, C, E, F, G and H during the 9 last weeks of 2004.

After concluding that ARMAX is the most suitable model for undertake forecasts, we now illustrate the result of ARMAX predicitons for 2005 and 2006. Fig. 4.9 shows curves for the 8 Power Substations, being blue lines for measured data and red lines depicting ARMAX forecasts. Both years will be important for our analyses within the next chapter, which will consider 2005 and 2006 forecasts.

4.2 Leipzig Map for the Power Substations

The main characteristics of each region is hereafter outlined, based on [52].

Power Station 1104 lies in the Meusdorf District, Southeast of Leipzig. This district has a very low Demographic Density, and on average has 1.9 up to 2.0 inhabitants per house. Population growth in this area is 9 % to 15 %. Circa 70 % of those residents are economically active, and on average they think that their economic situation is stagnated.

Power Station 1108 is located in in the Gohlis-Mitte District, Center of Leipzig. Power Station 1109 lies in the Gohlis-Nord District, Center-North of Leipzig. Both districts have high Demographic Densities, and on average both have 2.2 or more inhabitants per house. Population growth for the two districts is between 9 % and 15 %. Circa 70 % of Gohlis-Mitte residents are economically active, whereas this rate for Gohlis-Nord District is 80%.

Power Station 3092 has place in the Schonefeld-Ost District, Center-Northeast of Leipzig. Power Station 3219 lies in the Paunsdorf District, East of Leipzig. Both districts have medium demographic densities. Population growth in Schonefeld-Ost is 3 % to 9 %. In Paunsdorf, population growth is in the range from -3 % to 3 %. Approximately 60 % residents are economically active, for both areas.

Power Station 3244 is located in the Heiterblick District, East of Leipzig. This district has a medium Demographic Density, and on average has under 1.9 inhabitants per house. Population growth in this area is negative, lower than -3 %. Circa 50 % of those residents are economically active, and on average they think that their economic situation is worsening.



Figure 4.9: ARMAX forecasted maximum electric power demands for 2005 and 2006, for every PS: the blue line relates to measured data, whereas red lines describe ARMAX prediction data. 58

Power Station 4012 lies in the Grunau Siedlung District, West of Leipzig. This district has a high Demographic Density. Population growth in this area is 15 % or higher. Circa 50 % of those residents are economically active. Power Station 4254 lies in the Lausen Grunau District, West of Leipzig. This district has a very low Demographic Density, and on average has under 1.9 inhabitants per house. Population growth in this area is 9 % to 15 %. Circa 60 % of those residents are economically active, and on average they think that their economic situation is improving.

Preparing the spatial analysis, Leipzig map is introduced, and the location of each one of the 8 PS is signalized, as shown in Fig. 4.10.



Figure 4.10: Leipzig Map with the 8 Power Stations located [52].

Chapter 5 ENERGY DEMAND PREDICTION INCLUDING RENEWABLE ENERGY GENERATION

Along the previous chapter, we studied the adequacy of the insertion of temperature as an exogeneous variable, and we calculated n_a , n_b , n_c and n_k so that we achieved satisfactory auto-regressive models, as well as neural networks. These tasks were undertaken in order to compare the several models each other, from which we concluded ARMAX as the most suitable model in forecasting future maximum power demands, given the quantity and conditions of power demand data.

In the current chapter, we evaluate the insertion of renewable sources into consumer houses and buildings. The same 209 weeks maximum power demands (2001-2004) were admitted as input data. We considered possible energetic incomes from for microgeneration, whose values are based on the power that the three sources can afford steadily, as shown in Table 3.5. The three sources - RF, PV panels and electric eels - are supposed to operate only over the last 157 weeks, which corresponds to the last three years - 2002, 2003 and 2004.

New values for maximum power demands are then obtained from the difference between the realistic values and those coming from the renewable sources. These new set of values will thus be adopted as another set of input data.

Given that during 2001 there is not yet input from the emergent sources, but only during the 2002-2004 period, and given that these renewable sources come in the grid progressively, we simulated new trends for the power demand values seen by each PS. After that, we trained our ANN and adjust our ARMAX according to the weeks between January 2001 until October 2004, and we have then evaluated the adaptation ability of each method to forecast values for the 9 last weeks of 2004.

Figure 5.1 shows the underlying idea for characterization of input data. Let us say we have the realistic power demand d_1 , and the apparent power demand is d_2 : d_1 is the electric power that a set of consumers spends on their activities, whereas d_2 is the effective electric burden that the PS of these consumers has to sustain. The difference between d_1 and d_2 is microgeneration power $P_{\mu g}$. In the beginning of the process, d_1



Figure 5.1: Example of electric demand series: d1 is the overall demand from consumers, d2 is the energy the PS furnishes, and $P_{\mu g}$ is the power microgeneration, so that $P_{\mu g} = d_1 - d_2$.

and d_2 are equal. However, in a certain moment, microgeneration starts to play the role of diminishing the apparent power demand d_2 .

Our system is projected to forecast future values of power demands without the knowledge about how many microgenerators are connecting to the grid. Such premise is very likely to be accepted in real systems, since we can only estimate, in most of times, how many kW are being generated now, and how many kW will come up in the grid next.

One of the most important parameters that Fig. 5.1 contains is the moment in which the microgeneration starts. In our simulations, several instants were adopted so that we could evaluate the role the instant of beginning of generation plays in the whole system. In the same way, the rythm by which the generation increases was a changeable parameter in our analysis.

5.1 Parameters of Temporal Evolution for Renewable Sources

At prior, we estimate the rythm of adoption of the diverse items composing the renewable sources. Indoor light and RF energy harvesting systems are supposed to vary according to market insertion, since the photovoltaic (PV) panels are part of household appliances. We assume that RF harvesting systems will be spread over houses in the same rate as the PV panels, given the probability of a person who recycles energy from indoor light will also adopt RF recycling schemes.

5.1.1 Panels and antennas growing rates

The main task in this subsection is to project the number of antennas and panels that people into a city will install in their homes. We suppose that the indicators of this city and the general economic indicators of Brazil are alike. We divide this analysis into the main trend and the oscillation that exists in the evolution of real data. We calculate the main growing trend for PV panels and antennas, and we will apply over them white noise values in order to simulate an inconstant measurement, as we observe in the real world.

5.1.1.1 Main trend

According to Brazilian Telecommunication Regulatory Agency - ANATEL¹, in 1990 there were about 667.000 cell phones in the country; in 1991, the number arose to 6.7 million (about 10 times) and, in the next year, the quantity achieved around 31.726 million cell phones. There is a logarithmic pattern within the increasing quantities with regards to the beginning of the process. We depict the logarithm of the evolution of cell phones in Brazil, considering the period from 1990 on, in Fig. 5.2(a). Fig. 5.2(b) illustrates the evolution of cable television users in the country, between 1993 and 2010, still according to data from Anatel.



Figure 5.2: (a) Increasing annual data for cell phones in Brazil since 1990; (b) Increasing annual data for cable TV in Brazil since 1993 (Anatel).

¹In portuguese, Agência Nacional de Telecomunicações - ANATEL

Tangents in the beginnings of graphs in Figs 5.2 are, respectively, about 1.1 and 1.5. Since these tangents are annual, it might be worth to keep in hands weekly variations, which are therefore near to 2.53%, or $e^{0.025t}$, t in weeks.

Both logarithmic graphs perform ascendant curves of different slopes, seeming straight lines in their beginning parts. This observation can clarify and explain how a new commodity is acquired by people: the lower the quantity, the slower the rythm of adoption, as the increasing rate seems to be exponential.

In order to introduce our mathematical generation model, Fig. 5.3 illustrates two distinct exponential curves, which describe temporal evolution for g_1 and g_2 generation processes. Say that for each exponential we count on the structure $g_t = c_n e^{a_n t}$, where g_t is the current amount of power which is furnished by the generation source on the t-th week, c_n is the initial amount of the unit generation (antenna, eel or square meter of PV panel) and t is time, in weeks. On the exponent, there is the factor " a_n " as the logarithmic value of annual increasing rate, being tehrefore a factor which expresses how fast the source increases.



Figure 5.3: Two curves expressing temporal evolution parameters for different sources.

The initial amount c_n is the quantity of the source in the instant it starts to generate power, being arbitrated for each source through several simulation cases. Time t is measured in weeks from the instant the respective source steps in the grid. For instance, if the first eel comes up in the system on the 23rd week, and the first solar PV panel enters only on the 40th week, than in the 50th week we have $t_1 = 27$ for eels generation, but $t_2 = 10$ for solar panels.

5.1.1.2 Determining a, a better exponent of e

Say M_{pv} is the number of square meter (m²) of PV panels that people install in their houses. In order to confer a higher accuracy to our analysis, we look for determining a, the exponent of e, which is multiplied by c in the term [ce^{at}]. Figure 5.4, from Internatitional Energy Agency Photovoltaic Power System (IEA PVPS), year 2013, depicts the evolution of grid-connected and off-grid PV systems from 1992 on in Japan, country which has taken as an example for the analysis.



Figure 5.4: Evolution of grid-connected and off-grid PV panels, in MW (IEA PVPS), in Japan.

Fitting the sequence of annual data obtained from this graph by means of regressions, the linear function returned the determination coefficient $R^2 = 0.7193$; power function provided $R^2 = 0.9039$; polynomial function returned $R^2 = 0.9321$. But the best result was drawn from exponential function, which reached $R^2 = 0.9813$, by means of the equation

$$P = 16.431e^{0.2982t},\tag{5.1}$$

where P is the cumulative installed capacity [MW] and y is the time in years. Comparatively to the cable TV and cell phones to evolve in Brazilian market, the smaller exponent in Eq. 5.1 can be explained by differences among products and between both countries, and because the spread of power generation appliances is believed to be not so fast as electronic devices. We therefore take on the supposition by which the Brazilian market bears seeming rates of solar panels spread related to grid-connected generators, with an accepted value of a = 0.3 for the exponent of e.

Since an year has 52 weeks, the area of panels are likely to increase at the rate of $e^{(0.3 \times \frac{t}{52})}$, or $e^{0.0058t}$, t in weeks. It corresponds to a weekly increment of 0.58 %. By means of this parameter, we can obtain the weekly increment of the number of PV panels and RF harvester antennas, considering an initial value.

5.1.1.3 Energy effectively available coming from PV panels and RF harvesters

As shown in Chapter 2, weekly average of global solar irradiance in Brazilian cities is 5250 Wh/m². This value expresses the net energy which is available, after taking into account the efficiency of the panel. Therefore, the number of square meters of existing PV panels in the week n, M_{pv}^n , will be directly multiplied for 5250 Wh/m², in order to achieve the generated power in that week and for the respective PS.

With respect to the indoor light energy generation, we suppose at least 12 hours of indoor generation daily, based on the parameter demonstrated in Chapter 2, of 0.103 Wh/m^2 . This parameter will also be multiplied by M_{pv}^n in order to dimension the indoor part of weekly PV generation.

Regarding to antennas available energy, we have produced also in Chapter 2 an energetic reference of 2.058 mWh per antenna per day as the available energy. One can notice that the total power obtained from indoor light and RF energy harvesting reaches 0.105 Wh/m^2 , if we consider one antenna for each squared meter of PV panel.

Once we have obtained these values, we discount them from the original weekly data of maximum consumption, in order to produce new values of power which is drawn from the grid. The same step will be executed with regards to the power drawn by the eels.

5.1.2 Electric Eels Aquariums in the River

Let us suppose that the city hall is undertaking the duty of arising eels. The administrative staff has concluded that an acceptable growing rate of e^1 a year for the baskets. Each basket has area of $1m^2$ and contains one eel with 90 cm of size in average. Hence, at the end of the first year, there will be around 108 eels generating energy; in the end of the second year, the city will have approximately 296 eels distributed in 296 baskets in the river. The reason for adopting an exponential serie of data lies on the need of confering to eels the same treatment as to the other sources.

As demonstrated in Chapter 2, the available power in one hour during the peak is $P_{e-peak} = 24n_e$ [W], where n_e is the number of eels generating energy.

5.1.3 Unilateral white noise upon constant values: the k factor

Before we carry on, it is necessary to introduce a variable which provides the noise character for temporal evolution of our sources. We suppose that unilateral white noise values are indicated to confer oscillations to the previously calculated g_t .

Consider an increasing series of discrete data y(n), as the weekly power demands. A variable k assumes random instantaneous values between -A and A, being that A is arbitrated; then, each outcome k multiplies the difference $\delta(n)$ between two consecutive data y(n) and y(n+1), for each n. Each result is then summed to y(n). We establish that y(n+1) is not allowed to decrease in relation to y(n), and, therefore, occasional negative values of $\delta(n)$ makes the current value y(n+1) to be equal y(n). The term "unilateral" comes from this criterion. The boundary value of k is chosen to simulate variations which may overcome weekly increasing values, being rather an arbitrary interval A. For a better comprehension of was just been said, Figure 5.5 is depicted.



Figure 5.5: Diagram for the routine by which unilateral white noise k is inserted into original exponential series.

Each t corresponds to the week, $\delta(n)$ is the gap of exponential increasing on the n^{th} value, and k is set to vary up to A = 2.5 in absolute value.

Figure 5.6 presents two graphs (a) and (b). In (a), it illustrates the pure exponential ascendant, and then in (b) it shows the sequence of values bearing oscillations originated by the k factor adoption. The vertical axis, global rate, shows how many times the initial value is multiplied. It is possible to see that the initial value of c is multiplied by approximately 14, up to the end of the second year, whether we take into account the annual increasing rate of $e^{1.3}$, as in this example.



Figure 5.6: Global rate over two years, firstly considering pure exponential $e^{0.025t}$ evolution (a), and then handling values with oscillation upon $e^{0.025t}$ (b). These last values are believed to be more probable into the market environment.

From Fig. 5.6, we conclude that k factor stands for the step-shape of exponential ascendant curve. The greater the k, the more the curve becomes similar to a stair.

5.1.4 Results for PV panels and antennas

Results from the main trend were summed to white noise values to produce the "probable market series", that is, the values that people are likely to originate in practice.

In sequel, each value of the probable market series was truncated to the nearest integer, so that all values of PV panels area and RF antennas are also integer. However, truncating values is possible only when the final values are available, and to produce them, we have to elect initial values. Assuming the initial quantity of panels to be 50 m^2 per Power Substation, we simulate random values for each PS as explained previously. The result is depicted in Figure 5.7.



Weekly growing rates after applying k factor, k = 2.5

Figure 5.7: Weekly actual values of square meter PV panels and RF antennas for the initial value of 50 m² in the considered PS.

5.2 Forecasting overall results considering power micro-generation values

Considering the parameters we have just presented, a brief analysis will be carried out upon different values for each one of them. For an easier data comprehension, we remember the definitions of the parameters as follows:

- w, in weeks, expresses the time during which the renewble emergent sources will generate power, within the 209 weeks. For instance, if w = 52, then the generators will step in at the 156th week, generating during 52 weeks until the last week of the whole period;
- *a*, scalar, the annual exponent of *e* in the curve that illustrates quantities of each source over time. Parameter *a* comes from market. We have resorted to Japanese market data to arbitrate *a* as 0.3, in principle;
- k factor, which brings a noise pattern to be added to the original perfectly exponential evolution for number of household appliances;
- c, in m² of PV panels, units of antenas or eels: the amount of units of each source when t starts to run, corresponding to the initial amount of units of a source in the first week it generates energy.

We divide our analysis according to the solar panel utilization. The first case considers that during sunny hours outdoor generation takes place, and we make use of the panels for indoor light solely during evening and nights hours. In the second case, we avoid outdoor generation, employing panels 24 hours inside buildings. The other two sources - eels and RF antennas - will work in both cases, not depending on sun performance. This division makes sense since solar panels yield 125 times more power in outdoor environments, according to Table 3.5.

5.2.1 Solar PV Panels Generating Outdoor and Indoor

With solar panels generating in outdoor environments during day hours light, we developed simulations in two main cases: microgeneration running until 2 years, w going from 35 until 80, and microgeneration running over more than two years, w starting on 105 and going until 150.
5.2.1.1 Microgeneration Running Until 2 Years

We have firstly settled some different possible values for t, a and k: t assumed values 35, 50, 65 and 80; a assumed 0.3, 0.6, 0.9 and 1.2; and k was 1, 2 or 3. Since for t = 80 we have tested only a = 0.3 and a = 0.6, we have 42 results in hands.

Under these combinations, Table 5.1 gather results concerning to the MSE for each method of the previous chapter, and also the best number of hidden layers for the ANN structure, which has been used for ANN results.

For until one year of operation, the number of the best hidden layers for ANN system becomes 5 for the first hidden layer, and 3 for the second one. There is not a precise explication upon this fact, however we can remark that for the evaluated periods, both quantities kept themselves practically constant.

Results for MSE are in accordance with has been shown in Chapter 4 about the order of the most accurate model to the less accurate one. Table 5.1 shows that ANN MSE stays above 4.0 in the most of time, being 3.8152 the minimum value in the entire set of outcomes. The second less precise system, AR, had as its worst MSE the value of 2.7458, circa 7/10 of the best MSE for ANN. An overview let us notice that outcomes for this table do not vary significantly.

With respect to the hypothesis of generation running during more than one year, we have data from Table 5.2. We can behold the continuity with respect to the main patterns of Table 5.1: ARMAX presenting the less errors, but, now, the best numbers of neurons per layer in ANN system vary.

Some results are quite clear: the order by which methods are precise is the same as in Chapter 4, i. e., ARMAX, ARX log[T], ARX, AR and ANN. In other words, the fact of microgeneration starts to operate until 2 years from the end of 2004 does not breed alterations on the quality of each model. One can observe that ARMAX is the only method that reaches an MSE below 2.0, occurring on the input set w = 80, a = 0.6 and k = 2.

There is no longer an answer for the best couple of hidden layers into ANN mesh. One possible explanation for it lies on time stretching: comprising more weeks far from the end of the process, the exponential tendency has more time to appear. This tendency

w	a	k			MSE			ANN:	best
			AR	ARX	ARX_{log}	ARMAX	ANN	1^{st} layer	2^{nd} layer
35	0.3	1	2.7433	2.6472	2.5517	2.0847	4.4209	5	3
35	0.3	2	2.7434	2.6472	2.5516	2.0929	4.5037	5	3
35	0.3	3	2.7434	2.6472	2.5514	2.1294	4.3253	5	3
35	0.6	1	2.7433	2.6472	2.5515	2.0932	4.4973	5	3
35	0.6	2	2.7433	2.6471	2.5512	2.1136	4.3451	5	3
35	0.6	3	2.7433	2.6471	2.5508	2.0866	3.8152	5	5
35	0.9	1	2.743	2.6467	2.551	2.1078	4.3185	5	3
35	0.9	2	2.7431	2.6468	2.5504	2.1209	4.1445	5	3
35	0.9	3	2.743	2.6466	2.5497	2.1399	3.8683	5	5
35	1.2	1	2.7427	2.6463	2.5503	2.1261	4.1812	5	3
35	1.2	2	2.7427	2.6462	2.5492	2.1357	4.1277	5	3
35	1.2	3	2.7423	2.6457	2.548	2.1251	3.9861	5	5
50	0.3	1	2.7442	2.6533	2.5589	2.0522	4.3901	5	3
50	0.3	2	2.7443	2.6534	2.5588	2.05	4.9612	5	3
50	0.3	3	2.7443	2.6534	2.5586	2.0757	4.5628	5	3
50	0.6	1	2.7443	2.6534	2.5587	2.0562	4.425	5	3
50	0.6	2	2.7444	2.6533	2.5583	2.0801	4.2944	5	3
50	0.6	3	2.7445	2.6533	2.5579	2.085	4.6239	5	3
50	0.9	1	2.7445	2.6533	2.5584	2.0851	3.969	5	3
50	0.9	2	2.7447	2.6533	2.5577	2.0916	4.5995	5	3
50	0.9	3	2.745	2.6532	2.557	2.1974	4.2372	5	3
50	1.2	1	2.7447	2.6532	2.5579	2.087	4.1514	6	2
50	1.2	2	2.745	2.6535	2.5569	2.1633	3.8602	5	3
50	1.2	3	2.7458	2.653	2.5556	2.1236	3.8253	5	3

Table 5.1: MSE for AR, ARX, ARX $\log[T]$, ARMAX and ANN, and the best number of hidden layers for ANN system in function of each combination of w, a and k, for periods of 35 and 50 weeks.

w	a	k			MSE			ANN:	best
			AR	ARX	ARX_{log}	ARMAX	ANN	1st	2nd
65	0.3	1	2.7411	2.6491	2.5585	2.063	4.2143	4	4
65	0.3	2	2.7411	2.6491	2.5583	2.0622	3.9726	5	6
65	0.3	3	2.7411	2.6491	2.5582	2.0652	4.6301	5	6
65	0.6	1	2.741	2.649	2.5581	2.0671	4.027	5	4
65	0.6	2	2.7407	2.6485	2.5571	2.034	3.8377	5	4
65	0.6	3	2.7406	2.6482	2.5564	2.0991	5.8055	5	6
65	0.9	1	2.741	2.6489	2.5575	2.0848	4.3505	5	6
65	0.9	2	2.7409	2.6486	2.5563	2.0783	3.7064	4	6
65	0.9	3	2.7408	2.6486	2.5557	2.0424	3.7805	4	4
65	1.2	1	2.7412	2.6491	2.557	2.0449	3.6362	3	4
65	1.2	2	2.7404	2.6473	2.5537	2.0879	3.7963	3	4
65	1.2	3	2.7401	2.6464	2.5514	2.1454	3.7031	4	4
80	0.3	1	2.7432	2.6485	2.5629	2.0314	4.2768	6	2
80	0.3	2	2.7432	2.6484	2.5626	2.0421	3.7287	5	6
80	0.3	3	2.7433	2.6485	2.5626	2.0281	3.8034	3	4
80	0.6	1	2.7433	2.6485	2.5626	2.033	3.7119	3	2
80	0.6	2	2.7433	2.6486	2.5622	1.9882	4.0857	6	2
80	0.6	3	2.7429	2.6478	2.5609	2.0248	3.7552	4	4

Table 5.2: MSE for AR, ARX, ARX $\log[T]$, ARMAX and ANN, and the best number of hidden layers for ANN system in function of each combination of w, a and k, for periods of 65 and 80 weeks.

alterates the previous characterization of a straight line, which is more likely to be observed in the beginning part of the exponential.

5.2.1.2 Microgeneration Running More than 2 Years

At this point, we continue the characterization of model performances, enlarging the time of operation t. Although there exist infinite possible scenarios for combination of the parameters, we settle some diverse values for k in order to assess different combinations. Values assumed for k are now k = 2.5, k = 3.5 and k = 4.5, against k = 1, k = 2 and k = 3 from the previous tables.

Table 5.3 comprises values for MSE in function of t = 105 and t = 120, whereas table 5.4 brings forth the MSE for t = 135 and t = 150. Therefore, both tables cover periods that start in 2002, which is the third year before the end of process (December 2004).

Higher values for k do not perturb the performance of Auto-Regressive models as AR, ARX and ARMAX, neither do the enlargement of time from w = 35 or w = 50 weeks to w = 150.

We can still show another Table which handles the same input data as Tables 5.3 and 5.4, in order to confirm the results that are comprised in them. For each set of w, a and k, the number of the best MSE per model was registered. For instance, adopting w = 120, a = 0.9 and k = 3.5, AR was the best fit in 9 combinations week/PS; ARX was the winner in 7 combinations, whereas ARX log[T] was the best in 10 occasions; ARMAX won in 36 combinations of week/PS, and ANN was the best in 10 cases. The number of combinations week/PS in which one method achieves the best approximation is expressed in average, according to all input values that observe the criterion of the left column. Therefore, according to Table 5.5, for all scenarios which undergo w = 135, for instance, ARMAX offers the best fit in 31.42 cases in average, and so on.

At least in what concerns to k, no great tendencies are observed, and k might be discarded as a possible influencer in the competition among models. However, as aincreases, ANN seems to improve in relation to the other methods, like a stakeholder who shares an increasing market portion. The reason underlying this apparent trend might be the aptitude of ANN system to detect non-linear sequences better than ARMA models [53]. Since we are modifying the series in its middle with a very ascendant exponential curve, we break the tendency.

w	a	k			MSE			ANN: best	
			AR	ARX	ARX_{log}	ARMAX	ANN	1st	2nd
105	0.3	2.5	2.7467	2.6605	2.5661	2.1800	3.9443	4	6
105	0.3	3.5	2.7467	2.6603	2.5658	2.1709	4.0312	6	2
105	0.3	4.5	2.7466	2.6603	2.5655	2.1544	3.8445	2	6
105	0.6	2.5	2.7458	2.6591	2.5632	2.1595	3.7429	3	2
105	0.6	3.5	2.7455	2.6589	2.5619	2.1422	6.2852	5	4
105	0.6	4.5	2.7477	2.6610	2.5635	2.1780	3.9487	6	2
105	0.9	2.5	2.7432	2.6555	2.5569	2.2071	4.0244	5	4
105	0.9	3.5	2.7490	2.6609	2.5597	2.1894	7.4646	3	3
105	0.9	4.5	2.7478	2.6572	2.5539	2.1674	4.1488	4	4
105	1.2	2.5	2.7455	2.6557	2.5505	2.1148	4.6382	5	3
105	1.2	3.5	2.7346	2.6420	2.5321	2.1478	3.7890	3	2
105	1.2	4.5	2.7302	2.6379	2.5253	2.1059	3.9687	4	5
120	0.3	2.5	2.7427	2.6536	2.5673	2.1714	3.9408	5	4
120	0.3	3.5	2.7427	2.6537	2.5671	2.1641	4.1164	6	2
120	0.3	4.5	2.7426	2.6535	2.5669	2.1575	4.6348	3	3
120	0.6	2.5	2.7431	2.6536	2.5654	2.1785	4.3511	3	3
120	0.6	3.5	2.7434	2.6537	2.5645	2.1808	3.7543	4	4
120	0.6	4.5	2.7426	2.6532	2.5629	2.2001	4.3872	6	2
120	0.9	2.5	2.7405	2.6490	2.5570	2.2047	4.7171	3	3
120	0.9	3.5	2.7449	2.6524	2.5566	2.1740	4.4021	6	2
120	0.9	4.5	2.7388	2.6453	2.5488	2.1698	3.8399	6	2
120	1.2	2.5	2.7347	2.6372	2.5370	2.1389	4.3162	3	3
120	1.2	3.5	2.7503	2.6492	2.5415	2.1441	3.6337	5	4
120	1.2	4.5	2.7318	2.6374	2.5265	2.1071	4.1117	5	4

Table 5.3: MSE for AR, ARX, ARX $\log[T]$, ARMAX and ANN, and the best number of hidden layers for ANN system in function of each combination of w, a and k, for periods of 105 and 120 weeks.

w	a	k			MSE			ANN: best	
			AR	ARX	ARX_{log}	ARMAX	ANN	1st	2nd
135	0.3	2.5	2.7395	2.6420	2.5602	2.2294	3.9212	6	2
135	0.3	3.5	2.7392	2.6419	2.5598	2.2513	3.6218	5	4
135	0.3	4.5	2.7396	2.6421	2.5598	2.2319	3.8483	6	2
135	0.6	2.5	2.7386	2.6413	2.5572	2.2570	4.1820	6	2
135	0.6	3.5	2.7383	2.6410	2.5558	2.2665	4.2890	4	4
135	0.6	4.5	2.7390	2.6407	2.5541	2.2333	4.3708	5	3
135	0.9	2.5	2.7422	2.6398	2.5504	2.1660	18.331	4	4
135	0.9	3.5	2.7373	2.6370	2.5425	2.2028	5.7739	2	6
135	0.9	4.5	2.7455	2.6400	2.5454	2.1787	4.3471	4	2
135	1.2	2.5	2.7033	2.5959	2.4927	2.1318	3.7499	3	2
135	1.2	3.5	2.7302	2.6314	2.5202	2.3369	4.9214	4	2
135	1.2	4.5	2.6764	2.5615	2.4500	2.0434	3.5608	3	2
150	0.3	2.5	2.7430	2.6447	2.5632	2.3016	3.9518	6	2
150	0.3	3.5	2.7433	2.6450	2.5632	2.2988	3.8121	6	5
150	0.3	4.5	2.7428	2.6442	2.5623	2.3168	3.8483	6	2
150	0.6	2.5	2.7448	2.6458	2.5613	2.2833	4.0309	6	2
150	0.6	3.5	2.7455	2.6462	2.5603	2.2204	3.9102	6	2
150	0.6	4.5	2.7463	2.6466	2.5593	2.2066	3.8721	4	2
150	0.9	2.5	2.7379	2.6365	2.5445	2.2835	4.1765	6	2
150	0.9	3.5	2.7552	2.6512	2.5538	2.2943	4.3579	4	2
150	0.9	4.5	2.7345	2.6306	2.5317	2.2144	3.4395	4	6
150	1.2	2.5	2.7288	2.6254	2.5148	2.3137	4.2285	4	2
150	1.2	3.5	2.7268	2.6128	2.5047	2.1735	4.0368	4	2

Table 5.4: MSE for AR, ARX, ARX $\log[T]$, ARMAX and ANN, and the best number of hidden layers for ANN system in function of each combination of w, a and k, for periods of 135 and 150 weeks.

Criterion	Best	Approximation	Average	Occurrences	
	AR	ARX	ARX_{log}	ARMAX	ANN
w = 105	11.0000	7.0000	9.6667	31.4167	12.9167
w = 120	9.91670	6.7500	9.0833	31.5833	14.6667
w = 135	12.5000	9.5000	8.1667	29.1667	12.6667
w = 150	12.3636	9.7273	8.8182	27.4545	13.6364
a = 0.3	11.2500	8.9167	9.6667	29.5000	12.6667
a = 0.6	11.0833	8.4167	8.5000	30.9167	13.0833
a = 0.9	11.6667	7.8333	9.0833	29.6667	13.7500
a = 1.2	11.7273	7.6364	8.4545	29.7273	14.4545
k = 2.5	11.6875	8.3125	8.5625	30.0625	13.3750
k = 3.5	11.1875	8.2500	9.7500	29.0000	13.8125
k = 4.5	11.4000	8.0667	8.4667	30.8667	13.2000

Table 5.5: Number of cases in which a model achieves the best approximation, taking into account all cases that obbey to the criterion on the left column.

Interestingly, when w is enhanced, ARMAX seems to lose accuracy. Nevertheless, in its worst performance, which is 27.4545 combinations week/PS, for w = 150, ARMAX overcomes ANN in two times. But this may arise the question: if we enlarged our sample sufficiently, would ANN overcome the remaining models?

If we just raise the value of a, as we were doing, probably ANN could become a winner in a given instant, but even that is not so determinable. However, Fig. 4.5 remembers us that, even when two methods have the same number of greatest successes, one can overcome the other one with a large advantage, whether in a direct confront. We remind, from Fig. 4.5, that despite ANN and ARX log[T] have offered the same number of best predictions, if we carry out a match between ARX log[T] and ANN, we obtain ARX log[T] to prevail in 78% of the total matches.

5.2.2 Solar PV Panels Generating Only Indoor

Since in Fig. 4.7 we had illustrated the complete set of forecasts for Power Substation D, now we assess, over the same PS, how the prediction behaves as we adopt only indoor light as the PV panel source. The analysis will run upon Power Substation D only for the sake of a fast comparison among the diverse methods, in face of different initial quantities c for the sources and weekly increasing rates e^a .

Firstly, we will consider w = 104, a = 2, initial units c = 50 for all sources, and k = 2.5. Only indoor light will apply now. The result is depicted in Fig. 5.8.



Figure 5.8: New version for Fig. 4.6, now adopting w = 104, a = 2.0, k = 2.5 and c = 50. Realistic Data corresponds to Measured Data from Fig. 4.6.

AR curve has not changed its relative position to the output data. It is a very important aspect since it confirms the series assembling: only components from past outputs are present, and therefore they must accord only to y(t). ARX and ARX log[T] modify a little their position in terms of offset, which is explainable due to the eXogeneous input: ARX models are allowed to vary according other parameter which is different of y(t). ARMAX model, which has still the Moving Average component, is now in a different position, fairly adapted to the new position of y(t).

On the other hand, ANN reacts completely different, assuming rather a diverse shape from that one of Fig. 5.6. The non-linearity of ANN black-box stands for its apparently incomprehensible behavior.

Now setting w = 104, a = 1, initial units c = 80 for all sources, and k = 2.5, we try to perceive if is there any consequence by lowering a and lifting c. The result is depicted in Fig. 5.9. Only indoor light continues to apply.

In the example of Fig. 5.9, ARMAX has presented the best fit in 34 cases, against 16 from ANN and 10 from AR; ARX has achieved 7 best performances and ARX log[T] reached just 5 best shots.



Figure 5.9: New version for Fig. 4.6, now adopting w = 104, a = 2.0, k = 2.5 and c = 50. Realistic data corresponds to measured data from Fig. 4.7.

Despite we have enhanced the initial value c, the final contribution from sources to the load has diminished. We remark that even an a = 1.0 is a very optimistic hypothesis. For seeing how market would actually behave, we fix a = 0.3 and set c = 20, that is to say, we assume that the first source to run power will start with 20 m² of PV panels, a reasonably interesting number.

5.2.3 Generating Only with Eels and Antennas

Beside the original realistic data, three curves are depicted: curves of load without microgeneration, then the curve of load which is diminished by using the three sources, and then load with only eels and antennas. Fig. 5.10 condenses such results.



Figure 5.10: Curves of demand with no sources generating, with the three sources and with only eels and RF antennas.

The difference between results without sources and with only antennas and eels are very insignificant for the last 9 weeks. However, we should also test how would their contribution be like in the end of 2006, and try to assess if they are also irrelevant in the mean-term. Fig. 5.11 illustrates the tiny gap that antennas provide after 4 years of participation, which barely overcomes 1 %. The initial quantity, in January 2002, is 300 antennas, and the increasing rate is 1718% for year. A huge factor a of annual expansion and a very high quantity at the beginning of the process c were not enough to make antennas to become minimally feasible as a recycling energy source.



Figure 5.11: Antennas generating from January 2002, initial quantity of 300 antennas, increasing rate 172% for year, providing the tiny gap which is observed only in 2006 months.

With regards to the eels, Fig. 5.12 illustrates what happens if we choose a = 0.8 and vary the initial quantity c, making c = 50 and c = 100, we notice that a gap is observed, despite it is still very narrow. However, dealing with eels, we needed to resort to a smaller initial amount of generator units. Moreover, the power availed by the eel may be enhanced with further researches, since we have determined the lowest performance in kWh. These points work towards not discarding eels as a feasible source yet. In some cases, electric eels might show themselves as an advantageous source, but there is still a plenty of necessary tests in order to probe their real energetic profit. The curve

c = 0 simulates the absence of eels in generation. Comparing to c = 0, c = 100 reduces the final power demand in 1.44 %.





Figure 5.12: Eels generating from January 2002, initial quantity of 50 and 100 eels, increasing rate 123% for year, providing the tiny gap which is observed in the end of 2006.

One conclusion is that the probability of success without solar panels is remote. Adoption of solar panels has high importance for the emergent generators to be seen by the Power Substation, even for indoor recycling only.

5.3 Spatial Forecast

The following analysis consisted on assigning different colors for the regions of Leipzig according to their reduction on peak load in the future. Microgenerators were considered to start to produce energy from the beginning of 2002 on, so that the maximum value for w is $w_{max} = 156$. Simulations consisted on varying c and a for the sources in order to obtain different results for the last weeks of 2005 and 2006.

In order to predict values, we made use of ARMAX model, with $[n_a \ n_b \ n_c \ n_k] = [10 \ 10 \ 10 \ 1]$. Exogeneous parameter was logarithmic of the minimum weekly temperature values, for the entire period between 2001 and 2006.

Fig. 5.13 illustrates the code which assigns different colors to some reduction percent values. These colors will be used to fill the 8 regions of Leipzig map in which the 8 Power Substations lie. Therefore, a certain combination of c and a will result in a peak reduction which is classified in a visual way. The resulting peak load apply only to the 5 weekdays; hence, Saturday and Sunday are not included. One can notice that whether a district is filled in black, for instance, it means that the respective region bears a PS whose peak load has been totally attended by the microgenerators. In this situation, more than 100 % of the peak load is supplied by the emergent sources between Monday and Friday, and the district might therefore export energy during the peak.

Weel	kly Peak Load Relief = PLR
	PLR < 10 %
	10 % ≤ PLR < 30 %
	30 % ≤ PLR < 60 %
	60 % ≤ PLR < 100 %
	100 % ≤ PLR

Figure 5.13: Color codes for the weekly peak load relief percentages.

Let us establish c_a and a_a for antennas, c_i and a_i for PV panels indoor, c_o and a_o for PV panels outdoor, and c_e and a_e for the eels, where c is the initial amount of the source and a is the logarithm of annual increase for it. The starting point for some examples considers the following set of values: $c_a = 50$, $a_a = 0.3$; $c_i = 20$, $a_i = 0.1$; $c_o = 10$, $a_o = 0.1$; and $c_e = 100$, $a_e = 0.2$. Fig. 5.14 depicts the result. We notice that only regions of Substations C and H have a relief in peak load corresponding to less than 10 %. All the remaining regions lowered their peak consumptions in the range 10 % to 30 %. On the left part of each Figure we show the table of results and inputs, and below the colors code is again depicted, as weel as the year corresponding to the week which is predicted. Districts and Substations are understood to be related to the same letter.

Setting all a = 0.2, we have now that all regions have peak allowances higher than 10 %, however none of them bearing economy of more than 30 % in the peak. Fig. 5.15 illustrates it. In the Fig. 5.15 and thereafter, figures also contain a row-table which states some examples of the annual increasing rates depending on the value of a.

Deals Dallaf	Source	Ante	ennas	Ind	oor	Out	door	Eels	
Peak Keller	Substation	С	а	С	а	С	а	с	а
19,49%	Α	50	0,3	20	0,1	10	0,1	100	0,2
12,22%	В	50	0,3	20	0,1	10	0,1	100	0,2
7,33%	С	50	0,3	20	0,1	10	0,1	100	0,2
18,46%	D	50	0,3	20	0,1	10	0,1	100	0,2
19,90%	E	50	0,3	20	0,1	10	0,1	100	0,2
10,50%	F	50	0,3	20	0,1	10	0,1	100	0,2
10,71%	G	50	0,3	20	0,1	10	0,1	100	0,2
8,20%	H	50	0,3	20	0,1	10	0,1	100	0,2
Year	2005		P	ak Load	Relief - F	ercentag	es		
			10%	30%	60%	100%	more		



Figure 5.14: Simulation with $c_a = 50$, $a_a = 0.3$; $c_i = 20$, $a_i = 0.1$; $c_o = 10$, $a_o = 0.1$; and $c_e = 100$, $a_e = 0.2$, for the last week of 2005.



Figure 5.15: New simulation for scenario of Fig. 5.3, now setting all a = 0.2.

If the scenario of Fig. 5.15 endures until the end of 2006, the reduction of load along the peak overcomes 30 % in districts where Substations A, D and E lie, as in Fig. 5.16:





Figure 5.16: Scenario of Fig. 5.3, this time for the last week of 2006.

Increasing the exponent of all sources to 0.3 and maintaining the analysis over the last week of 2006, we can perceive the impact of a little enlargement of the exponent at Fig. 5.17. Districts B, G and F reduce their weekdays peak load betwen 10 % and 30 %, whereas Substation A enhances their relief during the peak to more than 60 %. Actually, peak load reduction for Substation A is 61.76 %, as the table in the left part of this Figure lets us know.

Finally, elevating the initial amount only of the outdoor PV panels for $c_o = 20$, which corresponds to doubling it, the impact over diminishing peak load, as depicted in Fig. 5.18, is even more visible. Substations A, D and E achieves their complete energetic sufficiency for weekdays peaks. Substations C and H goes to the next level of peak load reduction, and so do Substations G, B and F.

The previous sequence of examples confirms some general ideas: the earlier the sources are connected to the grid and the greater is the annual increase rate, the more relevant is the impact of their energetic participation. The sequence of Figures also show that different Power Substations avail themselves of the existing emergent power in different





Figure 5.17: Peak load reductions for all Substations for the last week of 2006 and all a = 0.3.



Figure 5.18: Scenario of Fig. 5.17, now considering the initial amount of outdoor panels being $c_o = 20$.

levels, as they present different average power demands. Finally, the increase of outdoor panels area is much more relevant than any other source expansion.

There is a very plenty of possibilities in such test results, taking into account these elements to vary. Therefore, Appendix C gathers many examples, in order to make richer the analysis here initiated.

Chapter 6 CONCLUSIONS

Load prediction by means of temporal and spatial analysis is one of the most challenging tasks addressed by public utilities. The assurance in sustaining load is not a trivial task. The employment of correct data for information treatment techniques is each time more needed in this realm. One factor that emphasizes this increasing necessity is the change in the load behavior, given that new regulatory dispositions allow people to produce their own energy. Such self energy generation occurs not only in Brazil, but in many countries. The question arises: how does one accomplishes the best planning method for load behavior prediction, and within an ever changing environment?

Along the present study, we aimed to offer an objective answer making use of prediction techniques. In order to undertake any feasible analysis, reliable load data are the first step. In Chapter 4 we have faced the problem of determining future values of maximum power demands, with the best precision as possible, based on such data. We have employed AR models in counterpoint with the ANN black-box. At prior, the supposition was that ANN would not prevail since the set of mathematical resources at disposal of AR models seems to be wider. Our first guess at this point was demonstrated to be correct. ARMAX models attended the most of predicitons with the smallest errors, and often the total set of best shots had a half of results coming from ARMAX.

It was also a recompense to find out that logarithmic temperature values yielded a fair estimator of power demands. The weather question has strongly inspired us to consider this parameter, but evidently other parameters should be investigated, as population growth or GDP rates. It is true that most of these parameters were not available, but we have produced an experience that proves exogeneous inputs to be welcome in order to enhance explanation accuracy.

At the beginning of the process, the inclusion of three renewable sources - PV panels for recycling, RF antennas and electric eels - made the work risky, once the real impact of their generation was unknown. However, testing these different sources was important, even because the assessment of traditional sources rises no longer very innovation. New approaches are needed, and we should to seek for these new solutions whether considering the little energy consumer, which is very sensitive to environmental issues and therefore might adopt nonstandard solutions. Moreover, we tried to demonstrate that several other forms of energy generation are likely to come up all over the cities. Future market dispositions must count on this possibility. For instance, electric eels to be considered as novel source is something that might cause some strangeness in the first moment, but solutions like this might appear in the future.

The recycling objectives concerning to RF antennas and indoor light are worth it for two reasons. The first one, in order to probe the real generation power from such tiny sources. But it is also important as a way of showing people how difficult generating energy steadily and assuredly can be. Protests against large hydroelectric plants, for instance, would be undermined if people knew exactly how electric markets work, or how hard is to make 1 kW to arrive satisfactorily in our houses and buildings. We do not take for grant that such day will arrive, but the spread of little sources, as the three tiny sources exploited along this dissertation, could help people to win the lack of information and to become more sensitive to infrastructure problems - that is the best hypothesis.

The insertion of such renewable emergent sources leads us back to the grid system prediction discussion. We faced to estimate if future power demands would "feel" the presence of the three studied emergent sources. Indirectly, it would show us whether these sources are likely to be worth it. We have concluded that the insertion of solar PV panels are perceived by the system undoubtedly if these panels convert energy from outdoor. Depending on the amount of initial panels and on the increasing rythm of adoption by people, these panels can be a relevant source even when they are used solely for harvesting energy from indoor environments.

Eletric eels would need to be hugely employed if we wanted them to furnish the minimum amount of energy to be seen by the grid. The generation by means of RF recycling antennas is certainly not a feasible option, i. e., we can discard this third way of generating power.

The insertion of different loads under diverse scenarios of Chapter 5 did not modify the efficiency ranking observed in pure load forecasting of Chapter 4. ARMAX proved to be the most accurate model, and in the second position we counted on either AR or ARX log[T], both disputing this second position, with a little advantage for ARX log[T]. Again, the consideration of logarithmic values for temperature came up as a good decision, since the performances of ARX fell short of the ones from ARX log[T]. All over this study, ANN system figured almost always as a sub-optimal forecasting model. It was a meaningful perception because it was consistently observed in both Chapters 4 and 5. Therefore, it allows us to conclude that, in the presence of Box & Jenkins models, ANN is not likely to provide the best approximations. One might say that 209 serial data are not a very large amount of information, and expanding it would lead us to different results. However, many different combinations of input parameters were undertaken in the present work. Only on the boundary of situations, as elevating disproportionately one of the parameters, we beheld the improvement of ANN performance. Even in such cases, ANN was not the best one.

We conclude that ARMAX is the best model for predicting load evolution, considering or not energy furnished from power sources, while such sources consider renewable solutions in the many diverse range of small power values. The prevalence of ARMAX, ARX log[T] and AR models remained constant with over several situations, which demonstrates the roubstness of these systems in the realm of time series predicition.

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APPENDICES

Appendix A Stationary Processes

A.1 Discrete-Time Stochastic Processes

Random variables can be easily found in our all day such as body temperature, heart beat and bus schedule in Brasilia. For instance, we can compute the average and the variance of the heart beat of a person. In a certain period and later on we compute these values for the same person, collecting a set of data. We can find the average and variance in the first period. If we discover that values to the average and variance in the second period remain constant, it is an indication that we handle a stationary process. From this example, we can define a stationary random variable as a random variable with constant statistical properties - in this example, average and variance.

From statistics, we know that with the probability density function (PDF) of a random variable, we can determine all its moments. We denote the moments as the statistical properties, where the first moment of random variable is the mean, the second moment is the variance, the third moment is the skewness and the forth moment is the kurtosis. However, to find the PDF of a random variable in order to compute the moments would require a large number of samples. Moreover, usually stationarity is only observed in some time periods of the random variable. If the period becomes too long, the random variable becomes nonstationary.

Therefore, instead finding the PDF, or the probability mass function (PMF), whether our variable is discrete, we only seek for determining the first and second moment of a random variable, which requires much less samples than determining the PDF. Discrete variables differ from the continuous variables according to the specification over time. We refer to time t as being time, however, a series can bear data which are organized according to another physical parameter, as volume, distance, etc [14]. If the set of time instants can be enumerated, we have a variable which is discrete. In other words, the variable is discrete if and only if the observations between two arbitrary moments are countable. Diversely, in the continuous case, the random variable is defined for all value of $t \in R$.

Discrete variables are the ones which interested us along this work, given that mathematical tools which were employed, as auto-regressive and moving average models, presume the variables to be discrete. Therefore, the distribution probability function to speak of is the PMF.

A discrete-time stochastic process is a succession of events whose numerical values are not related each other by any apparent logic and, moreover, presents discrete values. Suppose you are driving a car through the streets of a large city: the succession of the last numers of the cars' plaques you see in front of you is a completely random process, which does not undergo any logical principle. That is an example of stochastic process, given that those numbers are not predictable; in the best hypothesis, you can suppose, after seeing many "5" repeatedly, that the next number will not be a "5", but, even in this case, there is still a probability you mistake your prediction.

On the other hand, deterministic variables are ruled by physical laws like montion of bodies, propagation of waves and temperature variations of a body in a controlled environment. For instance, imagine you are inside a train and that the constant speed of the train is 100 km/h. Therefore, after one hour, you know that you will 100 km far from the original point and after two hours, 200 km far from the origin, and so on. Therefore, *ceteris paribus*, there is no uncertainty on the distance computation in such example.

The successive observations may compose a time series u(n), where, for example, u(1) = 3, u(2) = 0, u(3) = 9, ..., u(n) = 5, where n is the number of registered data or the sampling period. If we sum all these values and divide it by n, we calculate the mean of the series.

$$\mu(n) = \mathbf{E}\{\mathbf{u}[\mathbf{n}]\} = \overline{\mathbf{u}}[\mathbf{n}]. \tag{A.1}$$

The expression $\lim_{N\to\infty} \frac{1}{N} \sum_{n=1}^{N} u[n]$ holds for ergodic processes, if we consider n as the number of instantaneous observations in the present moment. However, if we assign n to be the number of sampled observations, and provided that u(n) are independent and identically distributed (iid), this limit becomes equivalent to (A.1), due to the Large Numbers Law [61].

The mean in (A.1) consists on the first statistical moment of the random variable. As we have shown above, the lacking of the PMF is rather compensated by the the moments of the variable. Stationary processes handle these moments instead of the distribution probability functions, which are supposed to be inaccessible. Retaking the first example, the cars' plaques into the city which are observed by one driver constitutes one process. If 50 drivers take their cars throughout the city, each one of them will register a different sequence of results. In this sense, each driver perception embodies a stochastic process separately. So, considering that each driver sees the second car in front of them, there will exist other 50 results, corresponding to t = 2. If the mean from these 50 values remain the same over all the remaining observations, there is an evidence that we have come across a *stationary process*. Also, each instant t is a stochastic process, given that X(t) is, for each t, a sequence of observations itself.

A process is said to be stationary if their properties are not dependent on the instant of time the series is observed [54]. In other words, if we look at a stationary series, the mean and variance will not vary, no matter the period of time the analysis upon data is done. A *wide-sense* stationarity is achieved when the following statements are true:

$$E\{X[t]\} = \mu(t) = \mu, \forall t \in T;$$
(A.2)

$$Cov\{X[t_1], X[t_2]\} = Cov\{X[t_1 + \tau], X[t_2] + \tau], \forall \tau \in \mathbb{R};$$
(A.3)

Probabilistic laws that actuate upon a stationary series are independent on when we observe their values, i. e., such laws do not change over time. It is a useful property specially when we must to treat an infinite or very large amount of data, once that we get rid of knowing them historically from their beginning. Hence, we turn to obtain stationary series from the realistic data we face, since (A.2) and (A.3) hold for stationary processes.

Given that we are going to analyze the series nature and behavior, it is worth it remind some concepts and principles underlying mean, variance, autocovariance and autocorrelation, covariance, cross-covariance, correlation, cross-correlation and convolution. The following analysis is split between functions with only one variable, and functions for two or more variables.

A.1.1 Statistical Functions with Respect to Only One Random Variable

For all concepts hereafter exposed, a random variable X must be considered, such that $\{X(t), t \in T\}$ is a real valued discrete stochastic process.

A.1.1.1 Mean

Mean value function (or expected value) is a well known concept: the sum of a set of observations, divided by the number of these registers, provide us the mean of this observed set, which is almost always a subset of a greater group. The mean of a random vector $\mathbf{X} = [X_1 \ X_2 \ \dots \ X_n]^T$ is defined as

$$E\{[X]\} = [E\{X_1\} \ E\{X_2\} \ \dots \ E\{X_n\}]^T$$
 (A.4)

The mean of a constant is the constant itself (A.5). Multiplying a random variable by a constant α makes its mean be multiplyied by the same constant α (A.6) [55].

$$\mathbf{E}\{\alpha\} = \alpha \tag{A.5}$$

$$\mathbf{E}\{\alpha \mathbf{X}\} = \alpha \mathbf{E}\{\mathbf{X}\} \tag{A.6}$$

Supposing X(j) and $X(j + \tau)$ to be independent, the respective marginal PMFs are equal to the joint PMF $P\{X_j, X_{j+\tau}\}$. In order to build covariance matrix, correlation matrix and their respective "cross" versions, we introduce the *deviation matrix*, which is given by

$$\mathbf{X} - \mathbf{E}\{\mathbf{X}\} = \mathbf{d}_{j} = \begin{bmatrix} X_{1} - \mathbf{E}\{X_{1}\} \\ X_{2} - \mathbf{E}\{X_{2}\} \\ \dots \\ X_{n} - \mathbf{E}\{X_{n}\} \end{bmatrix},$$
(A.7)

which establishes an instantaneous value for the deviation vector on time; as in (A.7), this vector is a list of the differences of each dimension of X_i to its mean. Although it is a very frequent definition, the bibliography in many times uses another idea of deviation vector, which involves successive values of deviation over time, taking an unique dimension i of X into account:

$$\mathbf{d}_{i} = [X_{i,1} - E\{X_{i}\} \quad X_{i,2} - E\{X_{i}\} \quad \dots \quad X_{i,j} - E\{X_{i}\}], \quad (A.8)$$

where j is the occurrence in time of a random vector X, and i indicates the dimension of the j-th measured value of X. This corresponds to say that if we take j snapshots for a n-dimensioned vector X, all the values $[X_j - E\{X\}]$ for a specific dimension n are used to calculate the variance.

Diversely, in (A.7), we take j snapshots for an *i*-dimensioned vector X, and calculating variance of X_i demands us to consider all values $X_{i,j} - \mathbb{E}\{X_i\}$ for each snapshot j. Then, variance of X results in a column matrix assembled with all the elements $\mathbb{E}\{X_{i,j} - \mathbb{E}\{X_i\}\}$. In this way, every dimension of X has its specific variance.

A.1.1.2 Variance

Variance function is a measure related to how data belonging to a series are disperse with regards to the mean of the series. The expression of variance is given by

$$Var{X} = \sigma_X^2 = E{(X - E{X})^2}$$
(A.9)

Eq. (A.9) is frequently written as $Var{X} = E{X^2} - (E{X})^2$, which is an equivalent expression. Variance is also written as $Var{X} = \sigma_X^2$.

The variance of a constant is zero, as shown in (A.10). Multiplying a random variable by a constant, its variance will be multiplied by the square of this constant (A.11). Summing (or subtracting) a constant with the random variable, its variance does not change (A.12) [55, 56].

$$\operatorname{Var}\{\alpha\} = 0 \tag{A.10}$$

$$\operatorname{Var}\{\alpha \mathbf{X}\} = \alpha^2 \operatorname{Var}\{\mathbf{X}\} \tag{A.11}$$

$$\operatorname{Var}\{\alpha + \mathbf{X}\} = \operatorname{Var}\{\mathbf{X}\} \tag{A.12}$$

In addition, it is true that $Var{X} \ge 0$. From (A.11) and (A.12), we can state that $Var{\alpha X + \beta} = \alpha^2 X$, throwing some light upon the nonlinear nature of variance.

A.1.1.3 Autocorrelation Function

Say the random variable X has a known probability mass function given by f. The autocorrelation function includes the operator " \star " and is given by [57]

$$f \star f = \sum_{m=-\infty}^{\infty} f^*(m) f(m+n) \tag{A.13}$$

Autocorrelation function provides X to "pass over" itself, such in an autoconvolution. Actually, it is an autocorrelation, in the sense it is more likely to seem a cross-correlation function rather than convolution, as we will show in the next section. Fig. A.1 provides a visual comparison among the three functions here mentioned.

A.1.1.4 Correlation Matrix - correlating X(i) with X(i+k), for different k

The deviation matrix d_i from Eq. (A.7) enables us to establish the relationship between covariance and correlation, which will be explored in the next Section. Let $L^2_{d_i}$ be

$$L_{d_i}^2 = d_i d_i^T = \sum_{j=1}^n (x_{i,j} - \overline{x_i})^2,$$
(A.14)

such that $x_{i,j}$ is the value of X considering *i* as the dimension selected into X_n vector and *j* as the order of the snapshot made over X_i . We also can see that Eq. (A.14) contains a term $L^2_{d_i}$ which is proportional to the variance of X_i . Therefore we can write as expression which is related to covariance, given by

$$d_i d_k^T = \sum_{j=1}^n (x_{i,j} - \overline{x_i})(x_{k,j} - \overline{x_k}), \qquad (A.15)$$

where k describes another dimension of X_n (of course, both i and k must be less or equal the dimension of X, n). If we want to know how redundant is information brought by X_i and X_k , (A.15) can be rewritten as

$$d_i d_k^T = L_{d_i} L_{d_k} \cos(\theta_{i,k}), \tag{A.16}$$

where $\theta_{i,k}$ indicates the angle formed between d_i and d_k . Using (A.14) and (A.15) in (A.16), we have

$$\cos(\theta_{i,k}) = \frac{d_i d_k^T}{L_{d_i} L_{d_k}} = \frac{\frac{\sum_{j=1}^n (x_{i,j} - \overline{x_i})(x_{k,j} - \overline{x_k})}{N}}{\sqrt{\frac{\sum_{j=1}^n (x_{i,j} - \overline{x_i})^2}{N}} \sqrt{\frac{\sum_{j=1}^n (x_{k,j} - \overline{x_k})^2}{N}}} = \cos\{X_i, X_k\} \frac{1}{\sqrt{\operatorname{Var}\{X_i\}}\sqrt{\operatorname{Var}\{X_k\}}} = c_{i,k},$$
(A.17)

in which $c_{i,k}$ represents the sample correlation coefficient, being equal to $cos(\theta_{i,k})$. If this coefficient is near 1, the redundance is direct and strong; either if $c_{i,k}$ approaches +1 or -1, it is still strong, although the signal "-" describes that when one variable increases, the other one diminishes, for instance. But if $c_{i,k}$ is zero or nearby, X_i and X_k information are uncorrelated, that is, they have no overlap between them, and the behavior of both variable are not associated each other. In this case, one process provides information the other one does not contain. Correlation coefficient between two random variables, say, a and b, for instance, is frequently denoted as ρ_{ab} . Therefore,

$$\cos(\theta_{i,k}) = \frac{\operatorname{Cov}\{X_i, X_k\}}{\sqrt{\operatorname{Var}\{X_i\}}\sqrt{\operatorname{Var}\{X_k\}}} = c_{i,k} = \rho_{X_i X_k}$$
(A.18)

If one desire to verify correlation function between values of snapshots X(j), the only necessary operation is $c(j, j+\tau)$. Since X describes a stationary process, the correlation function depends only on the lag τ between observations x(j) that are spaced by this lag:

$$c(j, j - \tau) = E\{(X(j) - E(X_j))(X(j - \tau) - E(X(j - \tau)))^*\}, k = 0, \pm 1, \pm 2, \dots, (A.19)$$

where asterisk (*) denotes complex conjugation.

The correlation between X(i) and X(i+k) is given by:

$$R_{XX}(i, i+k) = \frac{E\{X(i)X(i+k)\}}{Var\{X\}}.$$
 (A.20)

By means of (A.20), we try to establish the relationship between different existing i-dimensions the vector X(n). While (A.19) probes correlation over time, (A.20) accomplishes the correlation between two different dimensions of X. Both types of correlation are used in several analyses. Expression (A.19) is employed, for instance, if one want to know if the variable of insterest has any seasonality or repetition; even whether temporarily. On the other hand, (A.20) is more suitable if one intend to see if X is over-dimensioned, since a positive answer indicates that two dimensions contain the same information.

A.1.2 Statistical Functions with Respect to More than One Random Variable

Two random variables X and Y must be considered now, being $\{X(t), Y(t), t \in T\}$ two real valued discrete stochastic processes. More than two variables might be considered, however for the sake of simplicity we will include only X and Y in the following analysis.

Still talking about mean and variance, there are three interesting properties which have not been yet exploited. Say we have a joint distribution of the above-mentioned two random variables. The mean of their sum or difference equals the sum or difference of their means (A.21), and the mean of the product of two independent random variables is the product of their means (A.22):

$$E\{X + Y\} = \sum_{i} \sum_{j} (x_{i} + y_{j})p\{x_{i}, y_{j}\} =$$
$$= \sum_{i} \sum_{j} x_{i}p\{x_{i}, y_{j}\} + \sum_{i} \sum_{j} y_{j}p\{x_{i}, y_{j}\} = \sum_{i} x_{i}p\{x_{i}\} + \sum_{j} y_{j}p\{y_{j}\} =$$
$$= E\{X\} + E\{Y\}$$
(A.21)

$$E\{XY\} = \sum_{i} \sum_{j} x_{i} y_{j} p\{x_{i}, y_{j}\} = \sum_{i} \sum_{j} x_{i} y_{j} p\{x_{i}\} p\{y_{j}\} =$$
$$= \sum_{i} x_{i} p\{x_{i}\} \sum_{j} y_{j} p\{y_{j}\} = E\{X\} E\{Y\}$$
(A.22)

The variance of the sum of two independent random variables is equal to the sum of their respective variances (A.23):

$$\operatorname{Var}\{X+Y\} = E\{[(X+Y) - (\mu_X + \mu_Y)]^2\} = E\{[(X-\mu_X) + (Y-\mu_Y)]^2\} = E\{[(X-\mu_Y) + (Y-\mu_Y)]^2\} = E\{[$$
$$= E\{(X - \mu_X)^2\} - 2E\{(X - \mu_X)(Y - \mu_Y)\} + E\{(Y - \mu_Y)^2\} =$$
$$= Var\{X\} + Var\{Y\},$$
(A.23)

since $E\{(X - \mu_X)(Y - \mu_Y)\} = Cov\{X, Y\} = 0$, given that X and Y are independent [55, 56].

A.1.2.1 Covariance (or Auto-Covariance) Function

Covariance function between X and Y is defined as

$$Cov{X, Y} = E{(X - E{X})(Y - E{Y})}$$
(A.24)

Eq. (A.24) may be written also as $Cov{X, Y} = E{X, Y} - E{X}E{Y}$. Covariance between X and Y is also written as σ_{XY} .

Although its mathematic expression gives us a relatively simple statement, covariance function is not a immediately comprehensible concept in the real world. In few words, covariance expresses how much informations from X and Y are alike, taking into account the mean and absolute values of X and Y as relevant factors for the magnitude of covariance [58].

A pure indicator of the existent "coincidence" between information coming from both variables is rather the auto-correlation function, for the reason this function varies between -1 and 1 and brings in a directly geometric interpretation of in what degree the overall values x(t) and y(t) are in accordance or, say, contain redundant information, in the same way as in (A.18). Nevertheless, considering covariance might be advantageous in certain circumstances. For instance, in Principal Components Analysis, if the variances differ significantly each other, or if the measurement units are not compatible, the principal components of covariance matrix will be dominated by the variables which have the largest variances. This situation might be interesting given that usually few principal components keep the most of the variance from original data. Therefore, it consists on a situation wherein magnitudes of X and Y are important, or should be strengthened.

If X and Y are independent, $Cov{X, Y} = 0$. Among the properties listed below, some other are understandable if one reminds variance properties.

$$\operatorname{Cov}\{\alpha + \beta X, \gamma + \kappa Y\} = \beta \kappa \operatorname{Cov}\{X, Y\}(A.25)$$
$$\operatorname{Cov}\{X, Y\} = \frac{\operatorname{Var}\{X + Y\} - \operatorname{Var}\{A\} - \operatorname{Var}\{B\}}{2}$$
(A.26)

$$\operatorname{Cov}\{X+Y,Z\} = \operatorname{Cov}\{X,Z\} + \operatorname{Cov}\{Y,Z\}$$
(A.27)

$$Cov{X, X} = Var{X}$$
(A.28)

$$\operatorname{Cov}\{\alpha, \mathbf{X}\} = 0 \tag{A.29}$$

One may use $\text{Cov}\{X(t), X(t + \tau)\} = E\{(X(t) - E\{X(t)\})(X(t + \tau) - E\{X(t + \tau)\})\}$ to probe if X is stationary: whether $\text{Cov}\{X(t), X(t + \tau)\} \neq 0$, at least within the observed period between t and $t + \tau$, X bears dependence in time; X can, alternatively, be a cicle-stationary random variable if, keeping τ constant, it continues to happen in all t moments.

In matricial terms, (A.24) is now used to assemble covariance matrix of vector \mathbf{X} , which is here expressed as Σ_X :

$$\Sigma_{X} = \mathbf{E} \begin{bmatrix} X_{1} - \mathbf{E}\{\mathbf{X}_{1}\} \\ X_{2} - \mathbf{E}\{\mathbf{X}_{2}\} \\ \vdots \\ X_{n} - \mathbf{E}\{\mathbf{X}_{n}\} \end{bmatrix} \mathbf{E} \begin{bmatrix} X_{1} - \mathbf{E}\{\mathbf{X}_{1}\} & X_{2} - \mathbf{E}\{\mathbf{X}_{2}\} & \dots & X_{n} - \mathbf{E}\{\mathbf{X}_{n}\} \end{bmatrix}$$
(A.30)

Given that $Cov{X, Y} = 0$ when X and Y are uncorrelated, we have $E{X}E{Y} = E{XY}$, and thus it is correct to state that

$$\Sigma_{X} = \begin{bmatrix} E\{(X_{1} - \overline{X_{1}})(X_{1} - \overline{X_{1}})\} & E\{(X_{1} - \overline{X_{1}})(X_{2} - \overline{X_{2}})\} & \dots & E\{(X_{1} - \overline{X_{1}})(X_{n} - \overline{X_{n}})\} \\ E\{(X_{2} - \overline{X_{2}})(X_{1} - \overline{X_{1}})\} & E\{(X_{2} - \overline{X_{2}})(X_{2} - \overline{X_{2}})\} & \dots & E\{(X_{2} - \overline{X_{2}})(X_{n} - \overline{X_{n}})\} \\ \vdots & \vdots & \ddots & \vdots \\ E\{(X_{n} - \overline{X_{n}})(X_{1} - \overline{X_{1}})\} & E\{(X_{n} - \overline{X_{n}})(X_{2} - \overline{X_{2}})\} & \dots & E\{(X_{n} - \overline{X_{n}})(X_{n} - \overline{X_{n}})\} \\ \end{array}$$

$$(A.31)$$

where $\overline{X_i} = E\{X_i\}$. It is equivalent to

$$\Sigma_X = \begin{bmatrix} \sigma_1^2 & \sigma_{12} & \dots & \sigma_{1n} \\ \sigma_{21} & \sigma_2^2 & \dots & \sigma_{2n} \\ \vdots & \vdots & \ddots & \vdots \\ \sigma_{n1} & \sigma_{n2} & \dots & \sigma_n^2 \end{bmatrix}$$
(A.32)

According to (A.32), covariance matrix has, in its main diagonal, the variances of X_i and, for $i \neq j$, the a_{ij} elements contain the covariances between X_i and X_j .

In the case our aim is to examinate time-lag behavior, we consider X(t) and $X(t+\tau)$ as the variables of interest, and for them we can encounter the covariance matrix $r(t, t+\tau)$ which is given by

$$r_X(t,t+\tau) = Cov[X(t), X(t+\tau)] = E[X(t)X(t+\tau)] - E[X(t)]E[X(t+\tau)] \quad (A.33)$$

Covariance between different values of X in time depend only on the lag of the observations, if X delineates a stationary process. Therefore, given a certain τ , Eq. (A.33) holds for any value of τ . That is the reason why we rewrite it, obtaining

$$r_X(\tau) = \operatorname{Cov}\{X(t), X(t+\tau)\}$$
(A.34)

Covariance accepts commutation, i. e., $Cov{X, Y} = Cov{Y, X}$. As the title of this subsection let us know, covariance function is also named as *auto-covariance function*.

A.1.2.2 Cross-Covariance Function

The function

$$r_{X,Y}(t,t+\tau) = \text{Cov}\{X(t), Y(t+\tau)\} = E\{X(t)Y(t+\tau)\} - E\{X(t)\}E\{Y(t+\tau)\} \quad (A.35)$$

is called the *cross-covariance function* between X(t) and Y(t), being each variable observed in instants of time spaced by τ . Cross-covariance function is a very important tool in order to assemble the cross-covariance matrix: the elements out of the main diagonal are cross-correlation function values. Writing $r_{X,Y}(t, t+\tau) = \sigma_{XY}(\tau)$, we have

$$\Sigma_{XY}(\tau) = \begin{bmatrix} \{E\{(X_t - \overline{X_t})(X_t - \overline{X_t})\} & E\{(X_t - \overline{X_t})(Y_{t+\tau} - \overline{Y_{t+\tau}})\} \\ E\{(Y_{t+\tau} - \overline{Y_{t+\tau}})(X_t - \overline{X_t})\} & E\{(Y_{t+\tau} - \overline{Y_{t+\tau}})(Y_{t+\tau} - \overline{Y_{t+\tau}})\} \end{bmatrix}$$
(A.36)

which corresponds to

$$\Sigma_{XY}(\tau) = \begin{bmatrix} \sigma_{X(t)}^2 & \sigma_{XY}(\tau) \\ \sigma_{XY}(\tau) & \sigma_{Y(t+\tau)}^2 \end{bmatrix}$$
(A.37)

Every applicable property of covariance holds here. For that, one needs only to replace $Y(t+\tau)$ by Y in (A.24). One subtle difference between covariance and cross-covariance relies on the parameters: Cov{X(t), X(t + τ)} considers different dimensions of **X** (X₁, X₂, ..., X_n), whereas $r(\tau)$ handles two different variables delayed by τ . Hence, making $t + \tau = u$, we have

$$r_{XY}(\tau) = \mathbf{E} \begin{bmatrix} X_1(t) - \overline{X_1}(t) \\ X_2(t) - \overline{X_2}(t) \\ \vdots \\ X_n(t) - \overline{X_n}(t) \end{bmatrix} \mathbf{E} \begin{bmatrix} Y_1(u) - \overline{Y_1}(u) \end{bmatrix} \dots Y_n(u) - \overline{Y_n}(u) \end{bmatrix}$$
(A.38)

and using (A.31), we obtain the equivalent expression for the cross-correlation if all values X and Y are taken with the difference in time τ :

$$\Sigma_{XY}(\tau) = \begin{bmatrix} E\{(X_1(t) - \overline{X_1}(t))(Y_1(u) - \overline{Y_1}(u))\} & \dots & E\{(X_1(t) - \overline{X_1}(t))(Y_n(u) - \overline{Y_n}(u))\} \\ E\{(X_2(t) - \overline{X_2}(t))(Y_1(u) - \overline{Y_1}(u))\} & \dots & E\{(X_2(t) - \overline{X_2}(t))(Y_n(u) - \overline{Y_n}(u))\} \\ \vdots & \ddots & \vdots \\ E\{(X_n(t) - \overline{X_n}(t))(Y_1(u) - \overline{Y_1}(u))\} & \dots & E\{(X_n(t) - \overline{X_n}(t))(Y_n(u) - \overline{Y_n}(u))\} \\ & (A.39) \end{bmatrix}$$

A.1.2.3 Correlation and Correlation Coefficient

We have seen in (A.20) the correlation between X(i) by X(i+k). In the part of "one variable" analysis, we assumed that both i and k were different dimensions within

X(n) vector, and hence $i \leq n, k \leq n$. Now we will check the correlation (function and coefficient) between two different variables.

Correlation is simply the expected value of the product of two random variables: $E{XY}$ is the correlation between X and Y. However, there is another definition which is much more utilized: the *correlation coefficient*, corresponding to the covariance divided by the standard-deviations of each random variable, as in (A.40).

$$\rho_{xy} = \frac{\operatorname{Cov}\{\mathbf{X}, \mathbf{Y}\}}{\sqrt{\operatorname{Var}\{\mathbf{X}\}\operatorname{Var}\{\mathbf{Y}\}}} = \frac{\operatorname{E}\{(\mathbf{X} - \operatorname{E}\{\mathbf{X}\})(\mathbf{Y} - \operatorname{E}\{\mathbf{Y}\})\}}{\sigma_x \sigma_y}.$$
 (A.40)

Variables X and Y are supposed to not bear temporal lag in this case, i. e., $X(\tau)$ is tested to be correlated to $Y(\tau)$ in the same instant τ they happen. Correlation function provides more directly the indication of how two random variables (or measurements of the only random variables) vary in the same way. We say "directly" because, on the contrary of covariance, now the values are enclosed in the interval [-1, 1]. If $\rho_{xy} = 0$, we know that the values are uncorrelated, i. e., there are not any information overlap between X and Y. Expressions between (A.35) and (A.19) provide us to visualize that mean, autocorrelation and autocovariance functions are related by

$$c(n, n-k) = r(n, n-k) - \mu(n)\mu(n-k)^*$$
(A.41)

for a random variable describing a stationary process.

A.1.2.4 Cross-Correlation

Since the term "cross" informs that there is at least one other random variable in the function combined with a lag in time, say we have values of X_i and Y_i which describe two stationary processes. For cross-correlation, the lag is varied in order to promove the "passage" of the values of a function over the other function values. For discrete functions, in which we are interested, let us consider two PMFs given by f and g, respectively to X and Y. The autocorrelation function includes the operator " \star ", as in (A.42).

$$f \star g = \sum_{m=-\infty}^{\infty} f^*(m)g(m+n) \tag{A.42}$$

A.1.2.5 Convolution

Convolution comprises the concept of cross-correlation function, only modifying the signal of X: the probability density of the difference Y - X is given by the cross-correlation, whereas the sum X + Y is provided by the concolution function. It is not too much to remind that these functions apply to the random variables pmf's. Convolution is given by

$$f * g = \sum_{m=-\infty}^{\infty} f^*(m)g(n-m)$$
 (A.43)

Figure A.1 shows a visible comparison of convolution with cross-correlation function and autocorrelation.



Figure A.1: Three graphics with comparison of Convolution (a) with Cross-Correlation (b) and Auto-Correlation (c) [57].

A.2 Stationarity - Examples

A.2.1 Non-Stationary Processes - the Neighbours Musicians

Imagine your house lies on a street wherein everyone have musical gifts. This supposition might sound a little whimsical, but it makes possible an interesting example about non-stationary processes. The first four neighbours on the left side of the street play specially often. They live in the houses named A, B, C, and D.

An old man who plays violin and viola dwells in house A; in house B, there is a fat guy playing the bassoon. In house C lives a woman who plays the cello in her long spare time, and, in house D, there is a conductor, with his piano. At each minute of the daily 70 minutes they practice their instruments, the dominant tone is registered. This tone can be understood as the frequency, in Hertz (Hz), concerning to the harmonic tone that was sounded in the most within that minute. The dominant tone is, therefore, a random variable that establishes a relationship between each minute and and the corresponding number of Hertz of the tone that was executed in the most of seconds.

In house A, the man uses to start studying with the sharpest string (E) of his violin and, usually after that, he plays his viola, which looks like a violin, but is longer and has lower pitches of sound (even because it has a larger body). In the house B, the basson player is a pretty disciplined music student, being recently chosen to take part as a solist into one Mozart concerto. Therefore, he has been performing sistematically his part of the concerto, over and over.

His neighbor, the woman with the cello, is a free player, and enjoys to play by herself with no musical classes involved. Nevertheless, she follows instrumental methods. She thinks that a worthy studying routine becomes with the low tones, ending with the highest ones, and therefore she starts studying with the lower frequency levels. The inhabitant of the fourth house, the conductor, examines compositions he will conduct in the next concerts. In the 39th minute, on average, he goes to the kitchen, and his cat, which is watching him to play, steps down and uses the lowest keys on the keyboard to come down and follow his owner, bringing very low frequencies into the registers; on the 52th minute, the conductor goes out again, now to grab a book, and a second outlier comes up.

Let us say that each house embodies a stochastic process. If a random process is considered widely stationary (when we name it simply as *stationary*, we are talking about a *wide-sense stationary* process), and the joint distribution of $X(t_1), X(t_2), \ldots,$ $X(t_n)$ is equal to the joint distribution of $X(t_1 - k), X(t_2 - k), \ldots, X(t_n - k)$ for all times t_1, t_2, \ldots, t_n and for all lags k (constant), being n the number of processes into the considered distribution.

If n = 1, there is only one process into the distribution, and Z_t distribution is equal to the distribution of Z_{t-k} for all k; in other words, both Z_s are identically distributed, given that $E\{Z_t\} = E\{Z_{t-k}\}$ and $Var\{Z_t\} = Var\{Z_{t-k}\}$, for all t and k. If n = 2, the distribution comprises two processes; the joint distribution (Z_t, Z_s) is equal to that of (Z_{t-k}, Z_{s-k}) , in which we have $Cov\{Z_t, Z_s\} = Cov\{Z_{t-k}, Z_{s-k}\}$, for all t, s and k. Prior to further analysis, we look into the musicians houses as four processes of n = 1 each. Returning to the neighbours street, let us consider a period of 70 minutes which represents their daily routine. Their typical behaviors are depicted in Fig. A.2, therefore, the depicted sequences are average processes.



Figure A.2: Neighbours musicians, playing according to their own studying routines.

Which process, among the related to the four inabitants above, can be considered a stationary one? The answer is: none of them. All neighbours constitutes cases of non-stationarity phenomena, and each one for different reasons.

The old man that starts playing violin and from the 25th minute on, approximately, sounds his viola is modifying the mean of the series. When the series has a changeable average, stationarity does not apply to it, since $E\{t\} \neq E\{t+k\}$. The different offsets can be easily explained by the average difference of frequencies of the violin and viola strings. We write, from (A.5) and (A.21), $E\{\alpha + X\} = \alpha + E\{X\}$, assigning α as the gap between average tones.

The disciplined bassoon player is repeating periodically the same sequence of data. In such case, whe have the so-called *seasonality*. Such time series are also non-stationary, given that there is only one k for which $E\{t\} \neq E\{t + k\}$ holds. That is why this is named a cicle-stationary process, which consists on a stationarity that is verified only for a specific interval of observations - in this case, each 18 minutes.

Regarding now the woman on the cello, we face either a non-stationary series because there exists an observable trend, which shows increasing values. Such fact breaks stationarity down since variance is not constant, on account of the expression (A.11). Whether one desires to sweep off this trend, one makes use of the differentiation method. Moreover, the mean appears in constant modification, which disagrees with the concept of stationarity.

The conductor is the nearest one to $E\{t\} \neq E\{t+k\}$, except in what concerns to the problem of the outliers. They make variance suffer sudden oscillations, since the standard-deviation is instantaneously modified. Besides, the expected value does not remain the same all the time. If the conductor avoids discontinuing his study about the next concert, probably he would produce a stationary series. Despite of this fact, many time series like this one are treated with no previous outliers removal, and that brings no significant loss of exactitude with respect to the results.

A.2.2 Example of a Stationary Series

After visiting so many nonstationary examples, let us characterize a positive case of stationary process. One of the most curious chracteristics of π is its apparent normality. A normal number is a real number whose decimal numerals are randomly distributed, i. e., all numerals appear with the same frequency [59]. We do not know yet if π is actually normal [60], although there is a lot of speculation in this sense.

Such normality seems to be true, and in function of the several observations that are made about this number, we can build an example of stationary process. Taking the first 20000 decimal numerals, and computing the PMF of each one of the 10 digits, we yield the graph in Fig. A.3.



Figure A.3: PMF of the first 20000 decimal numerals of π .

Number 5 appears in 2082 occurrences, practically 4% above the average, whereas number 7 occurs about 2.5% below the average (coming up 1953 times). For the 100000 first numerals, the distribution becomes even flatter. This fact itself indicates a high probability of this series to be stationary. The numeral which appears in the most of times is 1 (10138 occurrences - 1.38% above) and the less to appear is 9 (9902 times - 1% below).

This fact inspires us to consider the stationarity of the numerals distribution. Since they show up randomly, the probability of appearance seems to be equal as we enlarge the sample. We cannot take it for grant, but it seems to be a strong vestige for the stationarity feature of this series.

Summing two digits in a sliding window, we compute new values for the PMF shown in Fig. A.4.



Figure A.4: PMF of the sum of each two numerals side by side, among the first 20000 decimal numerals of π .

For the sum of three and four digits, respectively, we produce the graphs in Fig. A.5. The pattern comes up slowly: what we are achieving is a gaussian curve, which characterizes stationary processes since the appearance of a given result has a probability more and more determinable.

As the sum of more neighbour numerals makes the curve more and more gaussian, the Central Limit Theorem (CLT) may apply. One might remember that the most important example of convergence in distribution [61] is the CLT, in which we consider



Figure A.5: PMFs of the sum of each three numerals (a) and four numerals (b) side by side, among the first 20000 decimal numerals of π .

random variables to be independent and identically distributed (iid). However, iid processes are strictly stationary processes. Numerals beyond the comma of π are, therefore, a probable example of stationary series.

A.2.3 Ergodicity

According to [62], the term *ergodic* results from the concatenation of two Greek words: *ergos*, which means work, and odos = way. This term was introduced in Physics by the physicist L. Boltzmann.

L. Boltzmann, J. C. Maxwell and J. C. Gibbs were interested in systems which are described by means of a Hamiltonian flow. Boltzmann believed that the typical flow orbits filled all the energy surface that contains such flows. Counting on these ergodic hypothesis, he deduced the the temporal averages of observable measurements along the typical orbits coincide with the respective spatial means in the energy surface, a crucial fact for his kinetic theory formulation.

This hypothesis is clearly false; however, in the long run, it became a habit to call as ergodic hypothesis what should be its consequence, that is to say, spatial and temporal means being equal.

From [58] we can draw a good explanation of what is an ergodic phenomenon. If X_t is a variable in a time series, or more generally in a stochastic process, the expectation is a function of time, $\mu_X(t) = E\{X_t\}$. But if the process is stationary, strictly or weakly, $\mu_X(t)$ is a constant, i.e., all variables have the same expectation. Therefore, a natural question is: can one replace repeated observations of X_t at a fixed t, with observations of the entire time series, and estimate the common expectation μ by the time average $\mu_n = \frac{1}{n} \sum_{t=1}^n x_t$, observed in one single realization of the process? The vector $[x_1, \ldots, x_n]$ is here one observation of $[X_1, \ldots, X_n]$.

We know, from the law of large numbers, that the average of repeated measurements converges to the ensemble average μ as $n \to \infty$. But would be it equally true for time average? Does it also converge to μ ?

The answer is yes, for some processes. These processes are called *ergodic*, or, more precisely, *linearly ergodic*. A linearly ergodic stationary sequence is a stationary process $\{X_n, n = 1, 2, ...\}$ where the common mean value μ , which is the ensemble average, can be consistently estimated by the time average

$$\frac{x_1 + x_2 + \ldots + x_n}{n} \to \mu, \tag{A.44}$$

where x_1, x_2, \ldots, x_n are observations in one single realization of $\{X_t\}$. [58]

The main underlying idea of an ergodic phenomenon is that everything that can happen in repeated experiments, also happens already in one single realization, if it is extended indefinitely. Let us remeber the example of the cars plaques given in the beginning of this Appendix. That team of 50 drivers are again with their cars throughout the city, each one of them will registering a different sequence of results. Say we select a random moment to collect the 50 observations. The mean of this of this set of values is suposed to rest nearby 4.5, since

$$4.5 = \frac{0+1+2+3+4+5+6+7+8+9}{10}.$$
 (A.45)

It is possible that this result stand not very close to 4.5; however, given that it is an ergodic process, if we enlarge the set of driver from 50 to 5000, the result will approach 4.5 with a considerable certainty. The ergodicity of this phenomenon relies on the fact that 4.5 is also the result we would get if we analyzed the outcomes from a single driver, along with several repetitions. The *ensemble mean* is equal to the *mean in time*; whenever it holds, the process is ergodic.

Appendix B Z-Transform Models

Modelling is the principle by which one can interact with a real process without facing it directly. Since the real observed processes and phenomena - mainly in nature and also considering human relations - are intrinsically nonlinear, models are developed in order to simplify information to be processed in an indirect perspective. For instance, one can cite Albert Einstein's Relativity Theory as a model that allows us to understand how do mass, space and time behave into specific conditions, truly inaccessible by any means but imagination. It is very hard to suppose how could anyone reproduce, without a hugely advanced technology and in a concrete way, the knowledge arisen from this little set of suppositions; nevertheless, such theory makes possible to build further theoretic knowledge. Without living or experimenting something by our own hands, we can still craft a group of ideas and principles that interprete realistic events, which enables us to understand and control the world that surrounds us.

The underlying idea all over this work, when we cite modelling, is stochastic analysis. Our capability of forecasting or systematizing apparently indeterminate events is the core activity into stochastic processes. Consider the ability of determining the increase in Gross Domestic Product (GDP) in the two next years. A plenty of data which are still to come up in the real world affect the GDP, nonetheless, the very majority of the forecasted future GDP lies in the confidence interval. For this reason, they are a sactisfying model, although they count on data that are seemingly apart from the reality.

Periodic data forecasts are very related to the Fourier and Laplace transforms, and the main difference between them is the attenuation term σ . Fourier transform employs only the frequency of the phasor $e^{-j\omega t}$ as parameter of integration, being bare of the this attenuation term. A signal x(t), sampled over periods of length T, has as its Fourier transform the following complex functions series [63]:

$$G(j\omega) = x_0 + x_1 e^{-j\omega T} + x_2 e^{-j\omega 2T} + x_3 e^{-j\omega 3T} + \dots$$

As said, Laplace transform introduces a little modification consisting on the real part of the exponent of e:

$$L(s) = x_0 + x_1 e^{-sT} + x_2 e^{-s2T} + x_3 e^{-s3T} + \dots$$

The e^{-sT} shift operator provides a T delay between samples, expressed by

$$e^{-sT} = z^{-1} = e^{-(\sigma - j\omega)T} = e^{-\sigma T} e^{-j\omega T} = r e^{-j\theta},$$
 (B.1)

therefore, $z = e^{sT}$ is the operator of unitary advance T.

B.1 Background: Z-transform

B.1.1 Definition

Let X be a signal whose values are registered at each period of length T. X is treated as a discrete process; even if X is a continuous signal, it is sampled in order to be possible to manage it discretely. The z-transform of a function f(x), which is expressed by means of its instantaneous values x(n), is given by

$$F(z) = \sum_{n = -\infty}^{\infty} x_n z^{-n}.$$
 (B.2)

The z-transform of a sampled signal is a series of z^{-n} terms. The z^n coefficients are the temporal sampled values $[x_n]$. Each sample is positioned in the sequence taking into account the T number of delays, which are provided by the correspondent z^n operator, where the reference is time-origin [63].

The term z, the central piece of this model, is a complex variable. Its two main properties in this transformation are length and phase.

We can see that Laplace transform is more likely to offer a good paralell to z-transform, due to the existence of an attenuation factor σ , within $s = \sigma + j\omega$. It is possible to establish a direct relationship between s and z. In fact, all points from s-plane have a correspondent point into z-plane. Let $s = \alpha + j\beta$. The corresponding s-plane point in z-plane is

$$z = e^{sT}$$
$$z = e^{(\alpha + j\beta)T} = e^{\alpha T} + e^{j\beta T},$$

therefore,

 $|z| = e^{\alpha T}$

and

(B.3) angle
$$\{z\} = \beta T$$
.

B.1.2 Region of Convergence (ROC)

The result of (B.2) must be summable, i. e., the infinite sum has to converge to a determined value. When that occurs, X(z) is said to be uniformly convergent. For any given time series x(n), the set of values of z for which X(z) achieves such convergence is referred to as the *Region of Convergence* (ROC).

In (B.3), if we set $\alpha = 0$, we obtain, in the *s*-plane, the points belonging to the vertical axis. If the poles of the Laplace transform of any function lie upon this axis, we have a critically stabilized function; whether poles are in the $\alpha < 0$ semi-plane, the function is stable. Otherwise, the stability is prejudiced. Turning back to $\alpha = 0$, the equivalent points into the *z*-transform are that corresponding to $s = j\beta$, therefore,

$$|z| = e^0 = 1$$

and

$$0 < \text{angle } \{z\} < 360^{\circ}.$$

When $\alpha > 0$, then |z| > 1, and, thus, in the z-plane we have the region outside the unitary circle. Inside this circle, the dual on s-plane is the set of points with $\alpha < 0$. The left-side of s-plane corresponds to the interior of the unitary circle of z-plane. Figure B.1 shows a summarized relationship among the points of the two planes.

This is a mathematical comparison between both planes, which still does not provide an analysis of convergence; as said, the ROC is the region of z-plane in which X(z) converges, and such convergence clearly depends on x(n) values. One first glance over (B.10) would lead us to conclude that, if |z| < 1, the necessary future for X(z)



Figure B.1: Relationship among z-plane and s-plane points.

would be infinite values, given the negative exponent of z in Eq. (B.10) and the equivalence established in Fig. B.2. However, |z| < 1 is an isolated information which does not allow us to conclude X(z) will necessarily diverge. Say, for instance, that $x(n) = a^{-n}$, $a \in R$ and $n = 1, 2, \ldots$, situation in which X(z) will converge for |z| < |a|:

$$\begin{aligned} X(z) &= \sum_{n=-\infty}^{\infty} a^{-n} z^{-n} \\ &|a^{-n} z^{-n}| < 1 \\ &|a^{-n}| < |z^n| \\ &|z| > \frac{1}{a} \end{aligned}$$

B.1.3 Behavior of z

The z-transform of the sample signal is a series based on z^{-1} terms; the coefficients are the temporal samples x_i [63]. Since z is a complex variable, each information within the time-domain process is assigned to an angle over z-plan, which comes up in function of the existent T delay-factor α on the e exponent. Therefore, the angular anti-clockwise displacement, in this plan, is the dual of time running. The length of z, as above-mentioned, is the parameter related to attenuation. Figure B.2 sketches three phasor sequences in z-plane, varying the the length of z and its exponent signal. As said, if z is unitary, only the phase will vary, with no further consequences to the



Figure B.2: Behavior of z, depending on |z| and d: (a) |z| < 1, positive exponent (clockwise); (b) |z| > 1, positive exponent; (c) |z| > 1, negative exponent (anti-clockwise).

absolute value of processed signal; this is, by the way, the case in which we have a Discrete Fourier Transform.

Fig. B.2, (a) shows an decreasing absolute value of z, which makes clockwise displacements. Whether z has a length greater than 1 and d > 0, the spiral will present an increasing value of its radius, corresponding to (b). If the length of z is less than 1, but d < 0, then the sequence of z^{-d} results in a spiral whose radius diminishes step by step, in an anti-clockwise sense (c).

B.1.4 Time-delayed data

The z-transform has great application in sampled signal filtering, with regards to its capability of delaying data over time. Consider the following system in s-plane, with a transfer function given by

$$H(s) = \frac{Y(s)}{X(s)} = \frac{1}{s},$$
$$sY(s) = X(s).$$

This expression is the same as to say, in time domain, that

$$\frac{dy(t)}{dt} = x(t).$$

Now, assume that a similar transfer function with the z-transform results in

$$Y(z) = z^{-1}X(z).$$

The last expression means that the output y(k) is equal to an input received by the filter one T_s instant ago, that is, y(k) = x(k-1). In other words, the observer on the output side of the filter needs to wait one unit of time to receive the value of input. The signal x(t) is necessarily associated to the *causal* part of the system, since y(t) will receive x(t) information only at the moment t = k + 1. That is the reason why such z-transform is called as a *delay shift operator*. Other properties are examined next.

B.1.5 *Z*-Transform Properties

In [64] there are some interesting properties in of z-transform, and the first one to view, linearity, turns z-transform into an useful tool for linear signal filters models.

B.1.5.1 Linearity

$$Z\{\alpha f_1(kT) + \beta f_2(kT)\} = \alpha Z\{f_1(kT) + \beta Z\{f_2(kT)\}$$
(B.4)

Proof:

$$Z\{\alpha f_1(kT) + \beta f_2(kT)\} = \sum_{k=0}^{\infty} \{\alpha f_1(kT) + \beta f_2(kT)\} z^{-k} =$$

= $\alpha \sum_{k=0}^{\infty} \{f_1(kT) z^{-k}\} + \beta \sum_{k=0}^{\infty} \{f_2(kT) z^{-k}\} =$
= $\alpha Z\{f_1(kT)\} + \beta Z\{f_2(kT)\}$

B.1.5.2 Time-Shifting

$$Z\{f(k+n)\} = z^{n}F(z) - z^{n}\sum_{k=0}^{n-1}f(k)z^{-k}$$
(B.5)

Proof:

$$Z\{f(k+n)\} = \sum_{k=0}^{\infty} f(k+n)z^{-k} = \sum_{l=n}^{\infty} f(l)z^{n-l} = z^n \sum_{l=n}^{\infty} f(l)z^{-l} =$$

= $z^n \{\sum_{l=0}^{\infty} f(l)z^{-l} - \sum_{l=0}^{n-1} f(l)z^{-l}\} =$
= $z^n F(z) - z^n \sum_{k=0}^{n-1} f(k)z^{-k}$

Equation (B.30) allows us to consider the compex variable Z as an unitary advance operator, provided that f(0) = 0. In other words,

$$zF(z) = Z\{f(k+1)\}$$
 (B.6)

This is the dual of the *delay shift operator* seen before.

B.1.5.4 Initial Value Theorem

If y(kT) has Y(z) as z-transform and the limit

$$\lim_{z\to\infty} Y(z),$$

then there exists

$$y(0) = \lim_{z \to \infty} Y(z) \tag{B.7}$$

Proof:

$$Y(z) = \sum_{k=0}^{\infty} y(kT) z^{-k} = y(0) + y(1) z^{-1} + y(2) z^{-2} + y(3) z^{-3} + \dots =$$
$$= y(0) + \frac{y(1)}{z} + \frac{y(2)}{z^2} + \frac{y(3)}{z^3} + \dots$$

Applying limit, we achieve $\lim_{z \to \infty} Y(z) = \lim_{z \to \infty} [y(0) + \frac{y(1)}{z} + \frac{y(2)}{z^2} + \frac{y(3)}{z^3} + \dots] = y(0)$

B.1.5.5 Final Value Theorem

If F(z) converges for |z| > 1 and if every pole of { (z-1)F(z) } are into the unitary circle, then

$$\lim_{k \to \infty} f(k) = \lim_{z \to 1} (z - 1) F(z)$$
(B.8)

B.1.5.6 Multiplication by γ^n

$$Z\{\gamma^n x[n]u[n]\} = F(\frac{z}{\gamma}) \tag{B.9}$$

B.2 Z-Filters

The models hereafter described are particular cases of linear filter models. The main supposition applicable to these models is that the temporal series has been generated from a linear filter, whereas the input is a white noise a_t , i. e., for each $t \in T$, a_t is a random variable with [54]

$$\begin{split} \mathbf{E}[a_{\mathrm{t}}] &= 0, \forall t, \\ \mathrm{Var}[a_{\mathrm{t}}] &= \sigma_{a}^{2}, \forall t, \\ \mathbf{E}[a_{\mathrm{t}}a_{\mathrm{u}}] &= 0, t \neq u. \end{split}$$

Therefore, the series can be expressed as follows:

$$Z_t = \mu + a_t + \psi_1 a_{t-1} + \psi_2 a_{t-2} + \ldots =$$

= $\mu + \psi(B) a_t$,

where μ , in general, is the parameter which determines the level of the series and

$$\psi(B) = 1 + \psi_1 B + \psi_2 B^2 + \dots$$

is the linear operator which converts a_t in Z_t , being denominated as the *transfer func*tion of the filter, as in Fig. B.3.

$$a_t \longrightarrow \psi(B) \longrightarrow Z_t$$

Figure B.3: Linear Filter of ψ_i weights.

When the series of weights ψ_1, ψ_2, \ldots is finite or convergent infinite, Z_t will be stationary with mean μ ; otherwise, Z_t is non-stationary and μ has no specific meaning.

One filter is causal if its impulse response is zero for negative time, $\phi(n) = 0$, for n < 0. Filters which operate in real time are obviously causal. The *time-invariance* property is present when, in face of an excitation v(n-k), the response of the filter is u(n-k), no matter what is the value of the time shift k. Considering these concepts, a causal and time-invariant filter is stable if and only if all of the poles of the filter's transfer function lie inside the unit circle in the z-plane. This statement, however, says nothing about the zeros of the transfer function, which can be situated everywhere in z-plane [65].

The structure of linear filters concerns basically to what is the type of interaction between input and output data:

- Autoregressive models, or AR, when just the past values of the output variable are considered as input information;
- Moving Average models, or MA, when the input data is based on white noise process values therefore, no past values of the output are taken into account;
- Autoregressive Moving Average models, the so called ARMA models, which combines the AR and MA concepts.

There are many possibilities of new models based on other aspects: in the presence of exogeneous inputs, the process will receive an "X" in its name to indicate "eXogeneous" or "eXternal" inputs, as in the ARMAX process. In this type of model, more than one exogeneous data, or auxiliary variables, can be adopted into the process. In the ARIMA model, we need to eliminate previously the non-stationarity of the input data, so that the "I" means integration of the values after processing. A large amount of processes can be generated combining many different characteristics, as in MA, AR, ARX, etc.

These possibilities may be organized using parameters of a generalized equation. One can represent [4] a large number of members of AR-MA families by means of this basic model, corresponding to Fig. B.4, which comprises two inputs, W and U, as well as their transfer functions, respectively, H(z) and G(z). The output is Y(z). Let the system be invariant on time and let X be a small perturbation related to stationary stochastic process.



Figure B.4: General System for Stochastic and Deterministic Inputs [65].

The z-transform of the output is

$$Y(z) = G(z)U(z) + H(z)W(z),$$
 (B.10)

where G(z) is the transfer function concerning to the deterministic input U and H(z) is the transfer function related to the stochastic part of the system W(z). The functions G(z) and H(z) are stable and rational. H(z) has still these properties:

- 1. $H(z)^{-1}$ is stable;
- 2. $\lim_{z\to\infty} H(z) = 1.$

It is admissible to suppose that the stochastic and the deterministic parts have some poles in common. From that, we can rewrite (B.10) in order to achieve (B.11):

$$Y(z) = \frac{z^{-d}B(z)}{F(z)A(z)}U(z) + \frac{C(z)}{D(z)A(z)}W(z),$$
(B.11)

where A(z), B(z), C(z), D(z) and F(z) are polynomials in z. The poles and zeros of z correspond to the stochastic and deterministic parts of the system, and d is the delay of system transport. Taking the inverse of the z-transform of (B.11), (B.12) arises.

$$A(q^{-1})y(k) = q^{-d} \frac{B(q^{-1})}{F(q^{-1})} u(k) + \frac{C(q^{-1})}{D(q^{-1})} w(k),$$
(B.12)



Figure B.5: Infinite Impulse Response (IIR) Digital Filter [63].

In (B.12), q^{-1} is the delay operator so that $y(k)q^{-1} = y(k-1)$, w(k) is the white noise and $A(q^{-1})$, $B(q^{-1})$, $C(q^{-1})$, $D(q^{-1})$ and $F(q^{-1})$ are polynomials defined as follows:

$$A(q^{-1}) = 1 + a_1 q^{-1} + \ldots + a_{n_a} q^{-n_a}$$

$$B(q^{-1}) = b_0 + b_1 q^{-1} + \ldots + b_{n_b} q^{-n_b}$$

$$C(q^{-1}) = 1 + c_1 q^{-1} + \ldots + c_{n_c} q^{-n_c}$$

$$D(q^{-1}) = 1 + d_1 q^{-1} + \ldots + d_{n_d} q^{-n_d}$$

$$F(q^{-1}) = 1 + f_1 q^{-1} + \ldots + f_{n_f} q^{-n_f}$$
(B.13)

Depending on what are the particular values of $A(q^{-1})$, $B(q^{-1})$, $C(q^{-1})$, $D(q^{-1})$ and $F(q^{-1})$, we obtain different models in terms of system identification [65].

B.2.1 Infinite Impulse Response (IIR) and Finite Impulse Response (FIR)

The complex variable z^{-i} delays the signal in *i* sample intervals. Assuming that the transfer function contains *a* and *b* terms, we can write the resulting function at a *k* instant as

$$y(k) + a_1 y_{k-1} + \ldots + a_{n_a} y_{k-n_a} = b_0 x_k + b_1 x_{k-1} + \ldots + b_{n_b} x_{k-n_b}.$$

We illustrate the z-transform system of Fig. B.5, in which we can observe the input x_k and the ouput y_k .

Figure B.5 depicts an *Infinite Impulse Response* (IIR) Filter. This name is due to the existence of output feeding the input signal back, enforcing the output to be a non-zero value even when the input disappear. If feedback terms are set to zero, an = 0, then the system constitutes a Finite Impulse Response (FIR) model, whose output values depend only on the input signal values [63].

The IIR filter exists when at least one nonzero pole remains in the model not being canceled by any zero. Hence, the impulse response on the filter has an infinite duration, since a nonzero pole from A(z) provides output to regenerate input signals indefinitely.

A FIR filter is sometimes described as another particular type of another IIR filter. The *convolution sum* of the system impulse response h(k) and the input signal x(n) is given by

$$y(k) = \sum_{j=0}^{\infty} h(j)x(k-j) + w(k),$$
(B.14)

where w(k) is a white noise actuating over the system. Equation (B.14) is also called as Infinite Impulse Response (IIR), but the cause of the "infinite response" here is the infinite sum which involves infinite input terms. If the system is stable, there is an $M < \infty$ such that $h(k) < \epsilon, \forall k > M$. Hence, we can truncate equation (B.14) in order to obtain the finite impulse response (FIR), which is given by

$$y(k) = \sum_{j=0}^{M} h(j)u(k-j) + w(k),$$
(B.15)

where M is the number of elements of impulse response. The FIR model can be obtained from Eq. (B.12) by making $A(q^{-1}) = C(q^{-1}) = D(q^{-1}) = F(q^{-1}) = 1$. Here, $B(q^{-1})$ consists in an arbitrary polynomial of order M, that is to say, $n_b = M$, in conformity with Eq. (B.13). In this particular case, FIR model can be written as

$$y(k) = q^{-d}B(q^{-1})u(k) + w(k),$$
(B.16)

since, in (B.16), there is an implicit sum until the $n_b - th$ term. Nevertheless this filter is sometimes named as FIR in the bibliography, its underlying principle depends on a non-truncated input signal, therefore consisting rather in an "Infinite Input Signal". The infinite impulse is actually obtained, as said, by means of the output to feed the input filter back, with at least one nonzero pole which is not canceled by any zero.

The forecasted output, $\hat{y}(k|k-1,\theta)$, corresponds to

$$\hat{y}(k|k-1,\theta) = q^{-d}B(q^{-1})u(k).$$
 (B.17)

The output predictor may be represented in the regression form, with the regression vector, $\varphi(k)$, and the parameters vector, θ , defined as in

$$\varphi(k) = [u(k-d) \dots u(k-d-n_b)]^T$$
$$\theta = [b_0 \dots b_{n_b}]^T$$

and

$$\hat{y}(k|k-1,\theta) = \varphi(k)^T \theta.$$
(B.18)

Given a situation in which only the deterministic inputs are considered, a dynamic system cannot be described sufficiently just by means of a FIR model. However, if system is stable and its impulse response decays rapidly, normally a dynamic system can approximately place a FIR model, and the model's precision will rely on the order of polynomial $B(q^{-1})$.

B.2.2 Forecasted Error Model (FEM)

An important parameter that helps us to predict future values of output signal is the "Forecasted Error Model" (FEM), given the fact that output values are, in general, unpredictable given its stochastic characteristic. Therefore, it is important to know in the current instant, (k - 1), which is the most probable output in the next instant k in order to allow the error reduction. That is the reason we estimate the parameters vector of the process at each instant, so that the predicted error ϵ is reduced as possible.

The equation of FEM is given by [66]

$$\epsilon(k,\theta) = y(k) - \hat{y}(k|y^{k-1}, u^{k-1}, \theta),$$
(B.19)

where $\hat{y}(k|y^{k-1}, u^{k-1}, \theta)$ is the predicted next output value, y(k), as a function of the model vector parameters θ and of the set of the observed values until instant (k-1) of y and u, i. e., $y^{k-1} = [y(k-1), y(k-2), \dots, y(0)]$ and $u^{k-1} = [u(k-1), u(k-2), \dots, u(0)]$. One can notice that y^{k-1} and u^{k-1} store values since the beginning of the process, which might represent an additional exigency in terms of computational memory.

Applying the inverse of z-transform to (B.10),

$$y(k) = G(q^{-1}; \theta)u(k) + H(q^{-1}; \theta)w(k)$$
(B.20)

In (B.11), we have recognized the existence of the z^{-d} term in the numerator with U, the predictable series of input values, meaning that there is, at least, one delay instant between output and predictable input. In other words, in the first instant of the whole process, the output does not "see" the values coming from the input related to the predictable series, which only happens from the d - th instant on. Therefore, we can write $G(0, \theta) = 0$. If we make $G(q^{-1}; \theta) = q^{-1}G_1(q^{-1}; \theta)$, it follows that

$$y(k) = q^{-1}G_1(q^{-1};\theta)u(k) + H(q^{-1};\theta)w(k),$$

$$y(k) = [1-H^{-1}(q^{-1};\theta)]y(k) + q^{-1}H^{-1}(q^{-1};\theta)G_1(q^{-1};\theta)u(k) + w(k).$$

Defining $P(q^{-1};\theta) = q[1 - H^{-1}(q^{-1};\theta)]$ and $Q(q^{-1};\theta) = H^{-1}(q^{-1};\theta)G_1(q^{-1};\theta)$, we achieve

$$y(k) = P(q^{-1}; \theta)y(k-1) + Q(q^{-1}; \theta)u(k-1) + w(k).$$

We can obtain from (B.21), explicitly, the forecasted next output of the system for a particular value of the estimator vector $\theta = \hat{\theta}$:

$$\hat{y}(k|y^{k-1}, u^{k-1}, \theta) = P(q^{-1}; \theta)y(k-1) + Q(q^{-1}; \theta)u(k-1)$$
(B.21)

For each model hereafter exploited, the FEM method is used to provide us the general structure of the predictor vector θ . In the analyzed linear systems, at each step \hat{y} is compared with the real y and, given that the observations y(k) and u(k) are available all process long, θ is the equalization factor between predicted value and the real one. Thus, θ is the vector of coefficients which appear along Eq. (B.13) lines.

B.2.3 AR(X) and MA Models

B.2.3.1 Auto-Regressive Model (AR)

The AR structure is obtained by rewriting (B.13) making $B(q^{-1}) = C(q^{-1}) = D(q^{-1}) = F(q^{-1}) = 1$, being $A(q^{-1})$ a polynomial, such that

$$A(q^{-1})y(k) = w(k).$$
 (B.22)

In time domain, (B.22) can be written in the form

$$(1 + a_1q^{-1} + a_2q^{-2} + \ldots + a_{n_a}q^{-n_a})y(k) = w(k).$$
(B.23)

The AR structure, as well as the ARX and ARMAX, considers the last n_a values of y(k), including the present value. Therefore, this set of values is made as to comprise data into a moving-time window. As E[w(k)] = 0, the weights a_{n_i} are supposed to convert the average value of A(z)Y(z) into zero within the time-domain.

B.2.3.2 Moving Average Model (MA)

The MA model comes up when we make $A(q^{-1}) = B(q^{-1}) = D(q^{-1}) = F(q^{-1}) = 1$, being $C(q^{-1})$ any polynomial with order n_c . The white noise signal have successive values processed over the filter weights. The output in each instant is supposed to approach the instantaneous y(k) value:

$$y_k = C(q^{-1})w_k,$$

 $y_k = (1 + c_1q^{-1} + \ldots + c_nq^{-n})w_k.(B.24)$

If $w_k = 0$, we have $E[y_k] = 0$. It means that, in the k-instant, if the white noise is zero instaneously, y(k) is *expected* to be zero, that is to say, if n moments of $w_k = 0$, the average value of the several y_k values must approach zero as n increases.

B.2.3.3 Auto-Regressive Model with eXogeneous Inputs (ARX)

The ARX structure is obtained by rewriting Eq. (B.13) in order to make $C(q^{-1}) = D(q^{-1}) = F(q^{-1}) = 1$, being $A(q^{-1})$ and $B(q^{-1})$ arbitrary polynomials, resulting in

$$A(q^{-1})y(k) = q^{-d}B(q^{-1})u(k) + w(k).$$
(B.25)

Since the noise signal w(k) appears explicitly in (eq08), ARX model belongs to the class of models which present an *error term in the equation*. As well as FIR model, ARX can be written in the form

$$y(k) = \frac{q^{-d}B(q^{-1})}{A(q^{-1})}u(k) + \frac{1}{A(q^{-1})}w(k).$$
 (B.26)

In this case, the noise which is added to the output, $e(k) = \frac{w(k)}{A(q^{-1})}$, is not white; in this model, the white noise w(k) is filtered along an auto-regressive model.

An alternative way of representing ARX model structure correspond to establish $G(q^{-1}; \theta)$ and $H(q^{-1}; \theta)$ as

$$G(q^{-1},\theta) = q^{-d} \frac{B(q^{-1})}{A(q^{-1})} \qquad \qquad H(q^{-1},\theta) = \frac{1}{A(q^{-1})}, \tag{B.27}$$

in which the predictor can be written in the form

$$\hat{y}(k|k-1,\theta) = q^{-d}B(q^{-1})u(k) + [1 - A(q^{-1})]y(k) =$$

= $\varphi^{T}(k)\theta$ (B.28)

with

$$\varphi(k) = [y(k-1) \dots y(k-n_a), u(k-d) \dots u(k-d-n_b)]^T (B.29)$$
$$\theta = [-a_1 \dots - a_{n_a}, b_0 \dots b_{n_b}]^T$$

(B.30)

From (B.30), one can notice that there is only an algebric relation between the forecasted output and the data comprising past inputs and measured outputs.

B.2.3.4 Auto-Regressive Moving Average with eXogeneous Inputs (ARMAX)

ARMAX models are produced when, in (B.13), $D(q^{-1}) = F(q^{-1}) = 1$ and $A(q^{-1})$, $B(q^{-1})$ and $C(q^{-1})$ are arbitrary polynomials, resulting in

$$A(q^{-1})y(k) = q^{-d}B(q^{-1})u(k) + C(q^{-1})w(k).$$
(B.31)

Like ARX model, ARMAX is said to bear "error in equation". Once more, the nonwhite noise (perturbation) is modeled as the output of an ARMA filter, in which the input is the white noise. In face of the A, B, C, D, E and F values, the polynomials G and H are chosen as below:

$$G(q^{-1},\theta) = q^{-d} \frac{B(q^{-1})}{A(q^{-1})} \qquad H(q^{-1},\theta) = \frac{C(q^{-1})}{A(q^{-1})},$$
(B.32)

from where we obtain the predictor:

$$\hat{y}(k|k-1,\theta) = q^{-d} \frac{B(q^{-1})}{C(q^{-1})} u(k) + \left(1 - \frac{A(q^{-1})}{C(q^{-1})}\right) y(k)$$
$$\hat{y}(k|k-1,\theta) = q^{-d} B(q^{-1}) u(k) + (C(q^{-1}) - A(q^{-1})) y(k) - (1 - C(q^{-1})) \hat{y}(k|k-1,\theta)$$
$$\hat{y}(k|k-1,\theta) = q^{-d} B(q^{-1}) u(k) + [1 - A(q^{-1})] y(k) + [C(q^{-1}) - 1] \epsilon(k,\theta)$$
$$= \varphi^T(k,\theta) \theta, (B.33)$$

where $\epsilon(k, \theta) = y(k) - \hat{y}(k|k-1, \theta)$ corresponds to the *forecast error* or *residue*. The regression vector and the parameter vector are defined, respectively, as

$$\varphi(k,\theta) = [y(k-1)\dots y(k-n_a), u(k-d)\dots u(k-d-n_b), \epsilon(k,\theta)\dots \epsilon(k-n_c,\theta)]^T,$$
(B.34)

and

$$\theta = [-a_1 \dots - a_{n_a}, b_0 \dots b_{n_b}, c_1, \dots c_{n_c}]^T (B.35)$$

In this case, the predictor model is specified by the roots of polynomial C(z), which must be inside the unitary circle on the z-plan to assure that the predictor is stable. Moreover, the presence of the poles in this predictor model causes the regression vector to depend on the model parameters, meaning that there exists a non-linearity in the predictor equation (B.33) with regards to θ .

Appendix C Examples of Peak Load Reductions over the Substations in Leipzig Map

In this Appendix, figures depict several examples of reductions for peak load on Leipzig districts.

In these simulations, we consider that the three emergent sources start to generate from the 1st week of 2002, therefore w = 156. Only parameters c and a vary. Predictions are based on ARMAX filter, whose parameters are [10 10 10 1]. As in the Chapter 5, there is a table into each figure containing the parameters c and a for each source, as well as the year whose last week is forecasted and the color code exposed on the Fig. 5.13 is inserted.

We should remember that c is the initial amount of generators for each source, such that it expresses units for antennas and eels, and m² for PV panels. For the sake of simplicity, all Power Substations present the same c and a related to the sources in each figure.



Figure C.1: Peak load reduction by district for the last week of 2006, c = 500 and a = 0.6, antennas only.

Peak Relief	Source	Irce Anten		nnas Indoor			door	Eels	
	Substation	С	а	с	а	С	а	с	а
5,06%	A	1500	1	0	0,3	0	0,3	0	0,3
3,12%	В	1500	1	0	0,3	0	0,3	0	0,3
1,83%	С	1500	1	0	0,3	0	0,3	0	0,3
4,77%	D	1500	1	0	0,3	0	0,3	0	0,3
4,91%	E	1500	1	0	0,3	0	0,3	0	0,3
2,62%	F	1500	1	0	0,3	0	0,3	0	0,3
2,74%	G	1500	1	0	0,3	0	0,3	0	0,3
1,98%	н	1500	1	0	0,3	0	0,3	0	0,3
Year	2006	T	Peak Load Relief - Percentages						
		-	10%	30%	60%	100%	more		
			-	¥					
exp(.)	0,10	0,20	0,40	0,60	0,80	1,00	1,50		
Annual increase .:	10,52%	22,14%	0,30%	82,21%	122,55%	171,83%	348,17%		



Figure C.2: Peak load reduction by district for the last week of 2006, c = 1500 and a = 1.0, antennas only.



Figure C.3: Peak load reduction by district for the last week of 2006, c = 1500 and a = 1.0 for antennas, c = 200 and a = 1.0 for eels.





Figure C.4: Peak load reduction by district for the last week of 2006, c = 300 and a = 1.0, eels only.



Figure C.5: Peak load reduction by district for the last week of 2006, c = 300 and a = 1.1, eels only.





Figure C.6: Peak load reduction by district for the last week of 2006, c = 300 and a = 1.2, eels only.



Figure C.7: Peak load reduction by district for the last week of 2006, c = 50 and a = 0.6, indoor panels only.





Figure C.8: Peak load reduction by district for the last week of 2006, c = 50 and a = 0.7, indoor panels only.



Figure C.9: Peak load reduction by district for the last week of 2006, c = 50 and a = 0.8, indoor panels only.

Peak Relief	Source	ource Ante		nnas Indoor		Outdoor		Eels	
	Substation	С	а	С	а	С	а	с	а
29,77%	A	0	0,3	50	0,9	0	0,3	0	0,3
18,37%	В	0	0,3	50	0,9	0	0,3	0	0,3
10,79%	С	0	0,3	50	0,9	0	0,3	0	0,3
28,10%	D	0	0,3	50	0,9	0	0,3	0	0,3
28,88%	E	0	0,3	50	0,9	0	0,3	0	0,3
15,41%	F	0	0,3	50	0,9	0	0,3	0	0,3
16,12%	G	0	0,3	50	0,9	0	0,3	0	0,3
11,67%	Н	0	0,3	50	0,9	0	0,3	0	0,3
Year	2006	T I	Peak Load Relief - Percentages						
			10%	30%	60%	100%	more		
exp(.)	0,10	0,20	0,40	0,60	0,80	1,00	1,50		
Annual increase.:	10,52%	22,14%	0,30%	82,21%	122,55%	171,83%	348,17%		



Figure C.10: Peak load reduction by district for the last week of 2006, c = 50 and a = 0.9, indoor panels only.



Figure C.11: Peak load reduction by district for the last week of 2006, c = 50 and a = 1.0, indoor panels only.





Figure C.12: Peak load reduction by district for the last week of 2006, c = 50 and a = 1.2, indoor panels only.



Figure C.13: Peak load reduction by district for the last week of 2005, c = 50 and a = 1.2, indoor panels only.




Figure C.14: Peak load reduction by district for the last week of 2005, c = 50 and a = 1.0, indoor panels only.



Figure C.15: Peak load reduction by district for the last week of 2006, c = 5 and a = 0.2, outoor panels only.

Peak Relief	Source	Ante	Antennas		Indoor		door	Eels	
	Substation	с	а	С	а	С	а	с	а
30,53%	A	0	0,3	0	0,3	5	0,3	0	0,
18,83%	В	0	0,3	0	0,3	5	0,3	0	0,3
11,07%	С	0	0,3	0	0,3	5	0,3	0	0,3
28,81%	D	0	0,3	0	0,3	5	0,3	0	0,3
29,61%	E	0	0,3	0	0,3	5	0,3	0	0,3
15,80%	F	0	0,3	0	0,3	5	0,3	0	0,3
16,53%	G	0	0,3	0	0,3	5	0,3	0	0,3
11,96%	н	0	0,3	0	0,3	5	0,3	0	0,3
Year	2006		Peak Load Relief - Percentages						
		-	10%	30%	60%	100%	more		
exp(.)	0,10	0,20	0,40	0,60	0,80	1,00	1,50		
Annual increase.:	10,52%	22,14%	0,30%	82,21%	122,55%	171,83%	348,17%		



Figure C.16: Peak load reduction by district for the last week of 2006, c = 5 and a = 0.3, outoor panels only.



Figure C.17: Peak load reduction by district for the last week of 2006, c = 5 and a = 0.4, outoor panels only.



Figure C.18: Peak load reduction by district for the last week of 2006, c = 10 and a = 0.4, outoor panels only.



Figure C.19: Peak load reduction by district for the last week of 2006, c = 15 and a = 0.4, outoor panels only.





Figure C.20: Peak load reduction by district for the last week of 2006, c = 20 and a = 0.4, outoor panels only.



Figure C.21: Peak load reduction by district for the last week of 2006, c = 20 and a = 0.3, outoor panels only.

Peak Relief	Source An		ennas Ind		door Out		door	Eels	
	Substation	С	а	С	а	С	а	С	а
86,01%	А	0	0,3	0	0,3	20	0,3	0	0,3
53,92%	В	0	0,3	0	0,3	20	0,3	0	0,3
32,35%	С	0	0,3	0	0,3	20	0,3	0	0,3
81,47%	D	0	0,3	0	0,3	20	0,3	0	0,3
87,83%	E	0	0,3	0	0,3	20	0,3	0	0,3
46,33%	F	0	0,3	0	0,3	20	0,3	0	0,3
47,27%	G	0	0,3	0	0,3	20	0,3	0	0,3
36,16%	Н	0	0,3	0	0,3	20	0,3	0	0,3
Year	2005)	Peak Load Relief - Percentages						
			10%	30%	60%	100%	more		
exp(.)	0,10	0,20	0,40	0,60	0,80	1,00	1,50		
Annual increase.:	10,52%	22,14%	0,30%	82,21%	122,55%	171,83%	348,17%		



Figure C.22: Peak load reduction by district for the last week of 2005, c = 20 and a = 0.3, outoor panels only.



Figure C.23: Peak load reduction by district for the last week of 2005, c = 15 and a = 0.3, outoor panels only.